



**GaNFast™ Power IC**  
with GaNSense™ Technology

**1. Features**

**GaNFast™ Power IC**

- Monolithically-integrated gate drive
- Wide  $V_{CC}$  range (10 to 30 V)
- Programmable turn-on  $dV/dt$
- 200 V/ns  $dV/dt$  immunity
- 800 V Transient Voltage Rating
- 700 V Continuous Voltage Rating
- Low 330 m $\Omega$  resistance
- Zero reverse recovery charge
- 2 MHz operation

**GaNSense™ Technology**

- Integrated loss-less current sensing
- Short-circuit protection
- Over-temperature protection
- Autonomous low-current standby mode
- Auto-standby mode input

**Small, low-profile SMT QFN**

- 6 x 8 mm footprint, 0.85 mm profile
- Minimized package inductance
- Large cooling pad

**Sustainability**

- RoHS, Pb-free, REACH-compliant
- Up to 40% energy savings vs Si solutions
- System level 4kg CO<sub>2</sub> Carbon Footprint reduction

**Product Reliability**

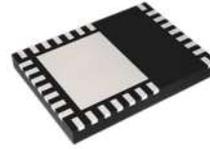
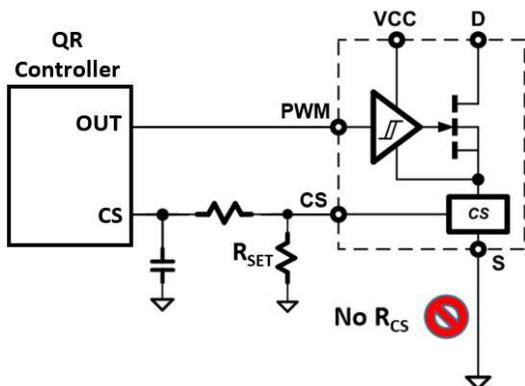
- 20-year limited product warranty  
(see Section 14 for details)

**2. Topologies / Applications**

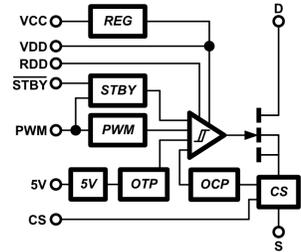
- AC-DC, DC-DC, DC-AC
- High frequency operation up to 2 MHz
- QR flyback, AHB, Buck, Boost, Half bridge, Full bridge, LLC resonant, Class D, PFC
- Wireless power, Solar Micro-inverters
- LED lighting, TV SMPS, Server, Telecom

**4. Typical Application Circuits**

Loss-less Current Sensing



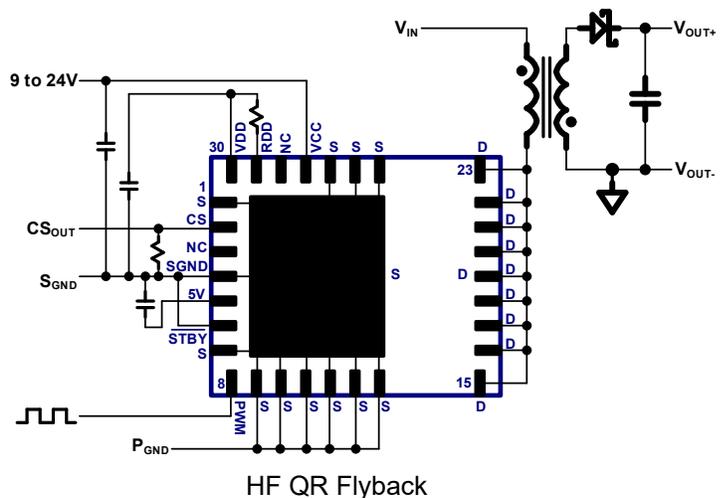
QFN 6 x 8 mm



Simplified schematic

**3. Description**

This GaNFast™ power IC integrates a high performance eMode GaN FET with integrated gate drive to achieve unprecedented high-frequency and high efficiency operation. GaNSense™ technology is also integrated which enables real-time, accurate sensing of voltage, current and temperature to further improve performance and robustness not achieved by any discrete GaN or discrete silicon device. GaNSense™ enables integrated loss-less current sensing which eliminates external current sensing resistors and increases system efficiency. GaNSense™ also enables short circuit and over-temperature protection to increase system robustness, while auto-standby mode increases light, tiny & no-load efficiency. These GaN ICs combine the highest  $dV/dt$  immunity, high-speed integrated drive and industry-standard low-profile, low-inductance, SMT QFN packaging to enable designers to achieve simple, quick and reliable solutions. Navitas' GaN IC technology extends the capabilities of traditional topologies such as flyback, half-bridge, buck/boost, LLC and other resonant converters to reach MHz+ frequencies with very high efficiencies and low EMI to achieve unprecedented power densities at a very attractive cost structure.



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## 6. Specifications

### 6.1. Absolute Maximum Ratings<sup>(1)</sup>

(with respect to Source (pad) unless noted)

SYMBOL	PARAMETER	MAX	UNITS
$V_{DS (CONT)}$	Drain-to-Source Voltage	-7 to +700	V
$V_{DS (TRAN)}$	Transient Drain-to-Source Voltage <sup>(2)</sup>	800	V
$V_{CC}$	Supply Voltage	30	V
$V_{DD}$	Drive Supply Voltage	7	V
$R_{DD}$	Input Voltage	7	V
$V_{STBY}$	Auto-Standby Mode Pin Voltage	-0.6 to +20 or $V_{CC}$	V
$V_{5V}$	5 V Pin Voltage	6	V
$V_{PWM}$	PWM Input Pin Voltage	-0.6 to +20 or $V_{CC}$	V
$V_{CS}$	CS Pin Voltage	5.3	V
$I_D$	Continuous Drain Current (@ $T_C = 100^\circ\text{C}$ )	4	A
$I_D \text{ PULSE}$	Pulsed Drain Current (10 $\mu\text{s}$ @ $T_J = 25^\circ\text{C}$ )	8	A
dV/dt	Slew Rate	200	V/ns
$T_J$	Junction Temperature	-55 to 150	$^\circ\text{C}$
$T_{STOR}$	Storage Temperature	-55 to 150	$^\circ\text{C}$

(1) Absolute maximum ratings are stress ratings; devices subjected to stresses beyond these ratings may cause permanent damage.

(2)  $V_{DS (TRAN)}$  allows for surge ratings during non-repetitive events that are <100 $\mu\text{s}$  (for example start-up, line interruption) and repetitive events that are <400ns (for example repetitive leakage inductance spikes). Refer to Section 8.9 for detailed recommended design guidelines.

## 6.2. Recommended Operating Conditions<sup>(3)</sup>

SYMBOL	PARAMETER	MIN	TYP	MAX	UNITS
$V_{CC}$	Supply Voltage	9		24	V
$V_{PWM}$	PWM Input Pin Voltage	0	5	15 or $V_{CC}$	V
$V_{STBY}$	Auto-Standby Mode Input Pin Voltage	0	5	15 or $V_{CC}$	V
$R_{DD}$	Gate drive turn-on current set resistor	10	50		$\Omega$
$T_J$	Operating Junction Temperature	-40		125	$^{\circ}\text{C}$

(3) Exposure to conditions beyond maximum recommended operating conditions for extended periods of time may affect device reliability.

## 6.3. ESD Ratings

SYMBOL	PARAMETER	MAX	UNITS
HBM	Human Body Model (per JESD22-A114)	2,000	V
CDM	Charged Device Model (per JESD22-C101F)	1,000	V

## 6.4. Thermal Resistance

SYMBOL	PARAMETER	TYP	UNITS
$R_{\theta JC}^{(4)}$	Junction-to-Case	2.34	$^{\circ}\text{C}/\text{W}$
$R_{\theta JA}^{(4)}$	Junction-to-Ambient	40	$^{\circ}\text{C}/\text{W}$

(4)  $R_{\theta}$  measured on DUT mounted on 1 square inch 2 oz Cu (FR4 PCB)

## 6.5. Electrical Characteristics (1)

Typical conditions:  $V_{DS}=400V$ ,  $V_{CC}=15V$ ,  $F_{SW}=1MHz$ ,  $T_{AMB}=25^{\circ}C$ ,  $I_D=2A$  (unless otherwise specified)

SYMBOL	PARAMETER	MIN	TYP	MAX	UNITS	CONDITIONS
<b>VCC &amp; VDD Supply Characteristics</b>						
$V_{CCUV+}$	$V_{CC}$ UVLO Rising Turn-On Threshold	8	8.5	9.3	V	
$V_{CCUV-}$	$V_{CC}$ UVLO Falling Turn-Off Threshold		7.3		V	
$I_{OCC-STBY}$	$V_{CC}$ Standby Current		275		$\mu A$	$\overline{STBY}=0V$
$I_{OCC}$	$V_{CC}$ Quiescent Current		0.55	0.8	mA	$V_{PWM}=0V$ , $\overline{STBY}=5V$ OR $V_{PWM}=5V$ , $\overline{STBY}=0V$
$I_{OCC-SW}$	$V_{CC}$ Operating Current		2.1		mA	$F_{SW}=1MHz$ , $V_{DS}=\text{Open}$
$V_{DD}$	$V_{DD}$ Supply Voltage	5.9	6.2	6.6	V	$V_{CC}=15V$ , $V_{PWM}=0V$ $\overline{STBY}=5V$
<b>5V Output (5V pin)</b>						
$V_{5V}$	5V Output Voltage	4.4	5	5.5	V	$\overline{STBY}=5V$
<b>Input Logic Characteristics (PWM, <math>\overline{STBY}</math>)</b>						
$V_{LOGIC-H}$	Input Logic High Threshold (rising edge)		2.5	2.8	V	
$V_{LOGIC-L}$	Input Logic Low Threshold (falling edge)	1.1	1.2		V	
$V_{LOGIC-HYS}$	Input Logic Hysteresis		1.3		V	
<b>Switching Characteristics</b>						
$F_{SW}$	Switching Frequency			2	MHz	$R_{dd}=10\Omega$
$t_{PW}$	Pulse width	30			ns	
$T_{ON}$	Turn-on Propagation Delay		25		ns	Fig 1
$T_{OFF}$	Turn-off Propagation Delay		22		ns	Fig 1
$T_R$	Drain rise time		9		ns	Fig 1
$T_F$	Drain fall time		7		ns	Fig 1

