

BIPOLAR ANALOG INTEGRATED CIRCUIT

μ PC8187TB

SILICON MMIC HI-IP₃ FREQUENCY UP-CONVERTER FOR WIRELESS TRANSCEIVER

DESCRIPTION

The μ PC8187TB is a silicon monolithic integrated circuit designed as frequency up-converter for wireless transceiver. This IC is higher operating frequency, lower distortion and higher conversion gain than conventional μ PC8163TB.

This IC is manufactured using NEC's 30 GHz f_{max} UHS0 (Ultra High Speed Process) silicon bipolar process.

FEATURES

- High output frequency : $f_{RFout} = 0.8$ to 2.5 GHz
- High-density surface mounting : 6-pin super minimold package
- Supply voltage : $V_{cc} = 2.7$ to 3.3 V
- Higher IP₃ : OIP₃ = +10 dBm @ $f_{RFout} = 1.9$ GHz

APPLICATION

- TDMA, PCS, CDMA etc.

ORDERING INFORMATION

| Part Number | Package | Marking | Supplying Form |
|---------------------|----------------------|---------|---|
| μ PC8187TB-E3-A | 6-pin super minimold | C3G | <ul style="list-style-type: none"> • Embossed tape 8 mm wide. • Pin 1, 2, 3 face the tape perforation side. • Qty 3 kpcs/reel. |

Remark To order evaluation samples, please contact your local sales office.

(Part number for sample order: μ PC8187TB-A)

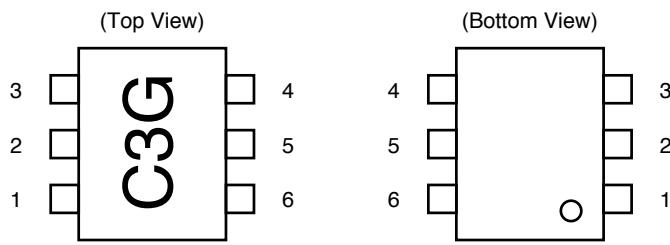
Caution Electro-static sensitive devices

The information in this document is subject to change without notice. Before using this document, please confirm that this is the latest version.

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1. PIN CONNECTIONS



| Pin No. | Pin Name |
|---------|-----------------|
| 1 | IFinput |
| 2 | GND |
| 3 | LOinput |
| 4 | GND |
| 5 | V _{cc} |
| 6 | RFoutput |

2. SERIES PRODUCTS ($T_A = +25^\circ\text{C}$, $V_{CC} = V_{PS} = V_{RFout} = 3.0 \text{ V}$, $Z_S = Z_L = 50 \Omega$)

| Part Number | I _{CC} (mA) | f _{RFout} (GHz) | CG (dB) | | |
|----------------|----------------------|--------------------------|-----------------------------|-------------|-------------|
| | | | @RF 0.9 GHz ^{Note} | @RF 1.9 GHz | @RF 2.4 GHz |
| μ PC8187TB | 15 | 0.8 to 2.5 | 11 | 11 | 10 |
| μ PC8106TB | 9 | 0.4 to 2.0 | 9 | 7 | - |
| μ PC8172TB | 9 | 0.8 to 2.5 | 9.5 | 8.5 | 8.0 |
| μ PC8109TB | 5 | 0.4 to 2.0 | 6 | 4 | - |
| μ PC8163TB | 16.5 | 0.8 to 2.0 | 9 | 5.5 | - |

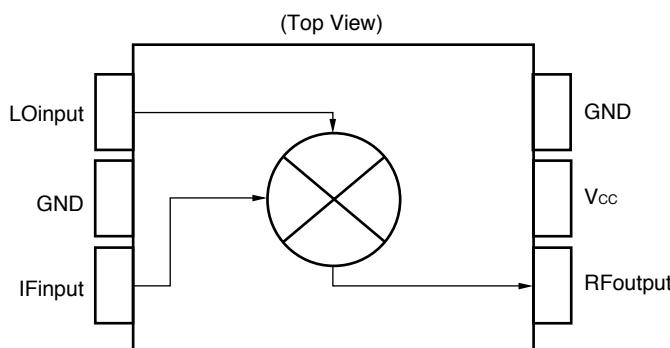
| Part Number | P _{O(sat)} (dBm) | | | OIP ₃ (dBm) | | |
|----------------|-----------------------------|-------------|-------------|-----------------------------|-------------|-------------|
| | @RF 0.9 GHz ^{Note} | @RF 1.9 GHz | @RF 2.4 GHz | @RF 0.9 GHz ^{Note} | @RF 1.9 GHz | @RF 2.4 GHz |
| μ PC8187TB | +4 | +2.5 | +1 | +10 | +10 | +8.5 |
| μ PC8106TB | -2 | -4 | - | +5.5 | +2.0 | - |
| μ PC8172TB | +0.5 | 0 | -0.5 | +7.5 | +6.0 | +4.0 |
| μ PC8109TB | -5.5 | -7.5 | - | +1.5 | -1.0 | - |
| μ PC8163TB | +0.5 | -2 | - | +9.5 | +6.0 | - |

Note f_{RFout} = 0.83 GHz @ μ PC8163TB and μ PC8187TB

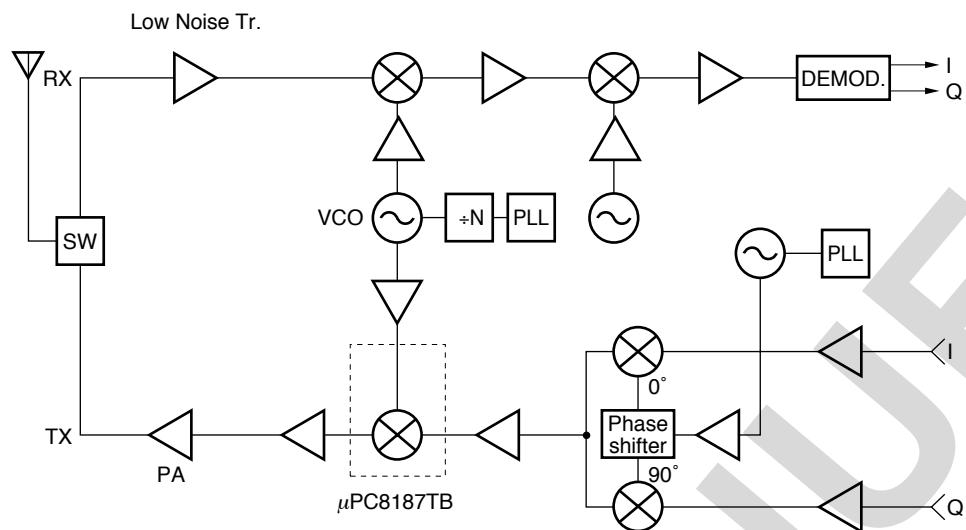
Remark Typical performance. Please refer to 8. ELECTRICAL CHARACTERISTICS in detail.

To know the associated product, please refer to each latest data sheet.

3. BLOCK DIAGRAM



4. SYSTEM APPLICATION EXAMPLES (SCHEMATICS OF IC LOCATION IN THE SYSTEM)



5. PIN EXPLANATION

| Pin No. | Pin Name | Applied Voltage (V) | Pin Voltage (V) ^{Note} | Function and Explanation | Equivalent Circuit |
|---------|----------|--|---------------------------------|--|--------------------|
| 1 | IFinput | - | 1.2 | This pin is IF input to double balanced mixer (DBM). The input is designed as high impedance. The circuit contributes to suppress spurious signal. Also this symmetrical circuit can keep specified performance insensitive to process-condition distribution. For above reason, double balanced mixer is adopted. | |
| 2 4 | GND | GND | - | GND pin. Ground pattern on the board should be formed as wide as possible. Track Length should be kept as short as possible to minimize ground impedance. | |
| 3 | LOinput | - | 2.1 | Local input pin. Recommendable input level is -10 to 0 dBm. | |
| 5 | Vcc | 2.7 to 3.3 | - | Supply voltage pin. | |
| 6 | RFoutput | Same bias as Vcc through external inductor | - | This pin is RF output from DBM. This pin is designed as open collector. Due to the high impedance output, this pin should be externally equipped with LC matching circuit to next stage. | |

Note Each pin voltage is measured at $V_{CC} = V_{RFout} = 2.8$ V.

6. ABSOLUTE MAXIMUM RATINGS

| Parameter | Symbol | Test Conditions | Rating | Unit |
|-------------------------------|------------------|--|-------------|------|
| Supply Voltage | V _{CC} | T _A = +25°C | 3.6 | V |
| Power Dissipation | P _D | Mounted on double-side copperclad 50 × 50 × 1.6 mm epoxy glass PWB, T _A = +85°C | 270 | mW |
| Operating Ambient Temperature | T _A | | -40 to +85 | °C |
| Storage Temperature | T _{STG} | | -55 to +150 | °C |
| Maximum Input Power | P _{IN} | | +10 | dBM |

7. RECOMMENDED OPERATING RANGE

| Parameter | Symbol | MIN. | TYP. | MAX. | Unit | Remarks |
|-------------------------------|--------------------|------|------|------|------|---|
| Supply Voltage | V _{CC} | 2.7 | 2.8 | 3.3 | V | The same voltage should be applied to pin 5 and 6 |
| Operating Ambient Temperature | T _A | -40 | +25 | +85 | °C | |
| Local Input Power | P _{LOIN} | -10 | -5 | 0 | dBM | Z _S = 50 Ω (without matching) |
| RF Output Frequency | f _{RFout} | 0.8 | - | 2.5 | GHz | With external matching circuit |
| IF Input Frequency | f _{IFin} | 50 | - | 400 | MHz | |