## 6.6. Electrical Characteristics (2, cont.)

Typical conditions:  $V_{DS}=400V$ ,  $V_{CC}=15V$ ,  $F_{SW}=1MHz$ ,  $T_{AMB}=25^{\circ}C$ ,  $I_D=2A$  (unless otherwise specified)

SYMBOL	PARAMETER	MIN	TYP	MAX	UNITS	CONDITIONS
<b>Current Sense Characteristics (CS pin)</b>						
$I_{CS}$	CS Pin Output Current	1.16	1.25	1.34	mA	$V_{PWM} = 5 V$ , $I_{DS} = 2.2 A$
Offset	CS Output Offset		18		$\mu A$	$V_{PWM} = 5 V$ , $I_{DS} = 0 A$
$t_{CSDLY}$	CS Pin Delay (from $I_{DS}$ to $V_{CS}$ , at 10% rated current)		55		ns	$di/dt = 40 A/\mu s$ , $R_{SET} = 400 \text{ Ohm}$ , $C_{CS} = 25 \text{ pF}$
<b>Over-Current Protection</b>						
OCP <sub>TH</sub>	OCP Threshold Voltage ( $V_{CS}$ Pin)		1.9		V	
<b>Standby Mode Characteristics</b>						
$t_{TO\_STBY}$	Time Out Delay to Enter Standby Mode		90		$\mu s$	$V_{PWM} = 0 V$
$t_{ON\_FP}$	First Pulse Propagation Delay		30		ns	$V_{PWM} = 5 V$ pulse, $\overline{STBY} = 0 V$
<b>Over-Temperature Protection</b>						
$T_{OTP+}$	OTP Shutdown Threshold		165		$^{\circ}C$	
$T_{OTP\_HYS}$	OTP Restart Hysteresis		60		$^{\circ}C$	

## 6.7. Electrical Characteristics (3, cont.)

Typical conditions:  $V_{DS}=400V$ ,  $V_{CC}=15V$ ,  $F_{SW}=1MHz$ ,  $T_{AMB}=25^{\circ}C$ ,  $I_D=2A$  (unless otherwise specified)

SYMBOL	PARAMETER	MIN	TYP	MAX	UNITS	CONDITIONS
<b>GaN FET Characteristics</b>						
$I_{DSS}$	Drain-Source Leakage Current		0.15	23	$\mu A$	$V_{DS} = 700 V$ , $V_{PWM} = 0 V$
$I_{DSS}$	Drain-Source Leakage Current, $T_C = 150^{\circ}C$		11		$\mu A$	$V_{DS} = 700V$ , $V_{PWM} = 0V$ , $T_C = 150^{\circ}C$
$R_{DS(ON)}$	Drain-Source Resistance		330	460	$m\Omega$	$V_{PWM} = 5 V$ , $I_D = 2 A$
$V_{SD}$	Source-Drain Reverse Voltage		3.5	5	V	$V_{PWM} = 0 V$ , $I_{SD} = 2 A$
$Q_{OSS}$	Output Charge		9.7		nC	
$Q_{RR}$	Reverse Recovery Charge		0		nC	
$C_{OSS}$	Output Capacitance		12		pF	$V_{DS} = 400 V$ , $V_{PWM} = 0 V$
$C_{O(er)}$ (Note 1)	Effective Output Capacitance, Energy Related		16		pF	$V_{DS} = 400 V$ , $V_{PWM} = 0 V$
$C_{O(tr)}$ (Note 2)	Effective Output Capacitance, Time Related		24		pF	$V_{DS} = 400 V$ , $V_{PWM} = 0 V$

(Note 1):  $C_{O(er)}$  is a fixed capacitance that gives the same stored energy as  $C_{OSS}$  while  $V_{DS}$  is rising from 0 to 400 V

(Note 2):  $C_{O(tr)}$  is a fixed capacitance that gives the same charging time as  $C_{OSS}$  while  $V_{DS}$  is rising from 0 to 400 V

## 6.8. Switching Waveforms

( $T_C = 25\text{ }^\circ\text{C}$  unless otherwise specified)

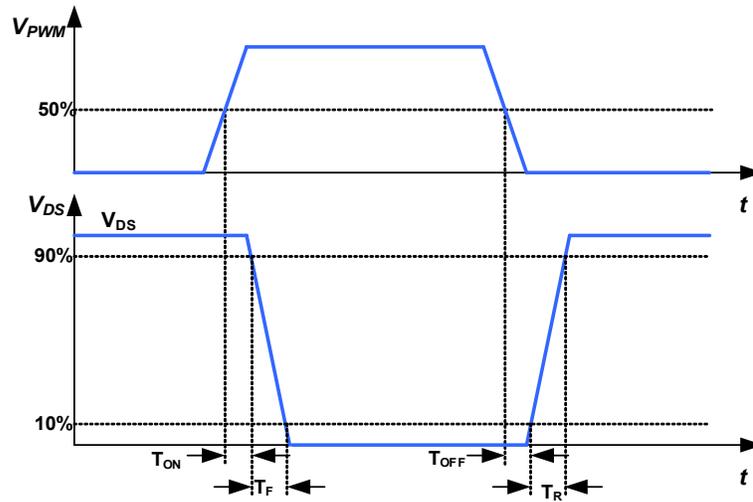


Fig. 1. Propagation Delay and Rise/fall Time Definition

### 6.9. Characteristic Graphs

(GaN FET,  $T_C = 25\text{ }^\circ\text{C}$  unless otherwise specified)

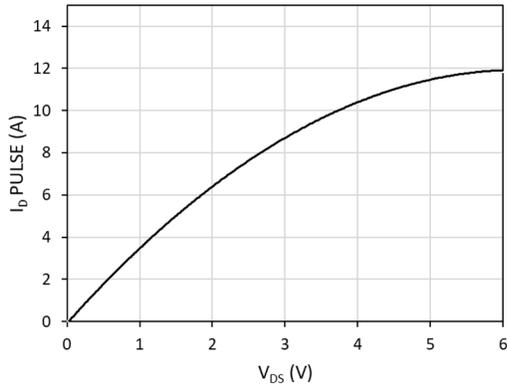


Fig. 2. Pulsed Drain current ( $I_D$  PULSE) vs. drain-to-source voltage ( $V_{DS}$ ) at  $T = 25\text{ }^\circ\text{C}$

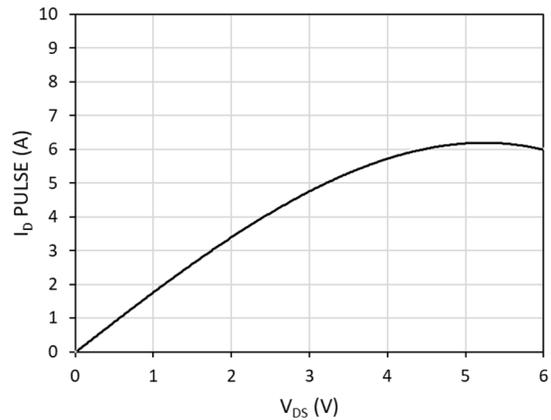


Fig. 3. Pulsed Drain current ( $I_D$  PULSE) vs. drain-to-source voltage ( $V_{DS}$ ) at  $T = 125\text{ }^\circ\text{C}$

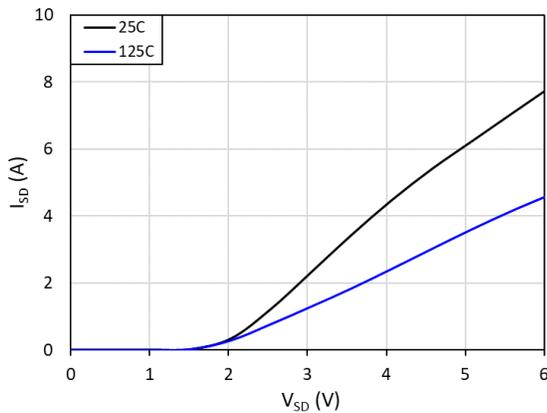


Fig. 4. Source-to-drain reverse conduction current vs. source-drain voltage

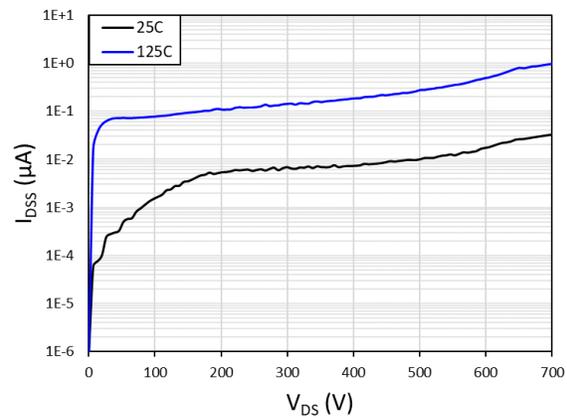


Fig. 5. Drain-to-source leakage current ( $I_{DSS}$ ) vs. drain-to-source voltage ( $V_{DS}$ )

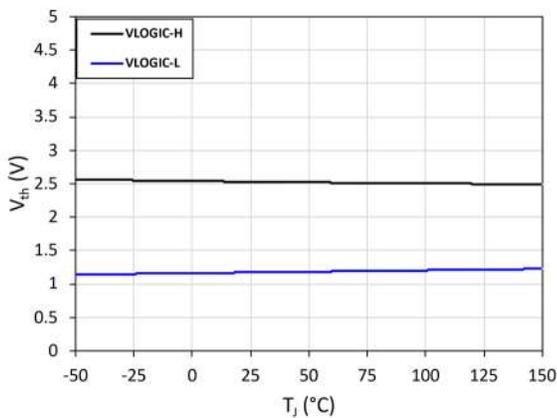


Fig. 6.  $V_{LOGIC-H}$  and  $V_{LOGIC-L}$  vs. junction temperature ( $T_J$ )

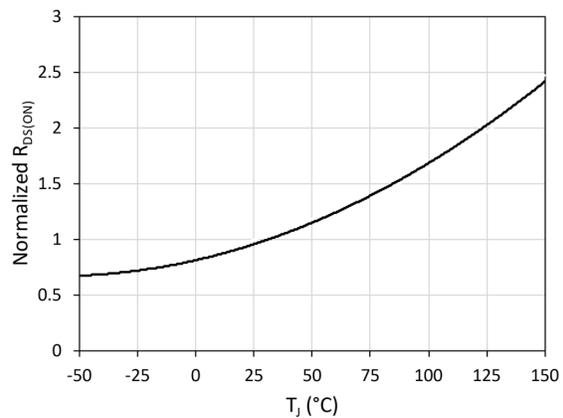


Fig. 7. Normalized on-resistance ( $R_{DS(ON)}$ ) vs. junction temperature ( $T_J$ )

**Characteristic Graphs (Cont.)**

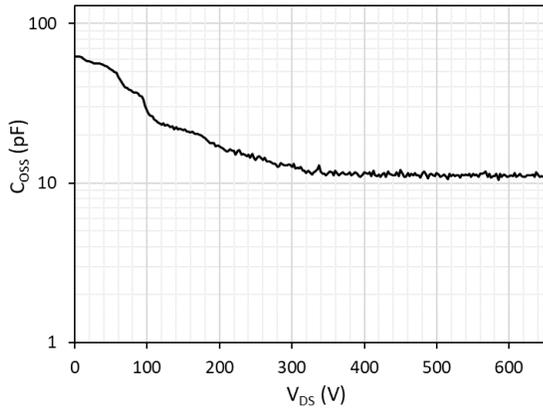


Fig. 8. Output capacitance ( $C_{OSS}$ ) vs. drain-to-source voltage ( $V_{DS}$ )

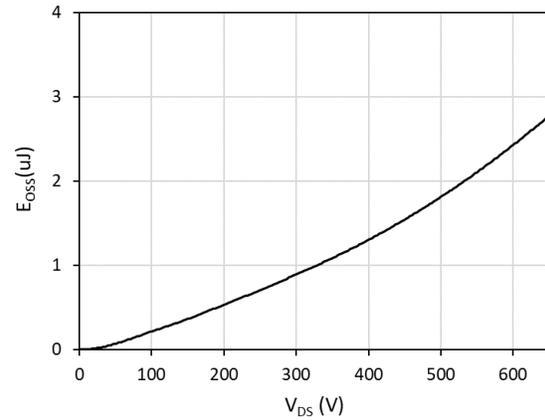


Fig. 9. Energy stored in output capacitance ( $E_{OSS}$ ) vs. drain-to-source voltage ( $V_{DS}$ )

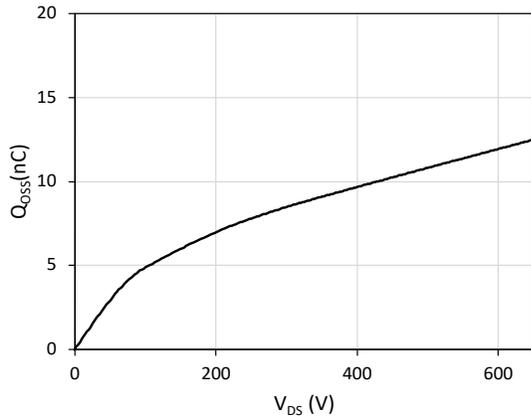


Fig. 10. Charge stored in output capacitance ( $Q_{OSS}$ ) vs. drain-to-source voltage ( $V_{DS}$ )

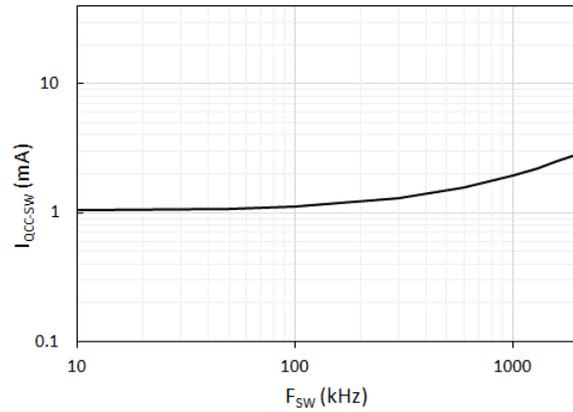


Fig. 11.  $V_{CC}$  operating current ( $I_{QCC-SW}$ ) vs. operating frequency ( $F_{SW}$ )

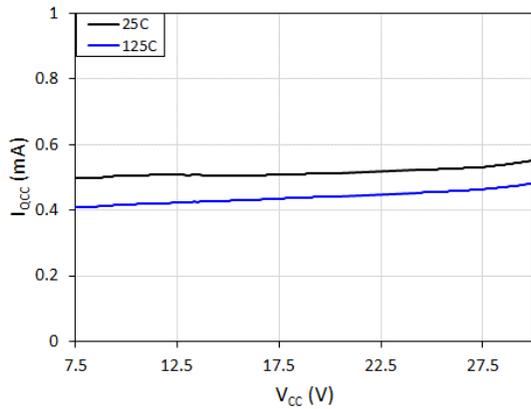


Fig. 12.  $V_{CC}$  quiescent current ( $I_{QCC}$ ) vs. supply voltage ( $V_{CC}$ )

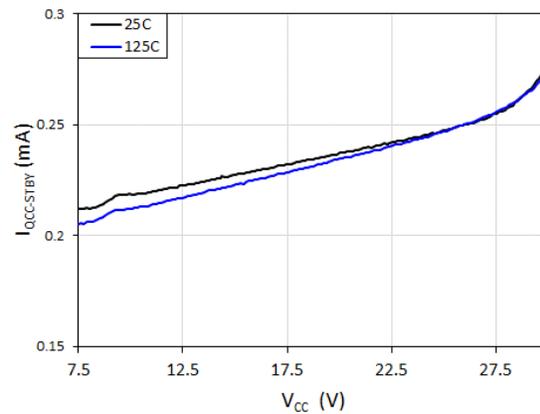


Fig. 13.  $V_{CC}$  stand-by quiescent current ( $I_{QCC-STBY}$ ) vs. supply voltage ( $V_{CC}$ )

**Characteristic Graphs (Cont.)**

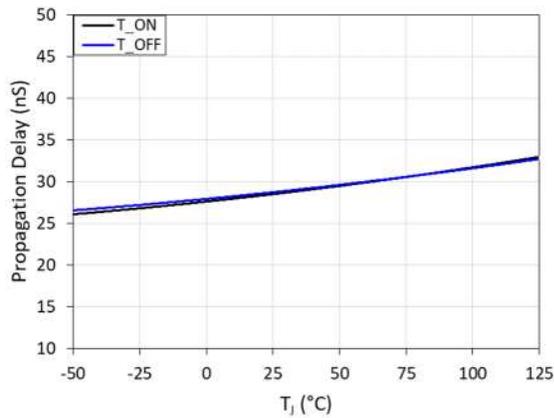


Fig. 14. Propagation delay ( $T_{ON}$  and  $T_{OFF}$ ) vs. junction temperature ( $T_J$ )

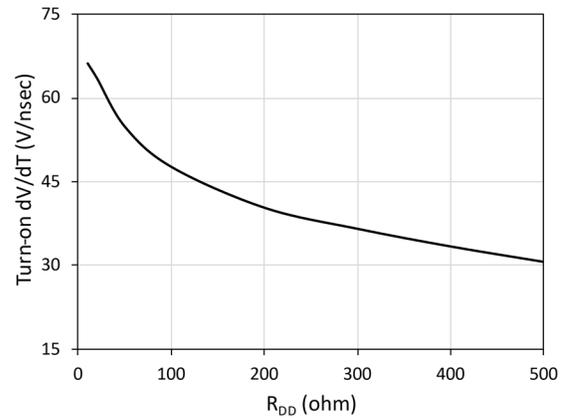


Fig. 15. Slew rate ( $dV/dt$ ) vs. gate drive turn-on current set resistance ( $R_{DD}$ ) at  $T = 25\text{ °C}$

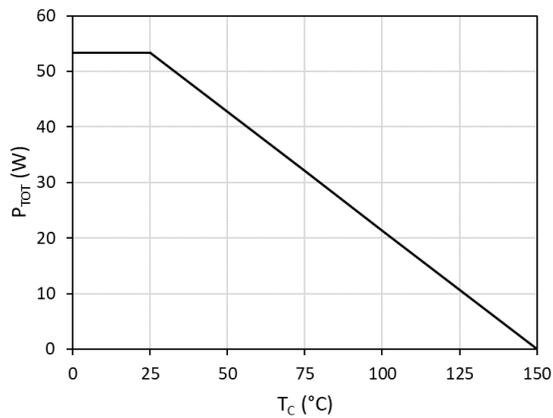


Fig. 16. Power dissipation ( $P_{TOT}$ ) vs. case temperature ( $T_C$ )

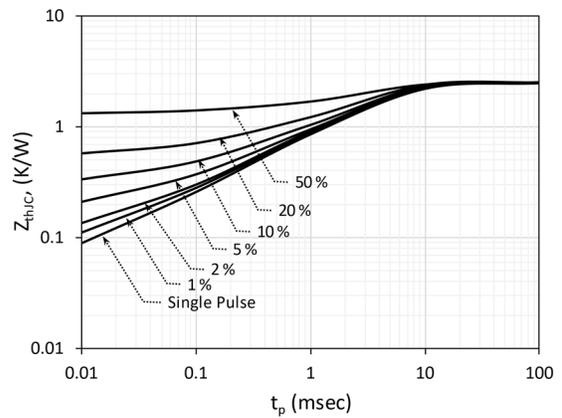


Fig. 17. Max. thermal transient impedance ( $Z_{thJC}$ ) vs. pulse width ( $t_p$ )

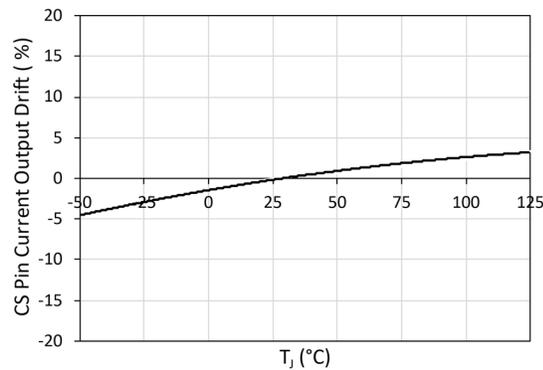


Fig. 18. CS Pin Current Output Drift vs. case temperature ( $T_C$ )