8. ELECTRICAL CHARACTERISTICS

(T_A = +25°C, V_{CC} = V_{RFout} = 2.8 V, f_{IFin} = 150 MHz, P_{LOin} = -5 dBm)

| Parameter | Symbol | Test Conditions ^{Note} | MIN. | TYP. | MAX. | Unit |
|------------------------|----------------------|--|------|------|------|------|
| Circuit Current | I _{CC} | No signal | 11 | 15 | 19 | mA |
| Conversion Gain | CG1 | f _{RFout} = 0.83 GHz, P _{IFin} = -20 dBm | 8 | 11 | 14 | dB |
| | CG2 | f _{RFout} = 1.9 GHz, P _{IFin} = -20 dBm | 8 | 11 | 14 | dB |
| | CG3 | f _{RFout} = 2.4 GHz, P _{IFin} = -20 dBm | 7 | 10 | 13 | dB |
| Saturated Output Power | P _{O(sat)1} | f _{RFout} = 0.83 GHz, P _{IFin} = 0 dBm | +1.5 | +4 | - | dBM |
| | P _{O(sat)2} | f _{RFout} = 1.9 GHz, P _{IFin} = 0 dBm | 0 | +2.5 | - | dBM |
| | P _{O(sat)3} | f _{RFout} = 2.4 GHz, P _{IFin} = 0 dBm | -1.5 | +1 | - | dBM |

Note f_{RFout} < f_{LOin} @ f_{RFout} = 0.83 GHz

f_{LOin} < f_{RFout} @ f_{RFout} = 1.9 GHz/2.4 GHz

9. OTHER CHARACTERISTICS, FOR REFERENCE PURPOSES ONLY

($T_A = +25^\circ\text{C}$, $V_{CC} = V_{RFout} = 2.8 \text{ V}$, $P_{LOin} = -5 \text{ dBm}$)

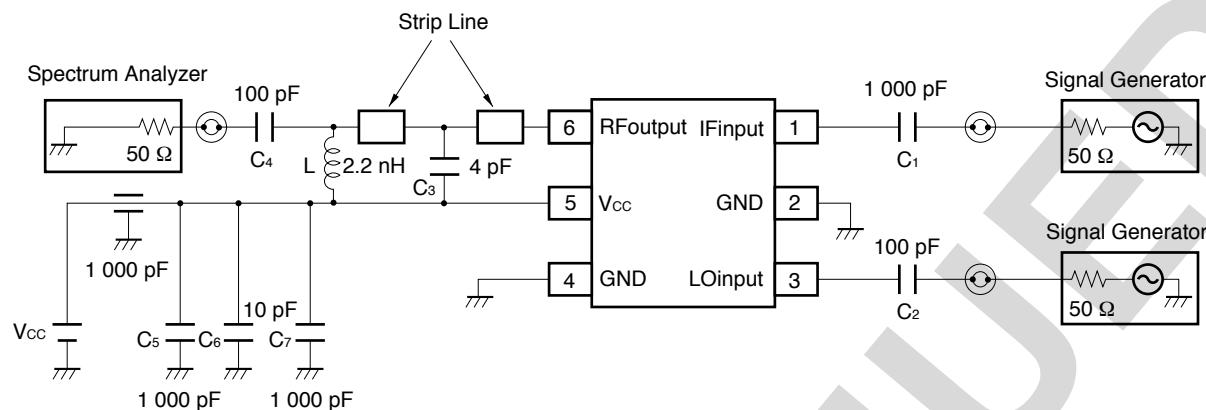
| Parameter | Symbol | Test Conditions ^{Note} | | Value | Unit |
|---|-------------------|---------------------------------|--|-------|------|
| Output 3rd Order Distortion Intercept Point | OIP ₃₁ | $f_{RFout} = 0.83 \text{ GHz}$ | $f_{IFin1} = 150 \text{ MHz}$ $f_{IFin2} = 151 \text{ MHz}$ | +10 | dBm |
| | OIP ₃₂ | $f_{RFout} = 1.9 \text{ GHz}$ | | +10 | dBm |
| | OIP ₃₃ | $f_{RFout} = 2.4 \text{ GHz}$ | | +8.5 | dBm |
| Input 3rd Order Distortion Intercept Point | IIP ₃₁ | $f_{RFout} = 0.83 \text{ GHz}$ | $f_{IFin1} = 150 \text{ MHz}$ $f_{IFin2} = 151 \text{ MHz}$ | -1.0 | dBm |
| | IIP ₃₂ | $f_{RFout} = 1.9 \text{ GHz}$ | | -1.0 | dBm |
| | IIP ₃₃ | $f_{RFout} = 2.4 \text{ GHz}$ | | -1.5 | dBm |
| SSB Noise Figure | SSB•NF1 | $f_{RFout} = 0.83 \text{ GHz}$ | $f_{IFin} = 150 \text{ MHz}$ | 11 | dB |
| | SSB•NF2 | $f_{RFout} = 1.9 \text{ GHz}$ | | 12 | dB |
| | SSB•NF3 | $f_{RFout} = 2.4 \text{ GHz}$ | | 12.5 | dB |

Note $f_{RFout} < f_{LOin}$ @ $f_{RFout} = 0.83 \text{ GHz}$

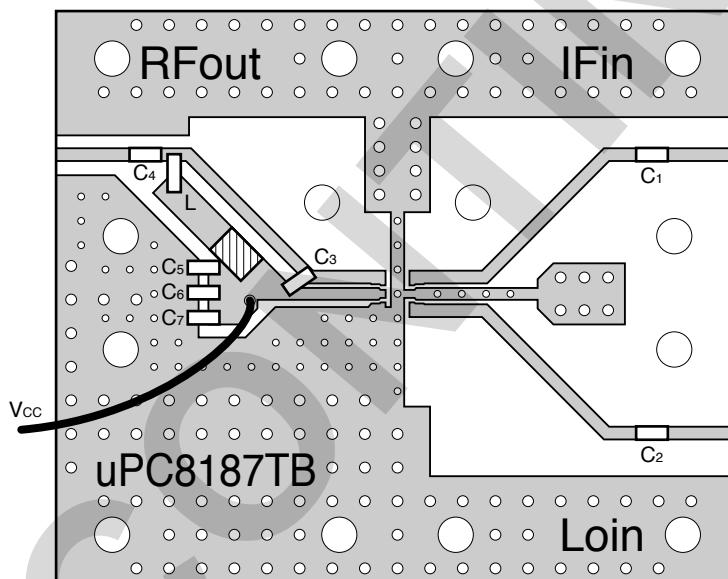
$f_{LOin} < f_{RFout}$ @ $f_{RFout} = 1.9 \text{ GHz}/2.4 \text{ GHz}$

★ 10. TEST CIRCUITS

10.1 TEST CIRCUIT 1 ($f_{RFout} = 0.83$ GHz)



EXAMPLE OF TEST CIRCUIT 1 ASSEMBLED ON EVALUATION BOARD

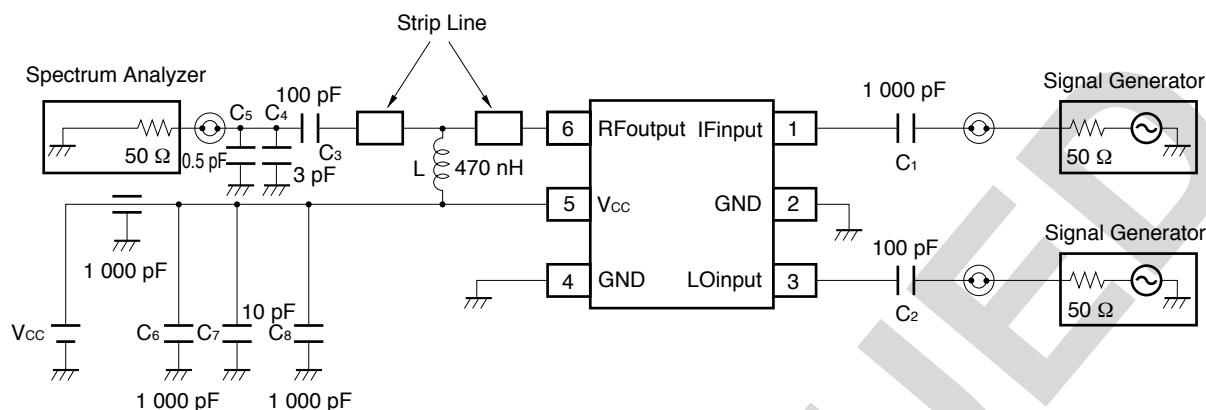


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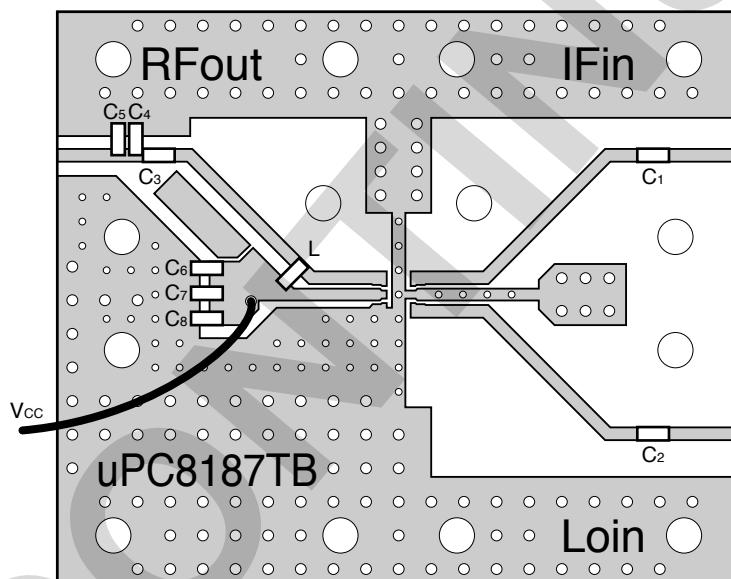
| Form | Symbol | Value |
|----------------|--|------------------------|
| Chip capacitor | C ₁ , C ₅ , C ₇ | 1 000 pF |
| | C ₂ , C ₄ | 100 pF |
| | C ₆ | 10 pF |
| | C ₃ | 4 pF |
| Chip inductor | L | 2.2 nH ^{Note} |

- (*1) 35 × 42 × 0.4 mm polyimide board, double-sided copper clad
- (*2) Ground pattern on rear of the board
- (*3) Solder plated patterns
- (*4) ○○○ : Through holes
- (*5) ━━━ : Join patterns with electrical tape

Note 2.2 nH: LL1608-FH2N25 (TOKO Co., Ltd.)

10.2 TEST CIRCUIT 2 ($f_{RFout} = 1.9$ GHz)

EXAMPLE OF TEST CIRCUIT 2 ASSEMBLED ON EVALUATION BOARD

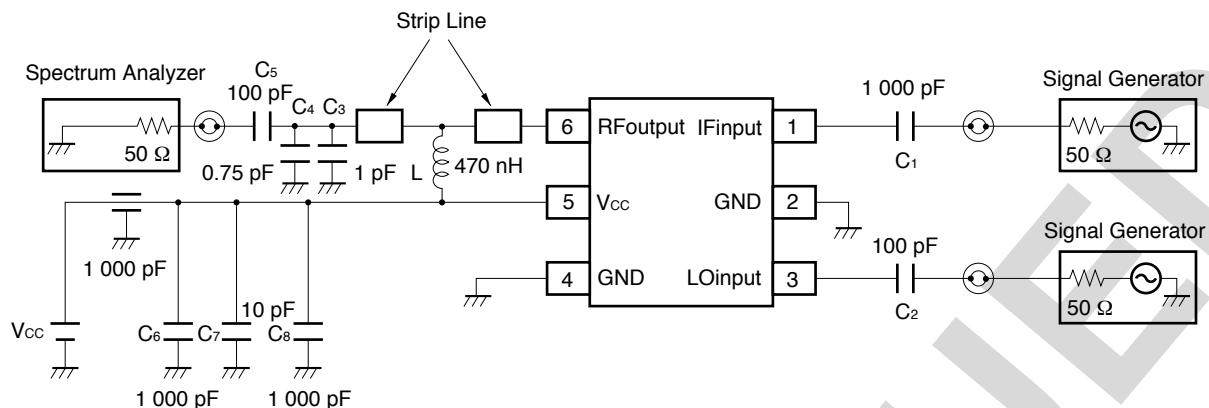


COMPONENT LIST

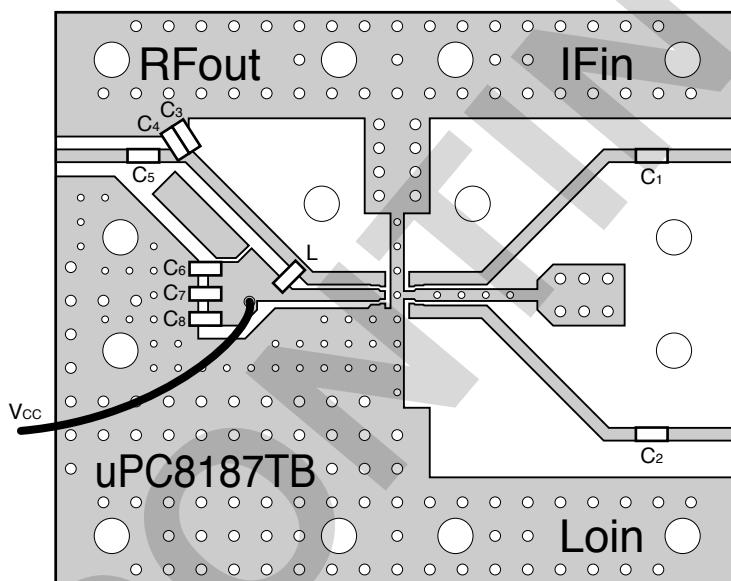
| Form | Symbol | Value |
|----------------|--|------------------------|
| Chip capacitor | C ₁ , C ₆ , C ₈ | 1 000 pF |
| | C ₂ , C ₃ | 100 pF |
| | C ₇ | 10 pF |
| | C ₄ | 3 pF |
| | C ₅ | 0.5 pF |
| Chip inductor | L | 470 nH ^{Note} |

- (*1) 35 × 42 × 0.4 mm polyimide board, double-sided copper clad
- (*2) Ground pattern on rear of the board
- (*3) Solder plated patterns
- (*4) ○○○: Through holes

Note 470 nH: LL2012-FR47 (TOKO Co., Ltd.)

10.3 TEST CIRCUIT 3 ($f_{RFout} = 2.4$ GHz)

EXAMPLE OF TEST CIRCUIT 3 ASSEMBLED ON EVALUATION BOARD



COMPONENT LIST

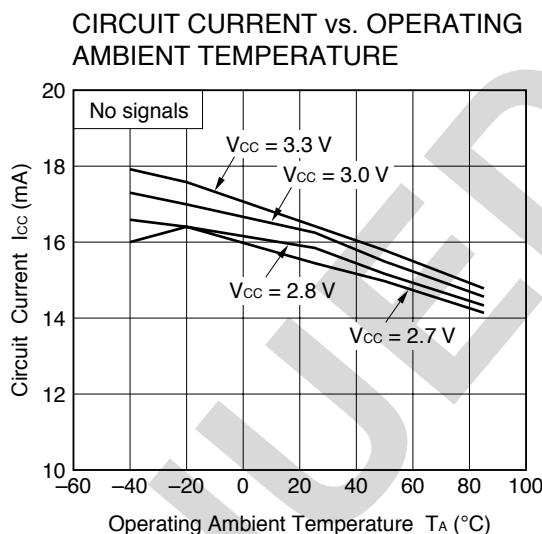
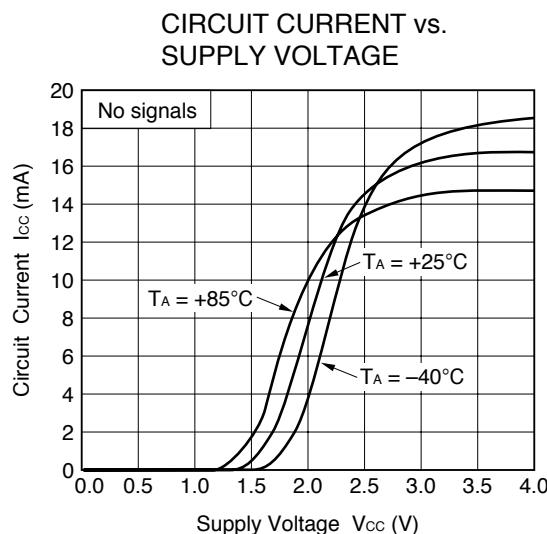
| Form | Symbol | Value |
|----------------|--|------------------------|
| Chip capacitor | C ₁ , C ₆ , C ₈ | 1000 pF |
| | C ₂ , C ₅ | 100 pF |
| | C ₇ | 10 pF |
| | C ₃ | 1 pF |
| | C ₄ | 0.75 pF |
| Chip inductor | L | 470 nH ^{Note} |

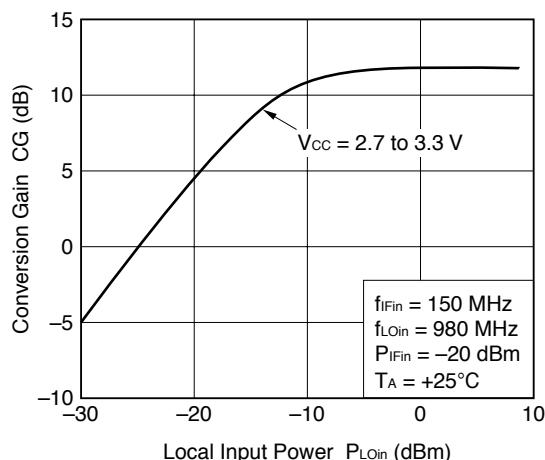
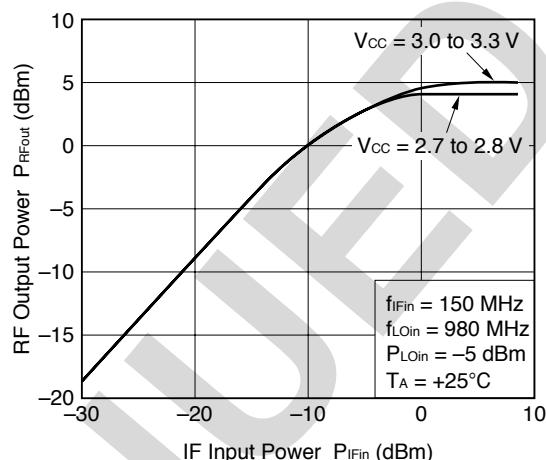
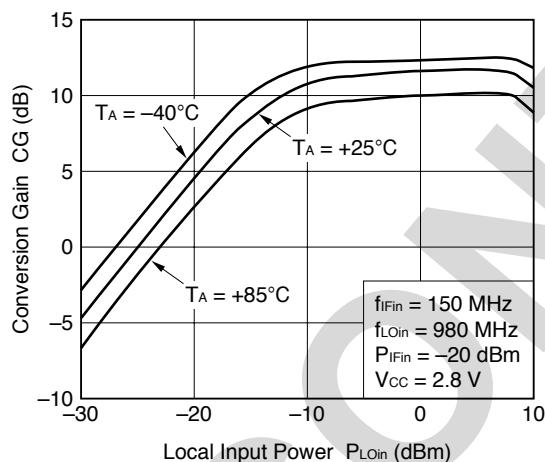
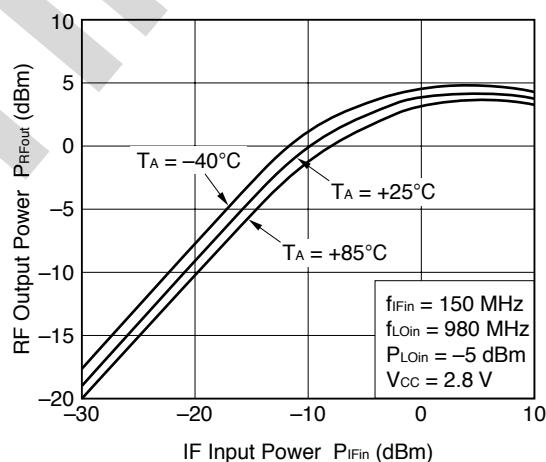
- (*1) 35 × 42 × 0.4 mm polyimide board, double-sided copper clad
- (*2) Ground pattern on rear of the board
- (*3) Solder plated patterns
- (*4) ○○○ : Through holes