## 7. Pin Configurations and Functions

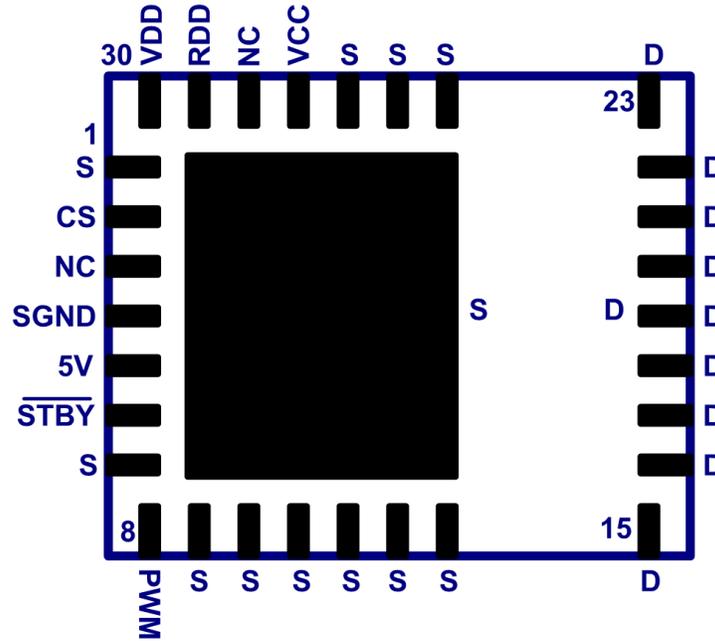


Fig. 19. Package Top View

Pin		I/O <sup>(1)</sup>	Description
Number	Symbol		
4	S <sub>GND</sub>	I	Connect directly to Source PAD
30	V <sub>DD</sub>	I	Gate drive supply voltage.
29	R <sub>DD</sub>	I/O	Gate drive turn-on current set pin (using R <sub>DD</sub> for dV/dt control)
2	CS	O	GaN FET I <sub>DS</sub> current sensing set pin. Internal current source and external resistor sets current measurement level. External resistor reference is S <sub>GND</sub> .
5	5V	O	5V internal supply voltage. Connect 10 nF capacitor between 5V pin and PAD
15 - 23	D	P	Drain of power FET
27	V <sub>CC</sub>	P	IC supply voltage. Provided externally.
8	PWM	I	PWM input (wrt S <sub>GND</sub> ).
6	STBY	I	Auto-standby mode input. Connect to S <sub>GND</sub> to enable auto-standby
1, 7, 9-14, 24-26, PAD	S	O, G	Source of power FET & IC supply ground. Metal pad on bottom of package.
3, 28	NC	NC	Must not connect any NC pins to Source

(5) I = Input, O = Output, P = Power, G = Ground, NC = No Connect



### 8.1. GaN Power IC Connections and Component Values (Cont.)

The typical connection diagram for a half-bridge configuration is shown in Fig. 21. The schematic includes a low-side GaN IC with the Source connected to P<sub>GND</sub> and a high-side GaN IC with the Drain connected to V<sub>IN</sub>. The Source of the high-side GaN IC is connected to the Drain of the low-side GaN IC to form the half-bridge switched node output (V<sub>SW</sub>). The external components for each GaN IC are placed directly next to their respective pins. A low-side supply voltage (9-24V) is required for V<sub>CC</sub> of the low-side GaN IC and a bootstrap diode and resistor (D<sub>BOOT</sub>, R<sub>BOOT</sub>) provide the necessary high-side supply voltage from V<sub>CC</sub> to V<sub>B</sub>. The low-side GaN IC includes low-side PWM input (PWM<sub>L</sub>) and a low-side current sense output (CS<sub>OUTL</sub>). The high-side GaN IC requires a high-side ‘floating’ PWM input signal for driving the PWM<sub>H</sub> input (typically provided by an external half-bridge driver, not shown). The R<sub>BOOT</sub> and C<sub>VB</sub> should be carefully selected such that the time constant of the R<sub>BOOT</sub> x C<sub>VB</sub> is greater than 0.5 μs (5 Ω x 100 nF).

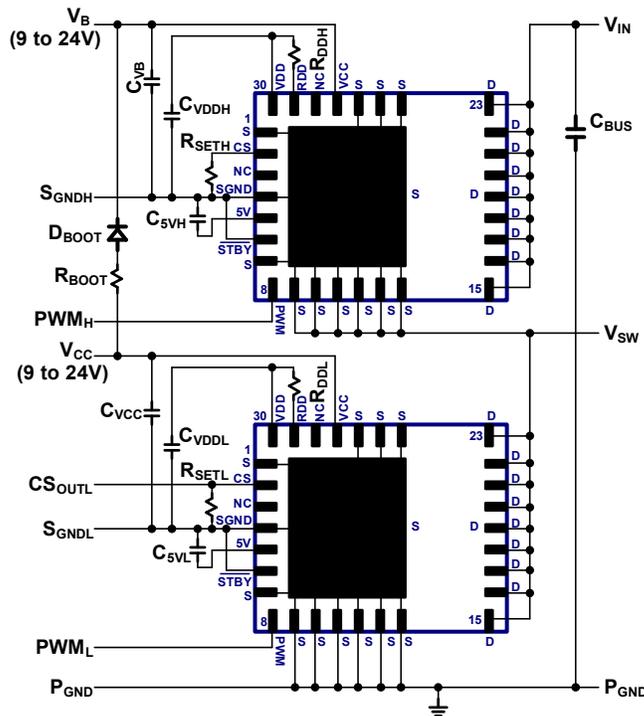


Fig. 21. Half-Bridge Configuration

SYM	DESCRIPTION	TYP	UNITS
C <sub>VCC</sub>	V <sub>CC</sub> supply capacitor	0.1	μF
C <sub>VB</sub>	V <sub>DB</sub> supply capacitor	100	nF
C <sub>VDDL,H</sub>	V <sub>DD</sub> supply capacitor	10	nF
R <sub>RDDL,H</sub>	Gate drive turn-on current set resistor	50	Ω
R <sub>SETL,H</sub>	Current sense amplitude set resistor	Depends on system design (See Section 8.6. Equation 1)	
C <sub>5VL,H</sub>	5V supply capacitor	10 (Max)	nF
R <sub>BOOT</sub>	Bootstrap resistor	5	Ω

Table II. Recommended component values (typical only)

## 8.2. UVLO Mode

This GaN Power IC includes under-voltage lockout (UVLO) circuits for properly disabling all of the internal circuitry when  $V_{CC}$  is below the  $V_{CCUV+}$  threshold (8.5V, typical) and  $V_{DD}$  is below the  $V_{DDUV+}$  threshold (4.5V, typical). During UVLO Mode, the internal gate drive and power FET are disabled and  $V_{CC}$  consumes a low quiescent current (275 $\mu$ A, typical). As the  $V_{CC}$  supply voltage increases (Fig. 22), the voltage at the  $V_{DD}$  pin also increases and exceeds  $V_{DDUV+}$ . The  $V_{DD}$  voltage continues to increase with  $V_{CC}$  until it gets limited to a constant voltage level (6.2V, typical) by the internal regulator. The  $V_{CC}$  voltage continues to increase until it exceeds  $V_{CCUV+}$  and the IC enters Normal Operating Mode. The gate drive is enabled and the control signal at the PWM input turns the internal GaN power FET on and off normally. During system power off, when  $V_{CC}$  decreases below the  $V_{CCUV-}$  threshold (7.3V, typical), the gate drive is disabled and the IC enters UVLO Mode.

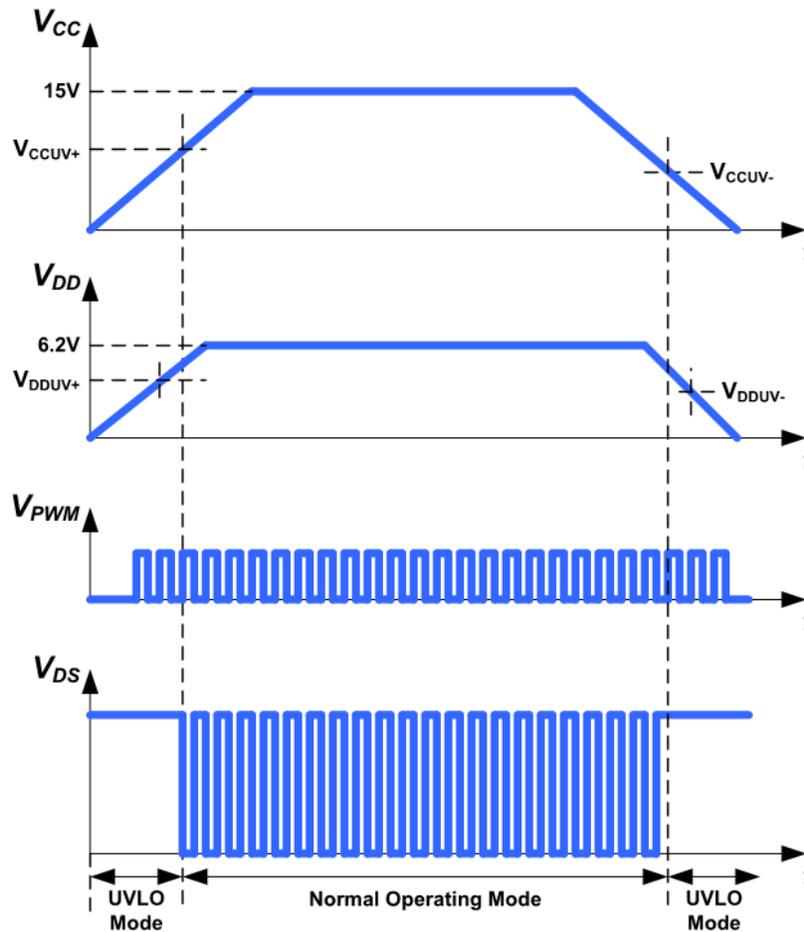


Fig. 22. UVLO Mode timing diagram

### 8.3. Normal Operating Mode

During Normal Operating Mode, all of the internal circuit blocks are active.  $V_{CC}$  is above 9V,  $V_{DD}$  is maintained at 6.2V by the internal voltage regulator, and the internal gate drive and power FET are both enabled. The external PWM signal at the PWM pin determines the frequency and duty-cycle of the internal gate of the power FET. As the PWM voltage toggles above and below the rising and falling input thresholds (2.8V and 1.1V), the internal power FET toggles on and off (Fig. 23). The drain of the power FET then toggles between the source voltage (power ground) and a higher voltage level (700V, max), depending on the external power conversion circuit topology. During each on-time, the CS pin outputs a voltage signal from the internal loss-less current sensing circuit. This circuit measures the current flowing in the GaN power FET without the need for an external current sensing resistor (see section 8.6 GaNSense™ Technology Loss-Less Current Sensing).

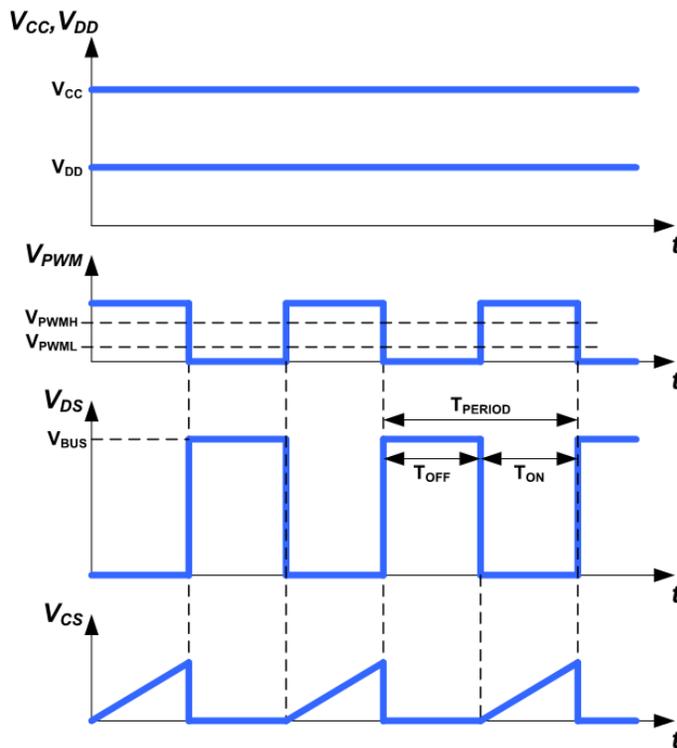


Fig. 23. Normal operating mode timing diagram

### 8.4. Low Power Standby Mode

This GaN Power IC includes an autonomous Low Power Standby Mode for disabling the IC and reducing the  $V_{CC}$  current consumption. During Normal Operating Mode, the PWM pin toggles high and low to turn the GaN power FET on and off. If the input pulses at the PWM pin stop and stay below the lower  $V_{PWML}$  turn-off threshold (1.1V, typical) for the duration of the internal timeout standby delay ( $t_{TO\_STBY}$ , 90usec, typical), then the IC will automatically enter Low Power Standby Mode (Fig. 24). This will disable the gate drive and other internal circuitry and reduce the  $V_{CC}$  supply current to a low level (275uA, typical). When the PWM pulses restart, the IC will wake up instantly at the first rising edge of the PWM input and enter Normal Operating Mode again. To enable auto Standby Mode, the auto-standby mode pin ( $\overline{STBY}$ ) should always be connected to  $S_{GND}$  (set low). To disable auto Standby Mode,  $\overline{STBY}$  pin should be connected to the 5V pin (set high).

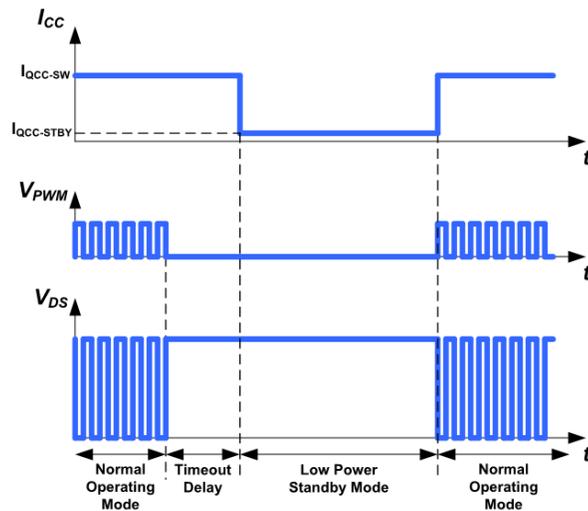


Fig. 24. Autonomous Low Power Standby Mode timing diagram

### 8.5. Programmable Turn-on dV/dt Control

During first start-up pulses or during hard-switching conditions, it is desirable to limit the slew rate ( $dV/dt$ ) of the drain of the power FET during turn-on. This is necessary to reduce EMI or reduce circuit switching noise. To program the turn-on  $dV/dt$  rate of the internal power FET, a resistor ( $R_{DD}$ ) is placed in between the  $V_{DD}$  pin 30 and the  $R_{DD}$  pin 29. This resistor ( $R_{DD}$ ) sets the turn-on current of the internal gate driver and therefore sets the turn-on falling edge  $dV/dt$  rate of the drain of the power FET (Fig. 25).

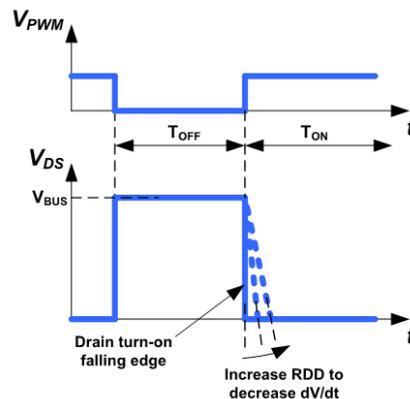


Fig. 25. Turn-on  $dV/dt$  slew rate control

### 8.6. GaNSense™ Technology Loss-Less Current Sensing

For many applications it is necessary to sense the cycle-by-cycle current flowing through the power FET. Existing current sensing solutions include placing a current sensing resistor in between the source of the power FET and P<sub>GND</sub>. This resistor method increases system conduction power losses, creates a hotspot on the PCB, and lowers overall system efficiency. To eliminate this external resistor and hotspot, and increase system efficiency, this IC includes GaNSense™ Technology for integrated and accurate loss-less current sensing. The current flowing through the internal GaN power FET is sensed internally and then converted to a current at the current sensing output pin (CS). An external resistor (R<sub>SET</sub>) is connected from the CS pin to the S<sub>GND</sub> pin and is used to set the amplitude of the CS pin voltage signal (Fig. 26). This allows for the CS pin signal to programmed to work with different controllers with different current sensing input thresholds.

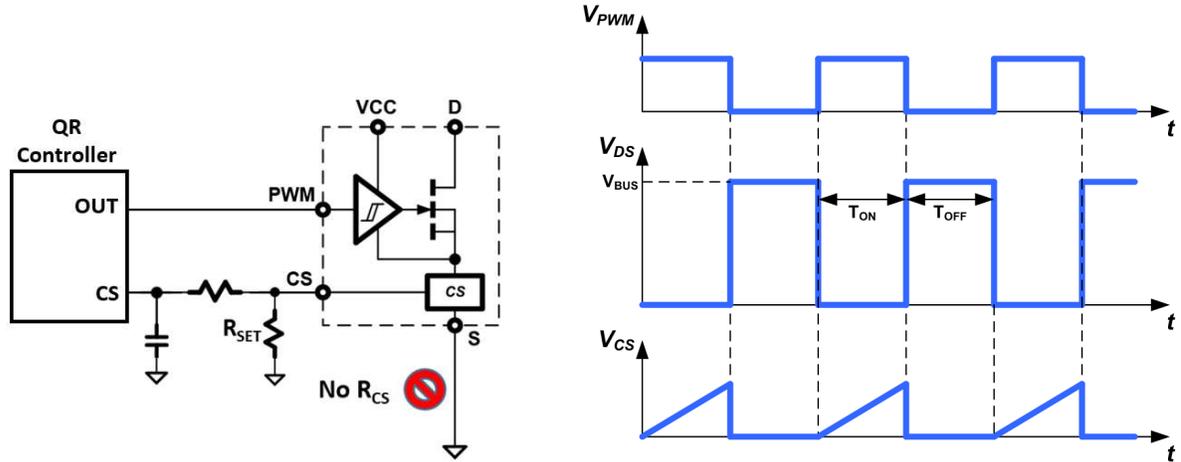
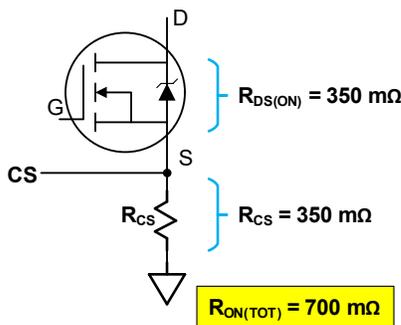


Fig. 26. Current sensing circuit and timing diagram

When comparing GaNSense™ Technology versus existing external resistor sensing method (Fig. 27), the total ON resistance, R<sub>ON(TOT)</sub>, can be substantially reduced. For a 65W high-frequency QR flyback circuit, for example, R<sub>ON(TOT)</sub> is reduced from 700m to 350m. The power loss savings by eliminating the external resistor results in a +0.5% efficiency benefit for the overall system.

#### External Resistor Sensing Method



#### GaNFast™ with GaNSense™

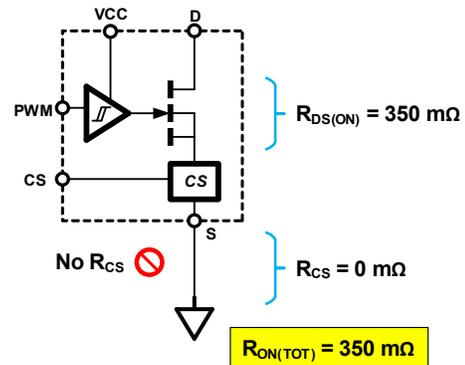


Fig. 27. External resistor sensing vs. GaNSense™ Technology

## GaNSense™ Technology Loss-Less Current Sensing (cont.)

To select the correct  $R_{SET}$  resistor value, the following equation (Equation 1) can be used. This equation uses the equivalent desired external current sensing resistor value ( $R_{CS}$ ), together with the gain of the internal sensing circuitry, to generate the equivalent  $R_{SET}$  resistor value. This  $R_{SET}$  value will then give the correct voltage level at the CS pin to be compatible with the internal current sensing threshold of the system controller.

$$I_{OUT} \text{ Ratio} = \frac{I_{DS}}{I_{CS}} = \frac{2.2A}{0.00125A} = 1760$$

$$R_{SET} = 1760 * R_{CS}$$

$$1760 * 350m\Omega = 616\Omega$$

Equation 1.  $R_{SET}$  resistor value equation

## 8.7. Over Current Protection (OCP)

This GaN Power IC includes cycle-by-cycle over-current detection and protection (OCP) circuitry to protect the GaN power FET against high current levels. During the on-time of each switching cycle, should the peak current exceed the internal OCP threshold (1.9V, typical), then the internal gate drive will turn the GaN power FET off quickly and truncate the on-time period to prevent damage from occurring to the IC. The IC will then turn on again at the next PWM rising edge at the start of the next on-time period (Fig. 28). This OCP protection feature will self-protect the IC each switching cycle against fast peak over current events and greatly increase the robustness and reliability of the system. The actual peak current threshold can be calculated using Equation 2 and is a function of the internal current-sensing ratio and the external  $R_{SET}$  resistor. The internal OCP threshold (1.9V, typical) is much higher than the OCP thresholds of many popular QR, ACF and PFC controllers. This ensures good compatibility of this IC with existing controllers without OCP threshold conflicts.