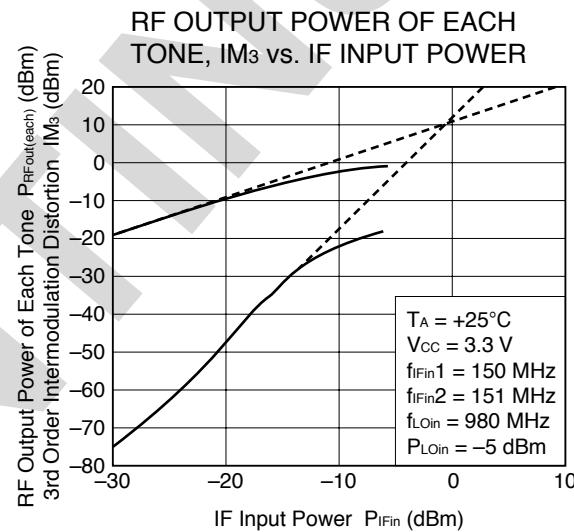
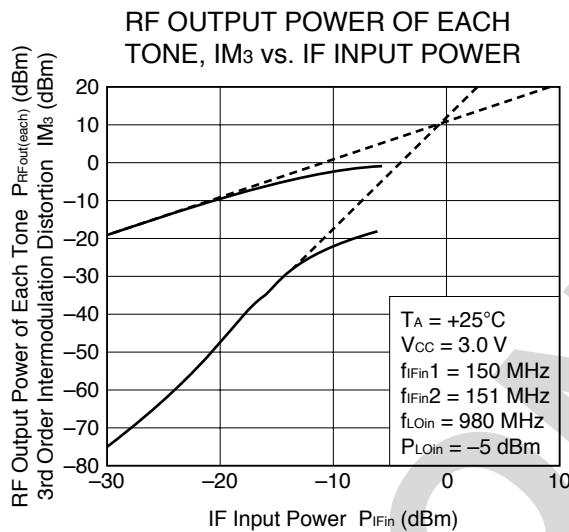
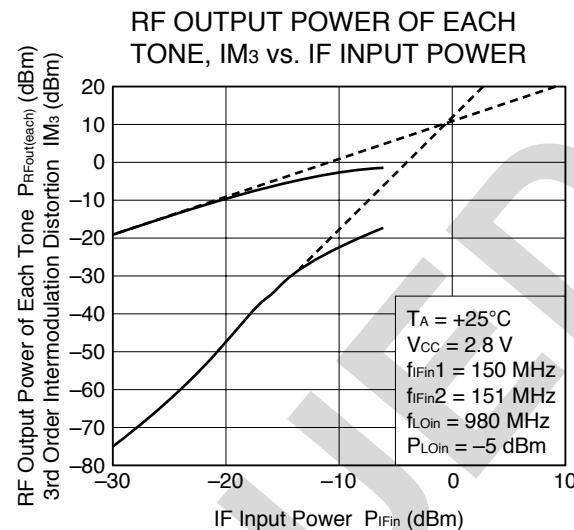
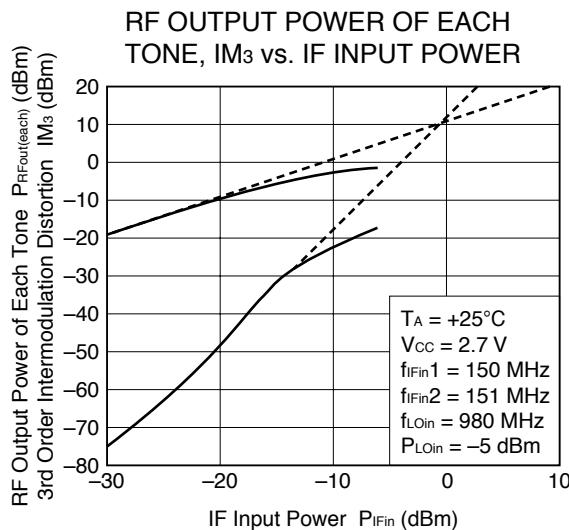
Note 470 nH: LL2012-FR47 (TOKO Co., Ltd.)

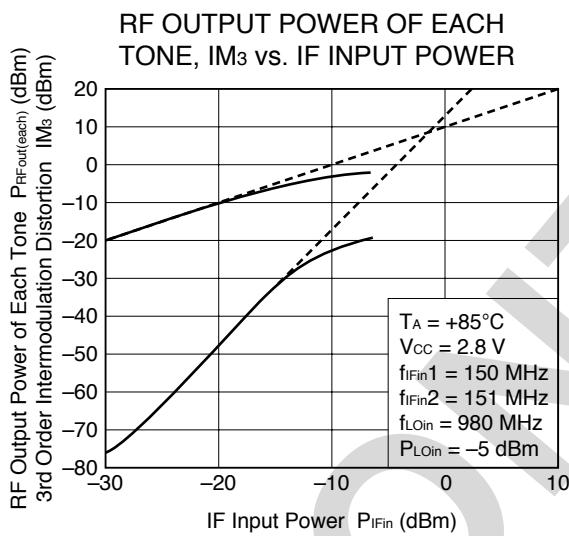
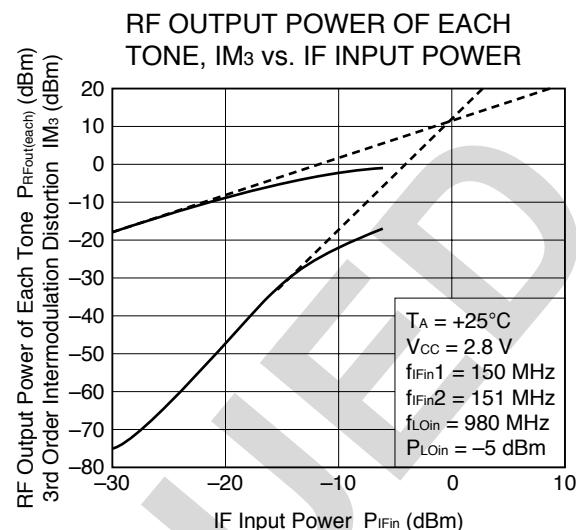
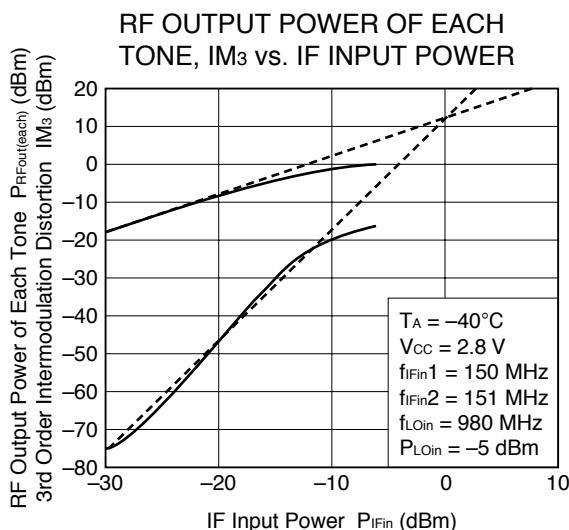
Caution The test circuits and board pattern on data sheet are for performance evaluation use only (They are not recommended circuits). In the case of actual design-in, matching circuit should be determined using S-parameter of desired frequency in accordance to actual mounting pattern.