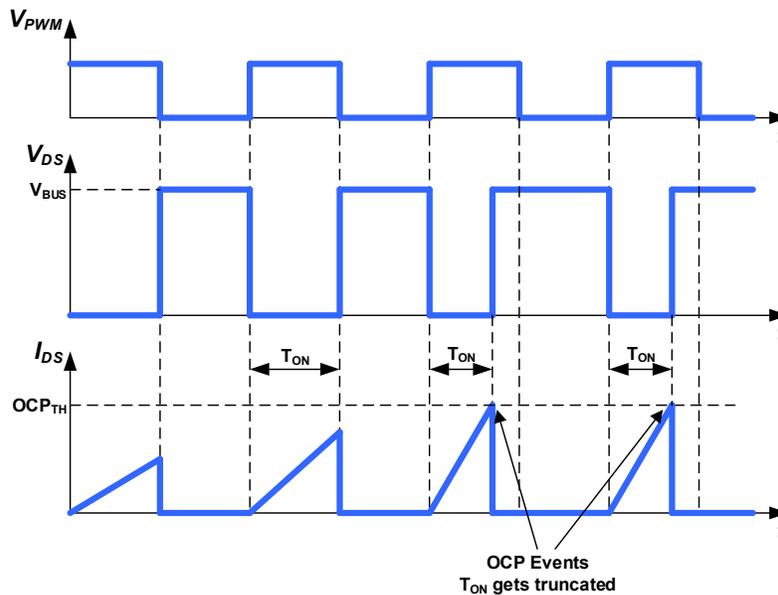


Fig. 28. OCP threshold timing diagram

$$I_{OCP} = \frac{[ 1.9 \text{ V} \times 1760 ]}{R_{SET}}$$

Equation 2. OCP current threshold equation

### 8.8. Over Temperature Protection (OTP)

This GaN Power IC includes over-temperature detection and protection (OTP) circuitry to protect the IC against excessively high junction temperatures ( $T_J$ ). High junction temperatures can occur due to overload, high ambient temperatures, and/or poor thermal management. Should  $T_J$  exceed the internal  $T_{OTP+}$  threshold (165C, typical) then the IC will latch off safely. When  $T_J$  decreases again and falls below the internal  $T_{OTP-}$  threshold (105C, typical), then the OTP latch will be reset. Until then, internal OTP latch guaranteed to remain in the correct state while  $V_{CC}$  is greater than 5V. During an OTP event, this GaN IC will latch off and the system  $V_{CC}$  supply voltage will decrease due to the loss of the aux winding supply. The system  $V_{CC}$  will fall below the lower UV- threshold of the controller and the high-voltage start-up circuit will turn-on and  $V_{CC}$  will increase again (Fig. 29).  $V_{CC}$  will increase above the rising UV+ threshold and the controller turn on again and deliver PWM pulses again.

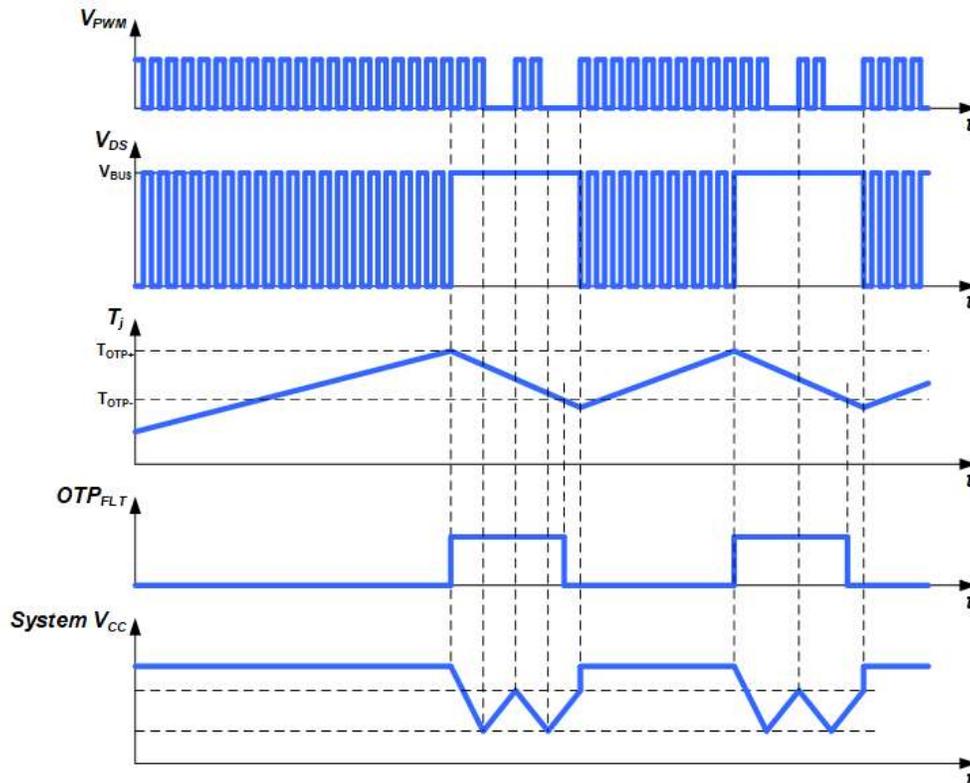


Fig. 29. OTP threshold timing diagram

### 8.9. Drain-to-Source Voltage Considerations

GaN Power ICs have been designed and tested to provide significant design margin to handle transient and continuous voltage conditions that are commonly seen in single-ended topologies, such as quasi-resonant (QR) flyback applications. The different voltage levels and recommended margins in a typical QR flyback can be analyzed using Fig. 30. When the device is switched off, the energy stored in the transformer leakage inductance will cause  $V_{DS}$  to overshoot to the level of  $V_{SPIKE}$ . The clamp circuit should be designed to control the magnitude of  $V_{SPIKE}$ . It is recommended to apply an 80% derating from  $V_{DS(TRAN)}$  rating (800V) to 700 V max for repetitive  $V_{DS}$  spikes under the worst case steady-state operating conditions. After dissipation of the leakage energy, the device  $V_{DS}$  will settle to the level of the bus voltage plus the reflected output voltage which is defined in Fig. 30 as  $V_{PLATEAU}$ . It is recommended to design the system such that  $V_{PLATEAU}$  follows a typical derating of 80% (560V) from  $V_{DS(CONT)}$  (700V). Finally,  $V_{DS(TRAN)}$  (800V) rating is also provided for events that occur on a non-repetitive basis, such as line surge, lightning strikes, start-up, over-current, short-circuit, load transient, and output voltage transition. 800V  $V_{DS(TRAN)}$  ensures excellent device robustness and no-derating is needed for these non-repetitive events, assuming the surge duration is  $< 100 \mu s$ . For half-bridge based topologies, such as LLC,  $V_{DS}$  voltage is clamped to the bus voltage.  $V_{DS}$  should be designed such that it meets the  $V_{PLATEAU}$  derating guideline (560V).

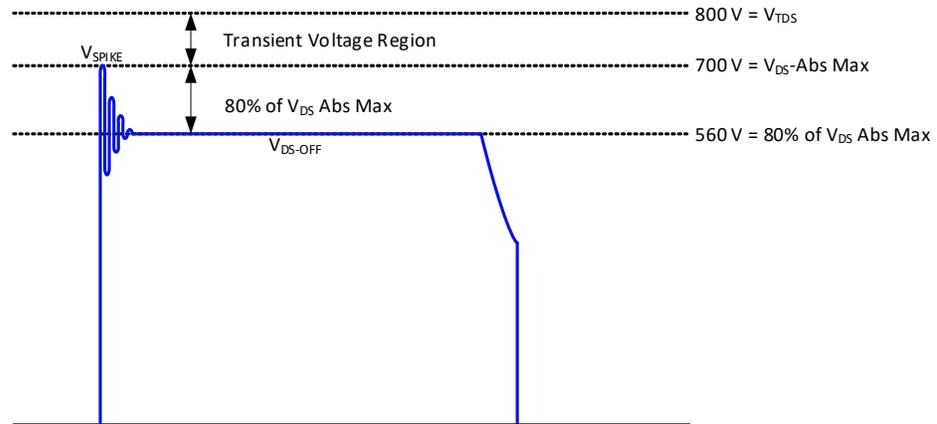
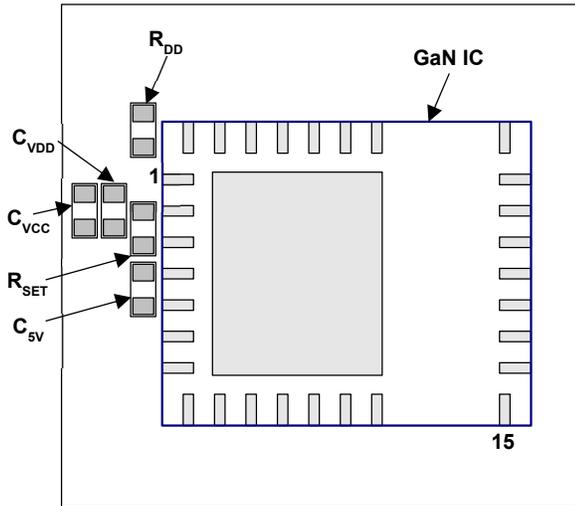


Fig. 30. QR flyback drain-to-source voltage stress diagram

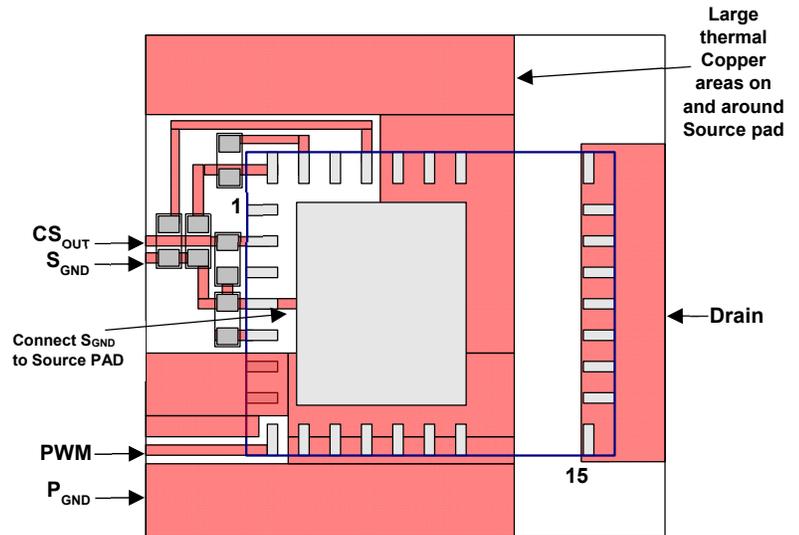
## 9. PCB Layout Guidelines

For best electrical and thermal results, these PCB layout guidelines (and 4 steps below) must be followed:

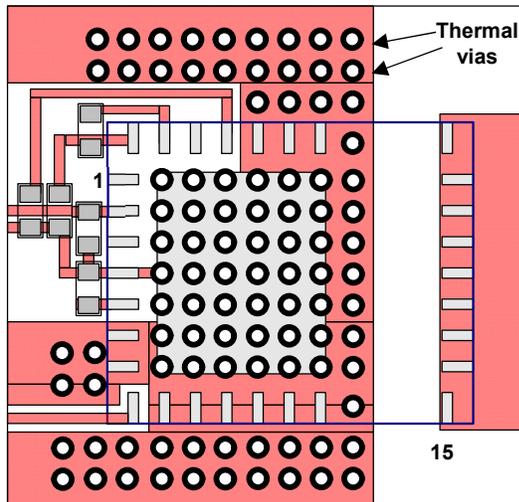
- 1) Place IC components as close as possible to the GaN IC. Place  $R_{SET}$  resistor directly next to CS pin to minimize high frequency switching noise.
- 2) Connect the ground of IC components to  $S_{GND}$  pin 4 to minimize high frequency switching noise. Connect  $S_{GND}$  pin 4 directly to Source PAD underneath IC.
- 3) Route all connections on single layer. This allows for large thermal copper areas on other layers.
- 4) Place large copper areas on and around Source pad.
- 5) Place many thermal vias inside Source pad and inside source copper areas.
- 6) Place large as possible copper areas on all other layers (bottom, top, mid1, mid2).



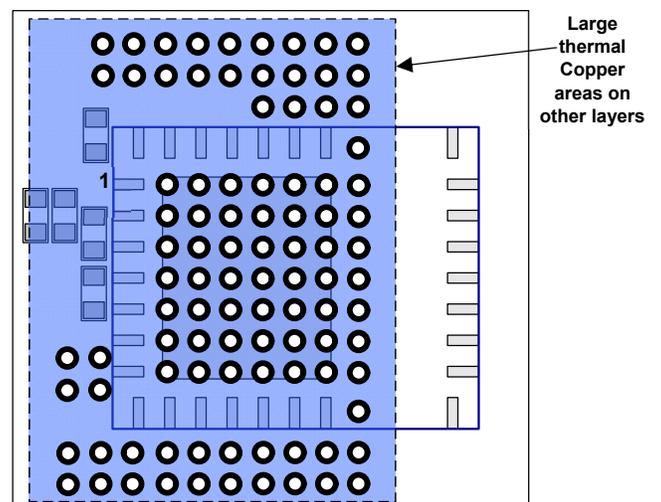
**Step 1.** Place GaN IC and components on PCB. Place components as close as possible to IC!



**Step 2.** Route all connections on single layer. Make large copper areas on and around Source pad!

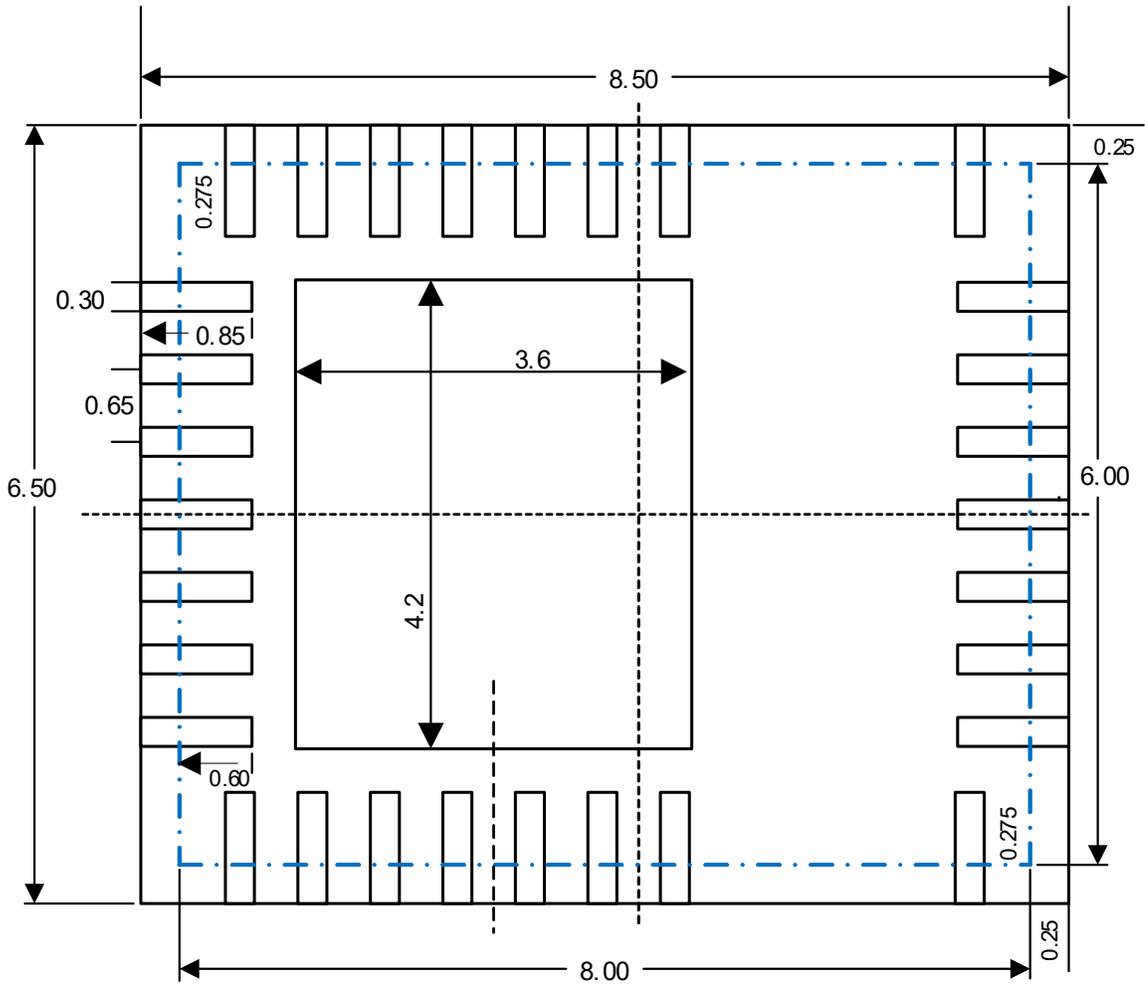


**Step 3.** Place many thermal vias inside source pad and inside source copper areas. (dia=0.65mm, hole=0.33mm, pitch=0.925mm, via wall=1mil)



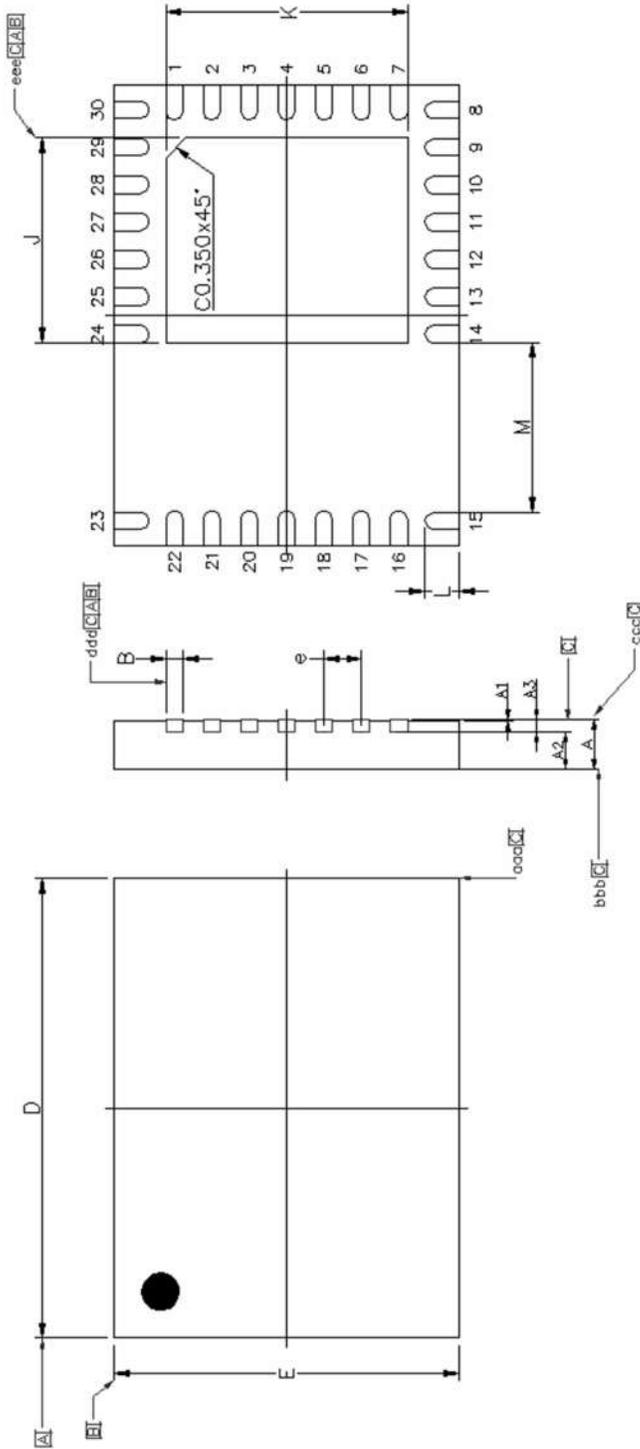
**Step 4.** Place large copper areas on other layers. Make all thermal copper areas as large as possible!