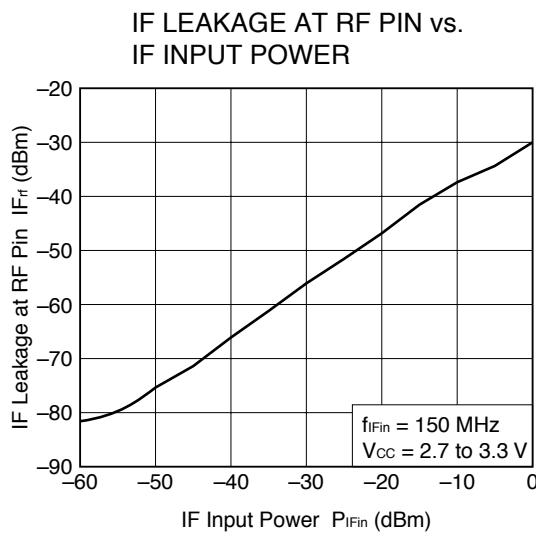
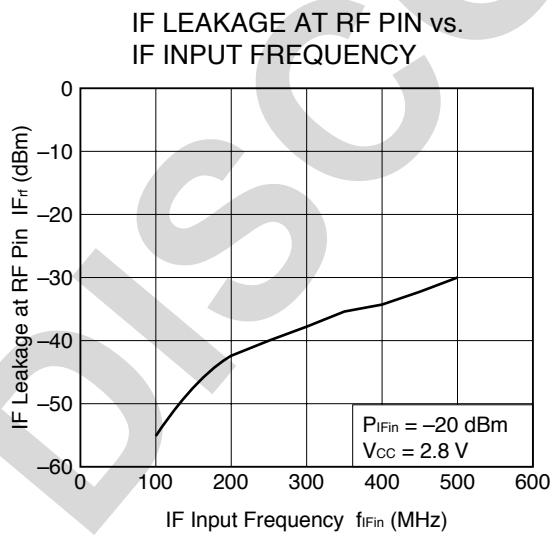
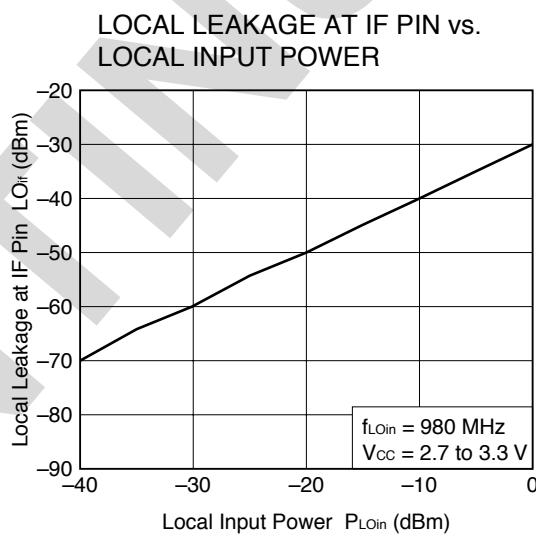
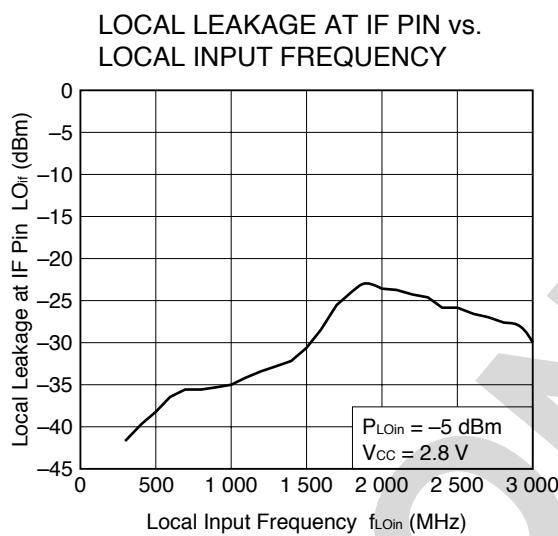
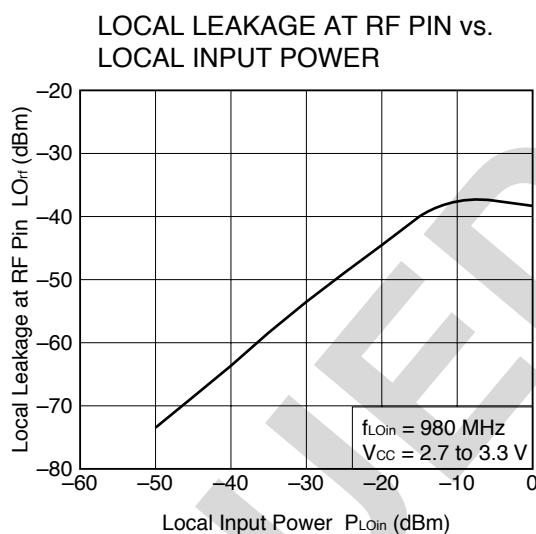
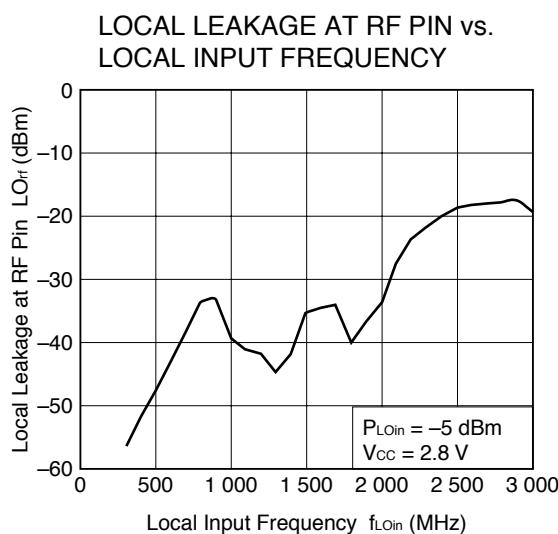
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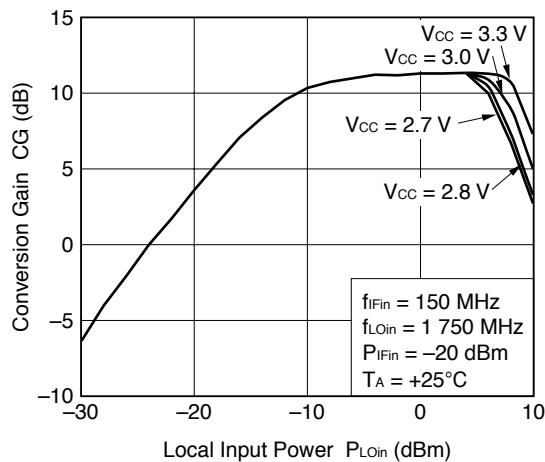
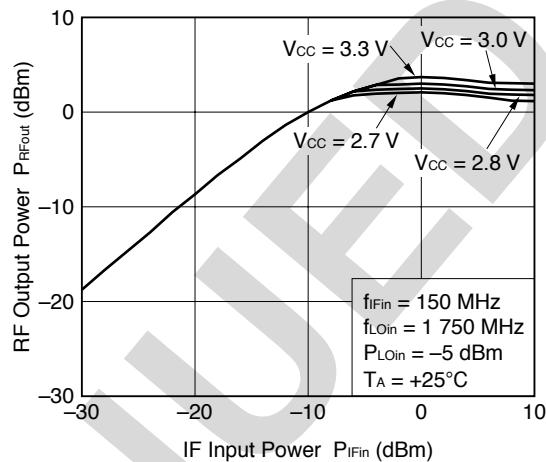
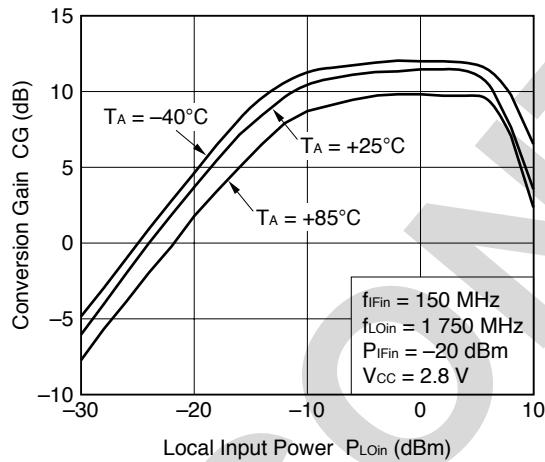
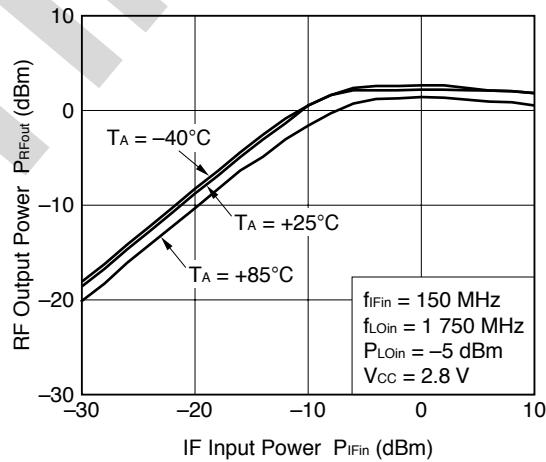
★ 11. TYPICAL CHARACTERISTICS (Unless otherwise specified, $T_A = +25^\circ\text{C}$, $V_{CC} = V_{RFout}$)

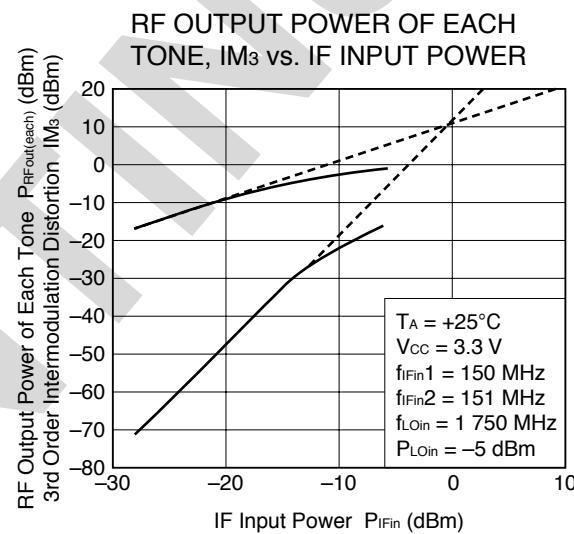
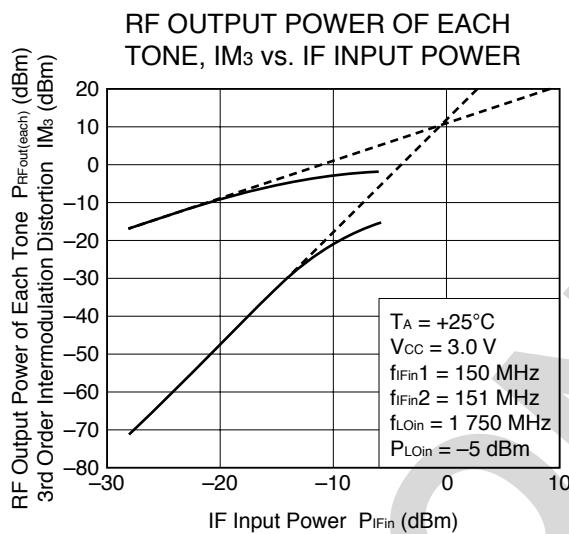
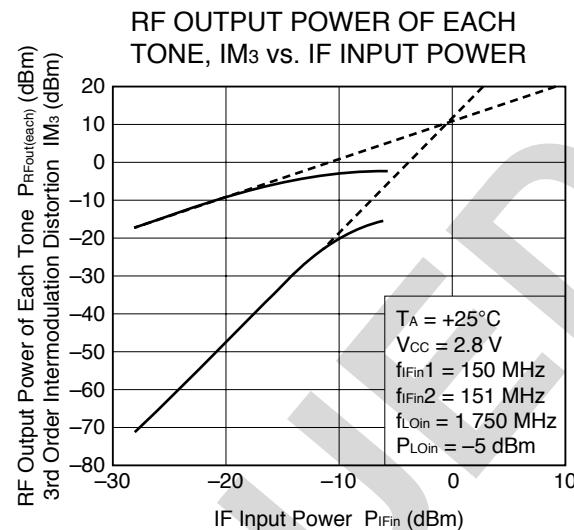
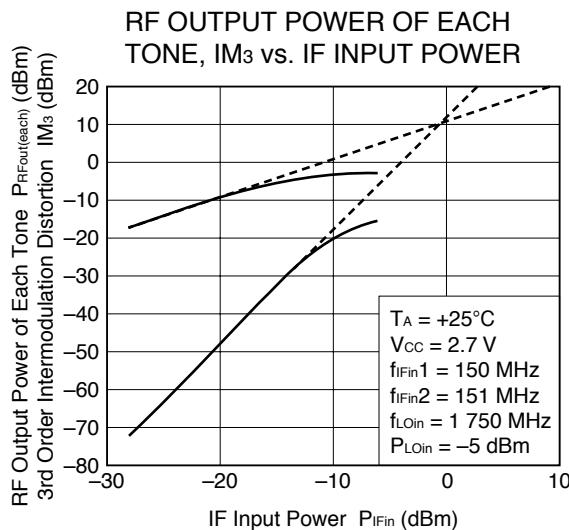
11.1 $f_{RFout} = 0.83$ GHzCONVERSION GAIN vs.
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IF INPUT POWERCONVERSION GAIN vs.
LOCAL INPUT POWERRF OUTPUT POWER vs.
IF INPUT POWER

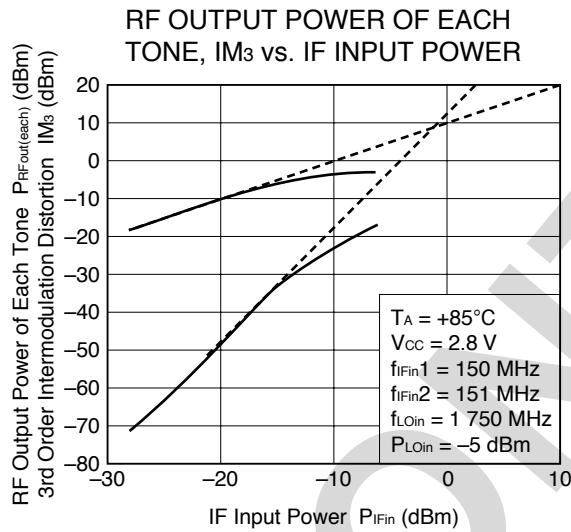
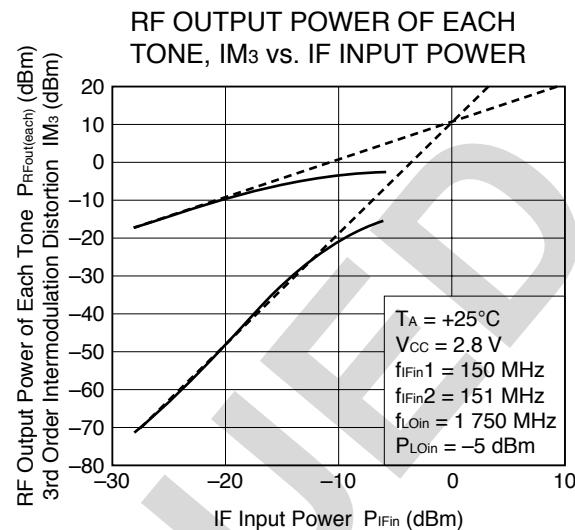
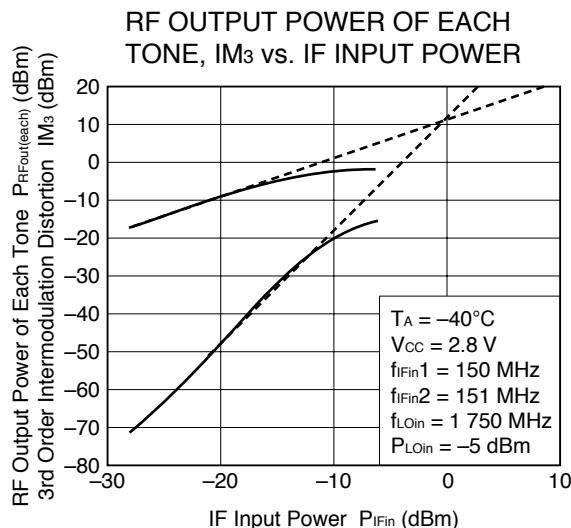


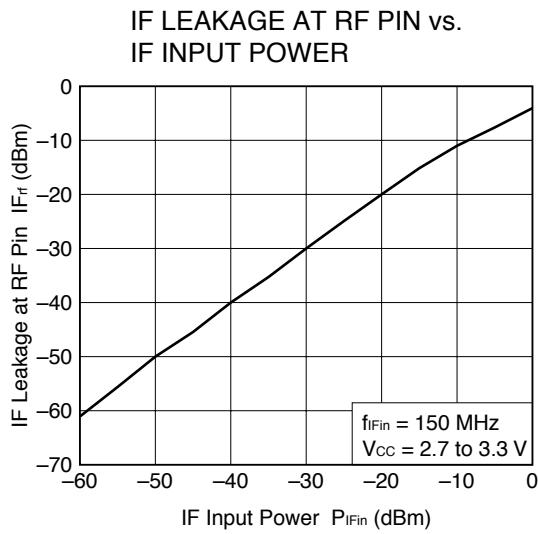
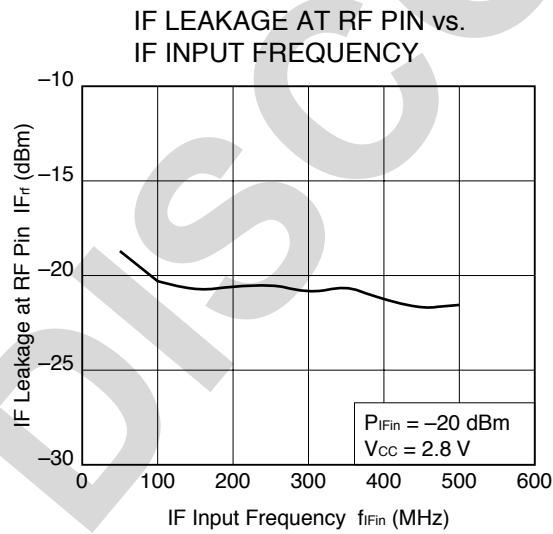
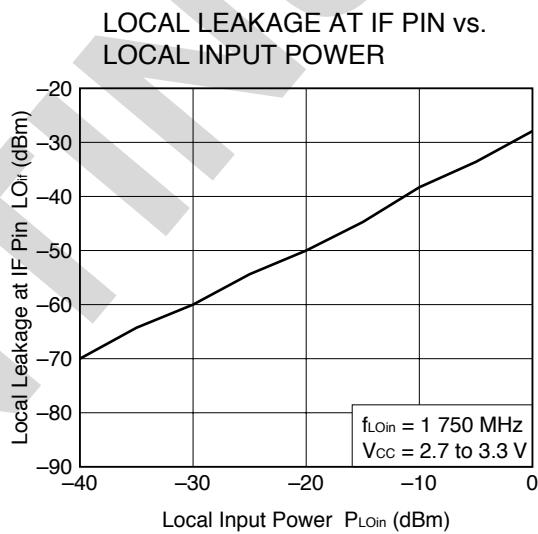
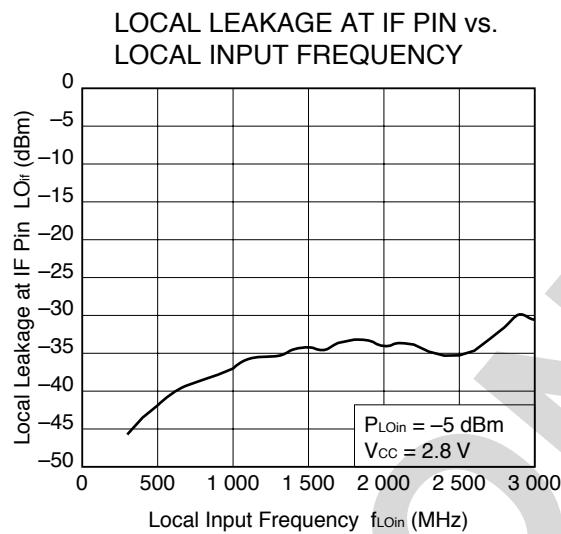
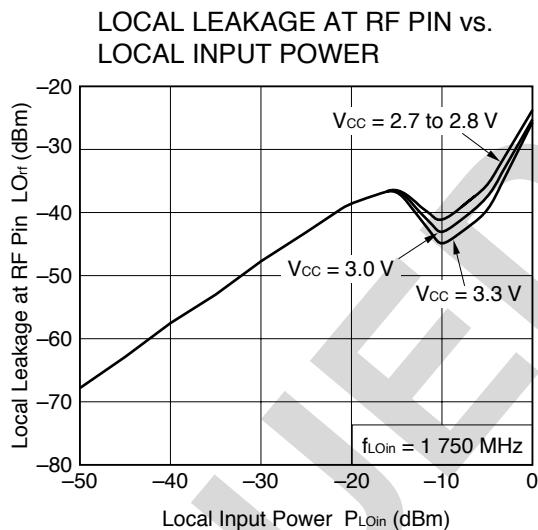
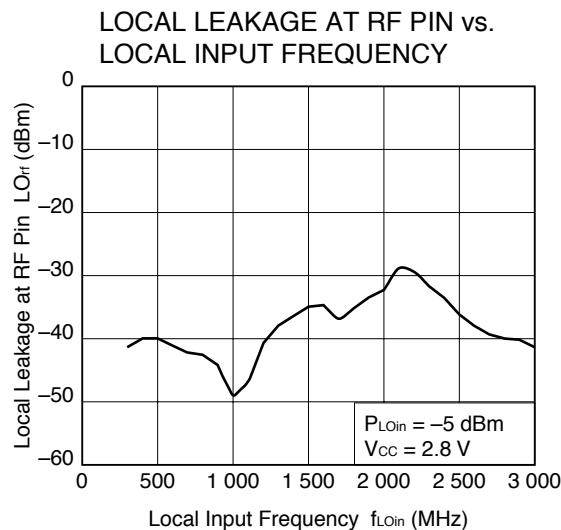