### 10. Recommended PCB Land Pattern



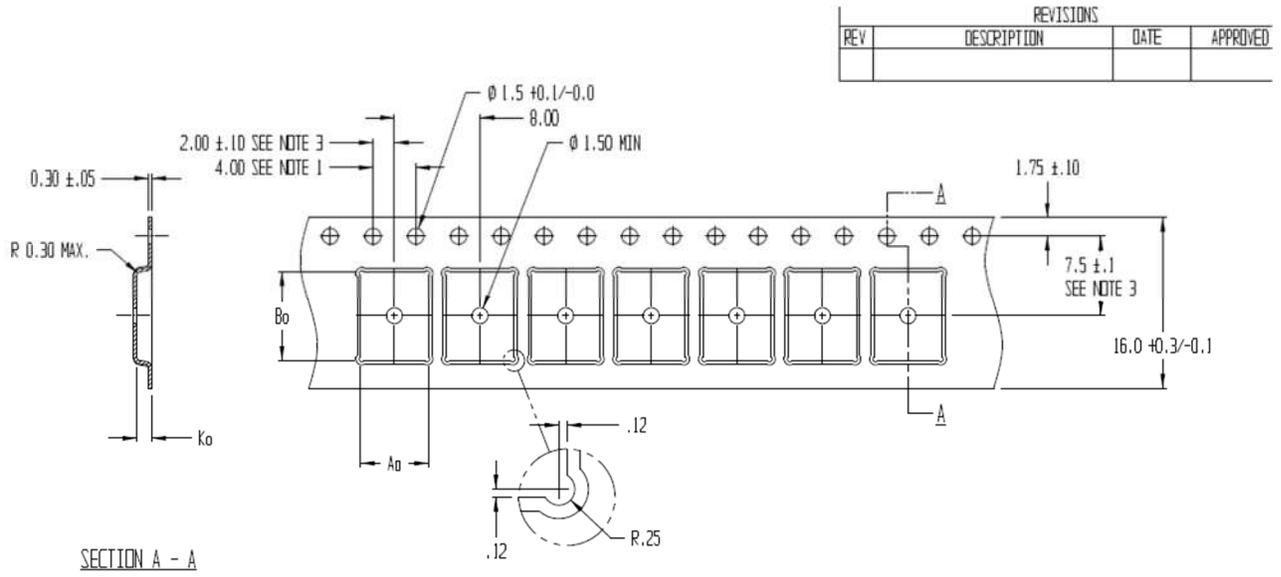
All dimensions are in mm

**11. Package Outline (Power QFN)**



	SYMBOL	MIN	NOM	MAX		SYMBOL	MIN	NOM	MAX
TOTAL THICKNESS	A	0.8	0.85	0.9		X	3.5	3.6	3.7
STANDOFF	A1	0.00	0.02	0.05	EP SIZE	Y	4.1	4.2	4.3
MOLD THICKNESS	A2	---	0.65	---	LEAD LENGTH	L	0.55	0.6	0.65
L/F THICKNESS	A3	---	0.203 REF	---	HIGH VOLTAGE SPACING	M	2.85	2.95	3.05
LEAD WIDTH	B	0.25	0.3	0.35	PACKAGE EDGE TOLERANCE	aaa		0.1	
BODY SIZE	D	8.00 BSC			MOLD FLATNESS	bbb		0.1	
		6.00 BSC			COPLANARITY	ccc		0.08	
LEAD PITCH	e	0.65 BSC			LEAD OFFSET	ddd		0.1	
					EXPOSED PAD OFFSET	eee		0.1	

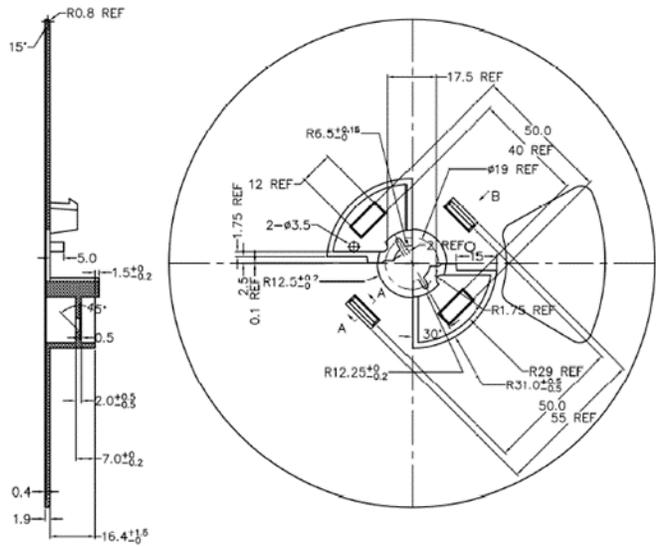
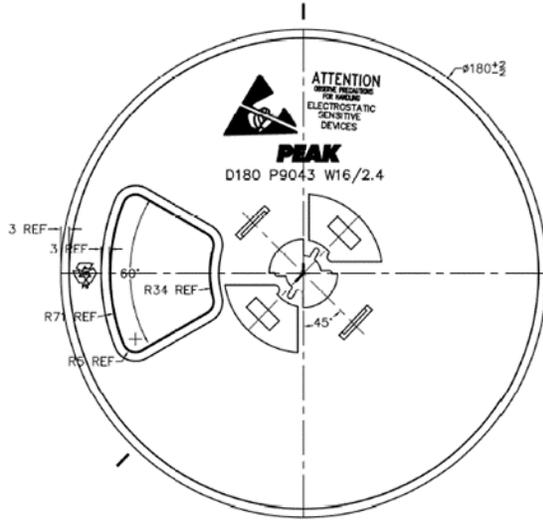
**12. Tape and Reel Dimensions**



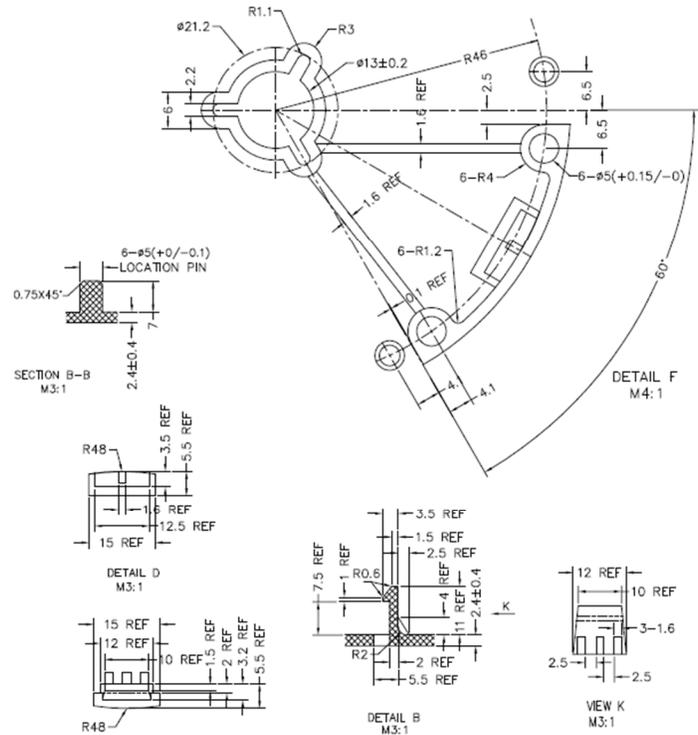
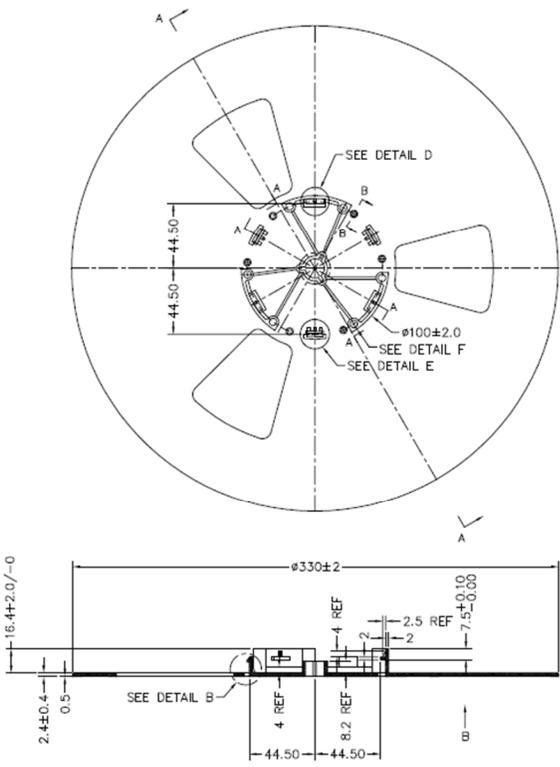
$A_0 = 6.35$   
 $B_0 = 8.35$   
 $K_0 = 1.40$

**12. Tape and Reel Dimensions (Cont.)**

7" Reel



13" Reel



### 13. Ordering Information

Part Number	Operating Temperature Grade	Storage Temperature Range	Package	MSL Rating	Packing (Tape & Reel)
NV6133A-RA	-55°C to +150°C T <sub>CASE</sub>	-55°C to +150°C T <sub>CASE</sub>	6 x 8 mm PQFN	3	1,000: 7" Reel
NV6133A	-55°C to +150°C T <sub>CASE</sub>	-55°C to +150°C T <sub>CASE</sub>	6 x 8 mm PQFN	3	5,000: 13" Reel

### 14. 20-Year Limited Product Warranty

The 20-year limited warranty applies to all packaged Navitas GaNFast Power ICs in mass production, subject to the terms and conditions of, Navitas' express limited product warranty, available at <https://navitassemi.com/terms-conditions>. The warranted specifications include only the MIN and MAX values only listed in Absolute Maximum Ratings, ESD Ratings and Electrical Characteristics sections of this datasheet. Typical (TYP) values or other specifications are not warranted.



### 15. Revision History

Date	Status	Notes
Dec. 17, 2021	FINAL	First publication
Apr. 13, 2022	FINAL	Increased V <sub>DD6</sub> Max to 6.6V; increased V <sub>CCUV+</sub> Max to 9.3V; added 20-Year Limited Product Warranty
Oct. 11, 2022	FINAL	Added Carbon Neutral certificate

### Additional Information

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