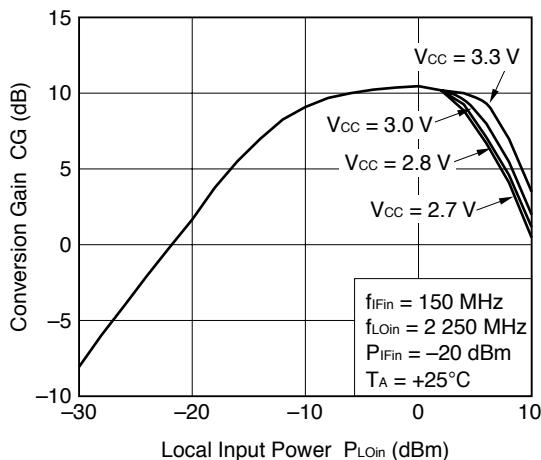
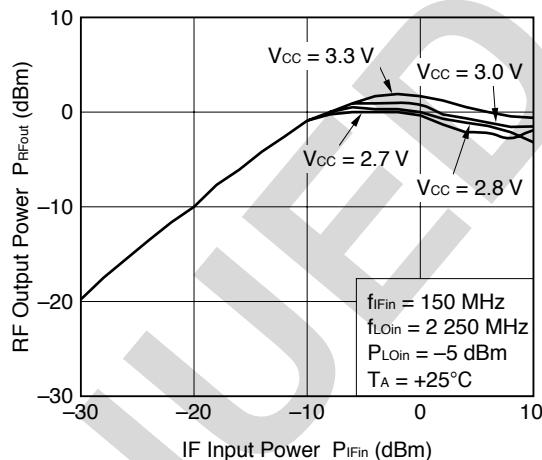
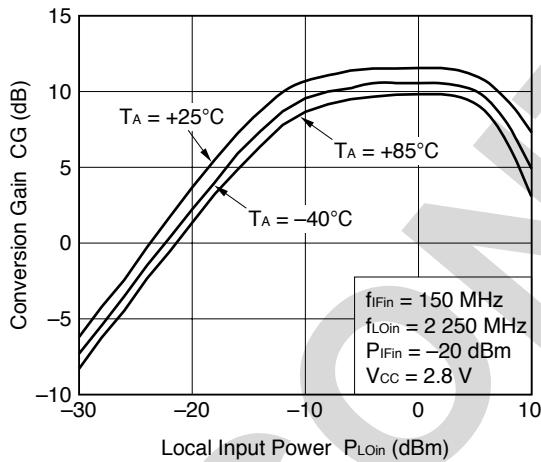
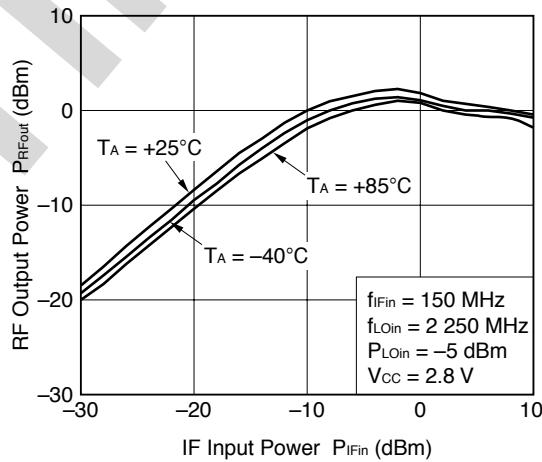


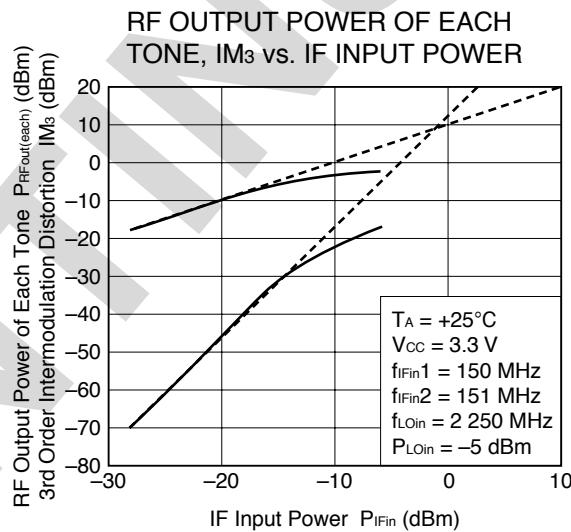
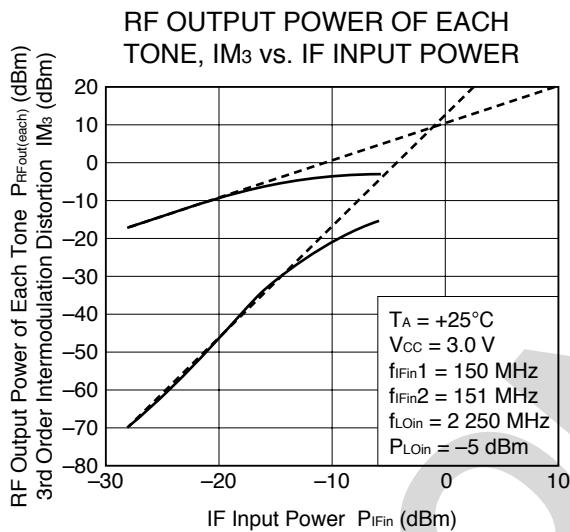
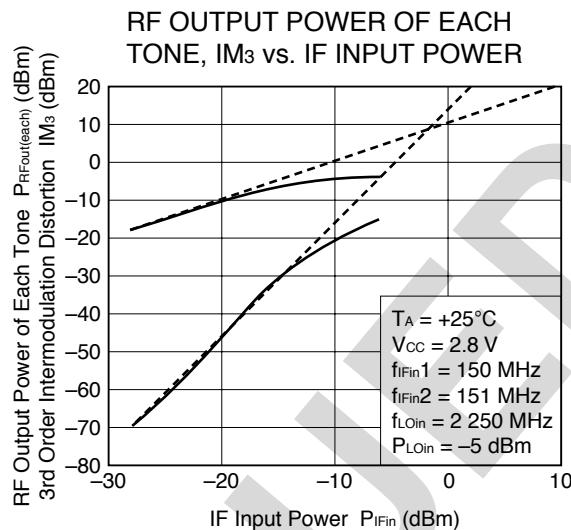
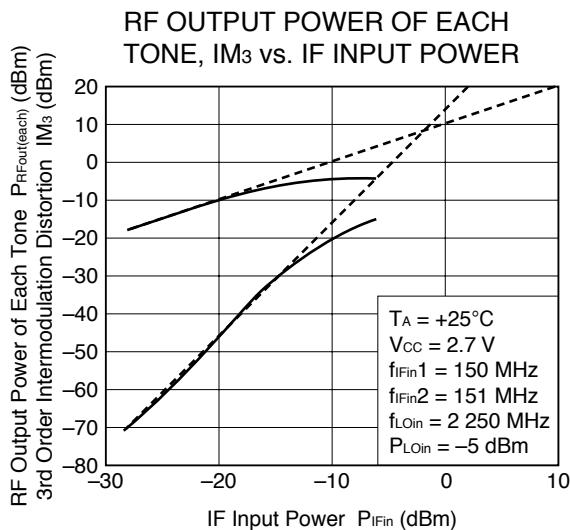
11.2 $f_{RFout} = 1.9$ GHzCONVERSION GAIN vs.
LOCAL INPUT POWERRF OUTPUT POWER vs.
IF INPUT POWERCONVERSION GAIN vs.
LOCAL INPUT POWERRF OUTPUT POWER vs.
IF INPUT POWER

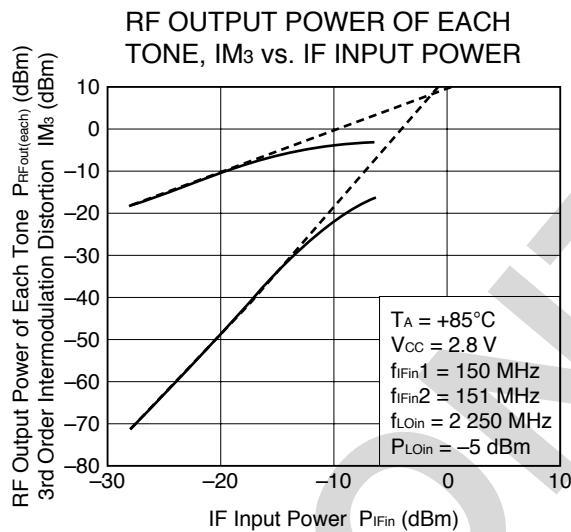
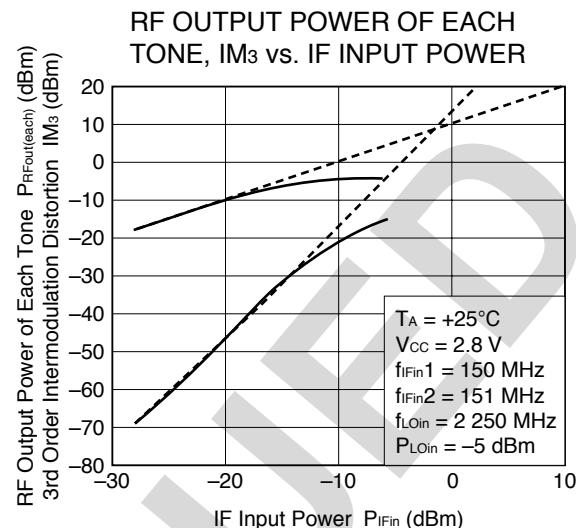
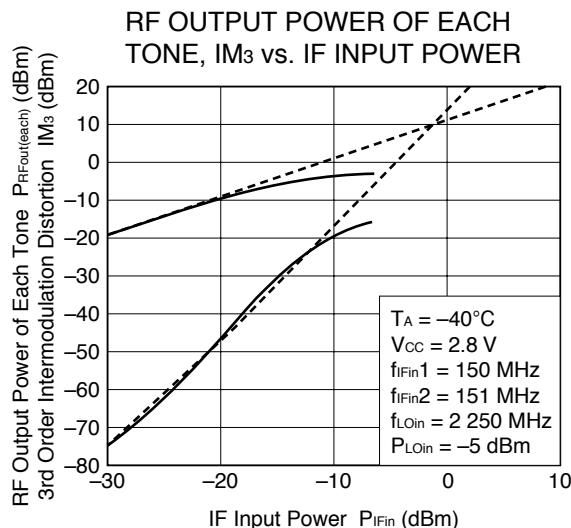


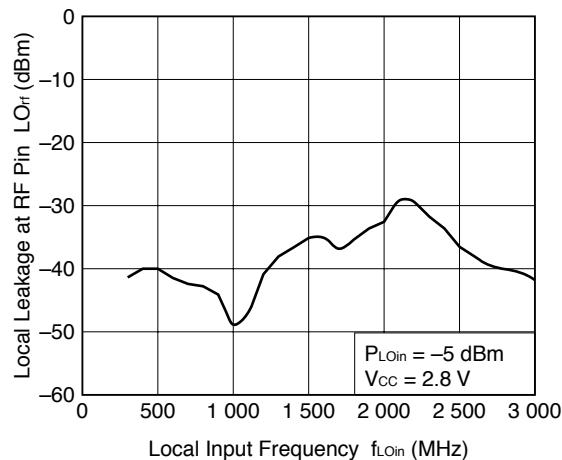
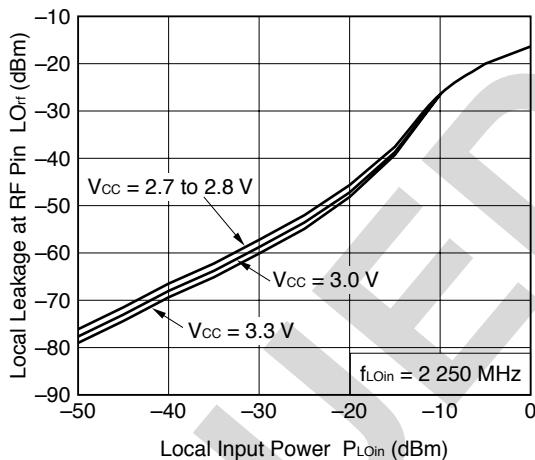
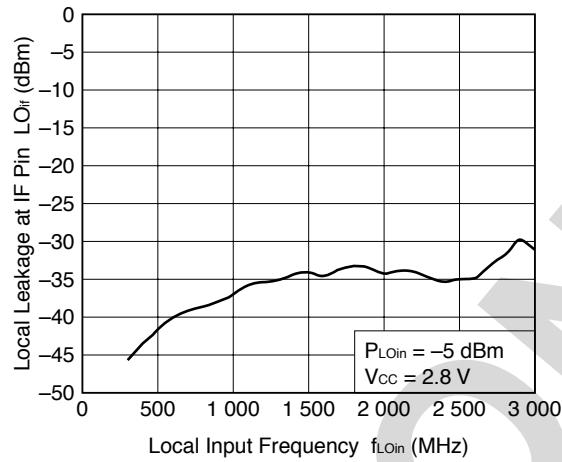
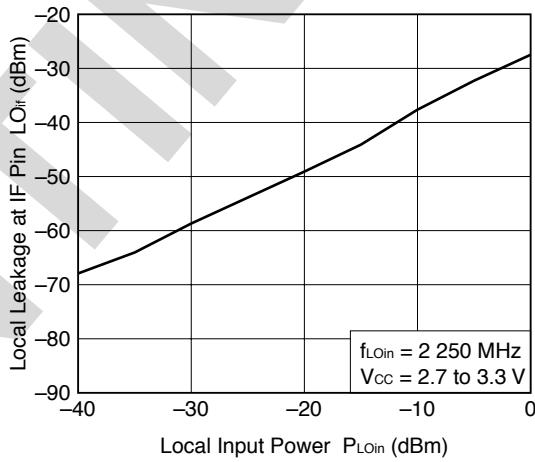
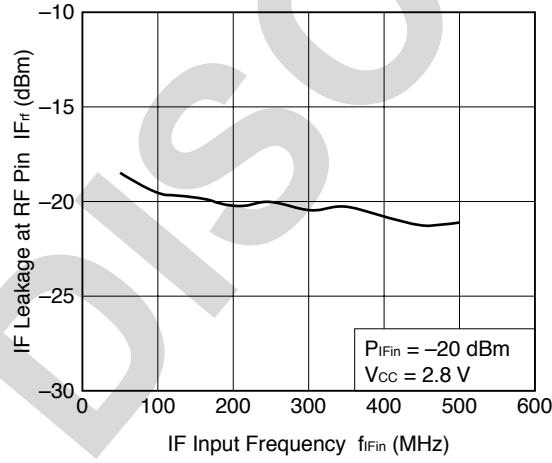
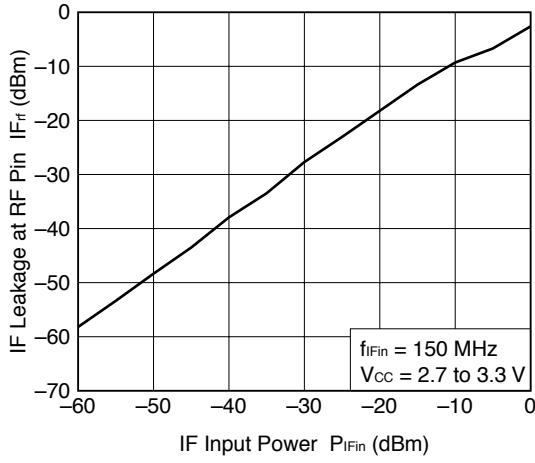




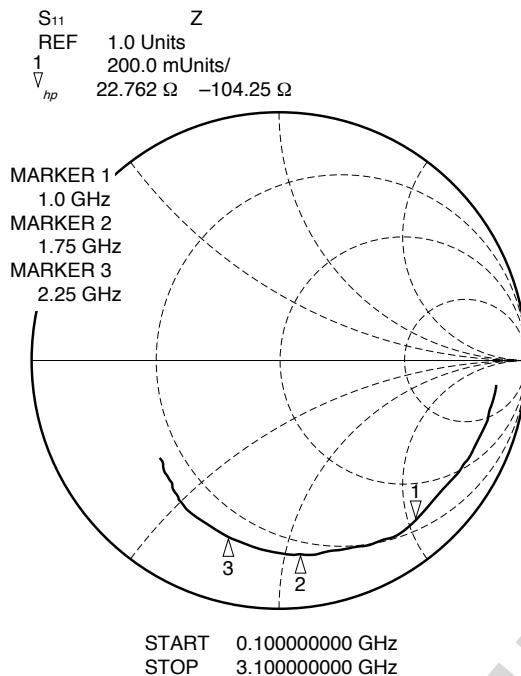
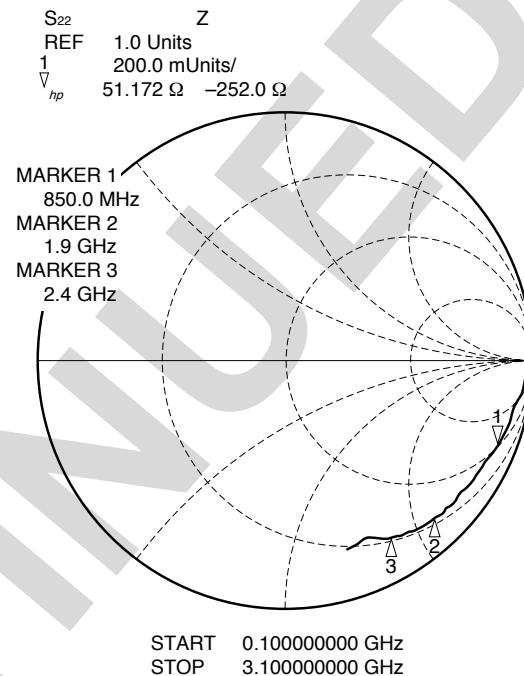
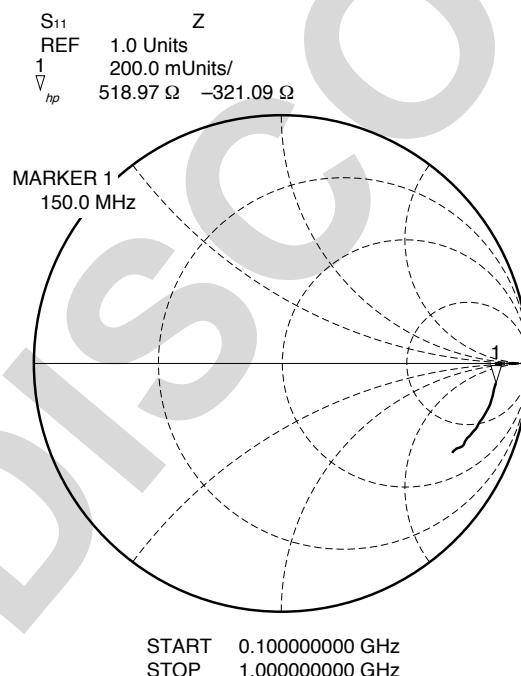
11.3 $f_{RFout} = 2.4$ GHzCONVERSION GAIN vs.
LOCAL INPUT POWERRF OUTPUT POWER vs.
IF INPUT POWERCONVERSION GAIN vs.
LOCAL INPUT POWERRF OUTPUT POWER vs.
IF INPUT POWER





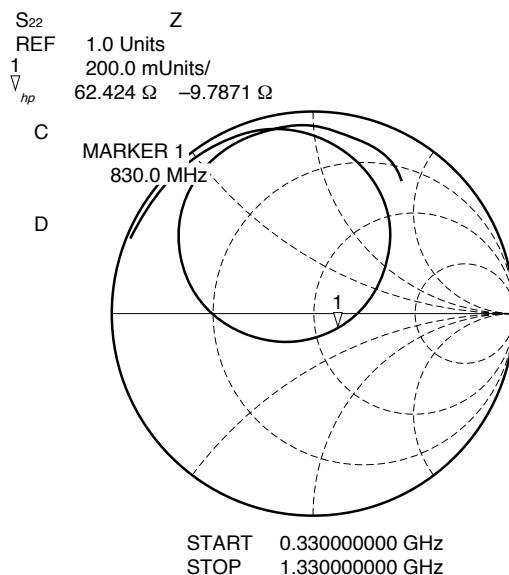
LOCAL LEAKAGE AT RF PIN vs.
LOCAL INPUT FREQUENCYLOCAL LEAKAGE AT RF PIN vs.
LOCAL INPUT POWERLOCAL LEAKAGE AT IF PIN vs.
LOCAL INPUT FREQUENCYLOCAL LEAKAGE AT IF PIN vs.
LOCAL INPUT POWERIF LEAKAGE AT RF PIN vs.
IF INPUT FREQUENCYIF LEAKAGE AT RF PIN vs.
IF INPUT POWER

★ **12. S-PARAMETERS FOR EACH PORT ($V_{cc} = V_{RFout} = 2.8$ V)**
 (The parameters are monitored at DUT pins)

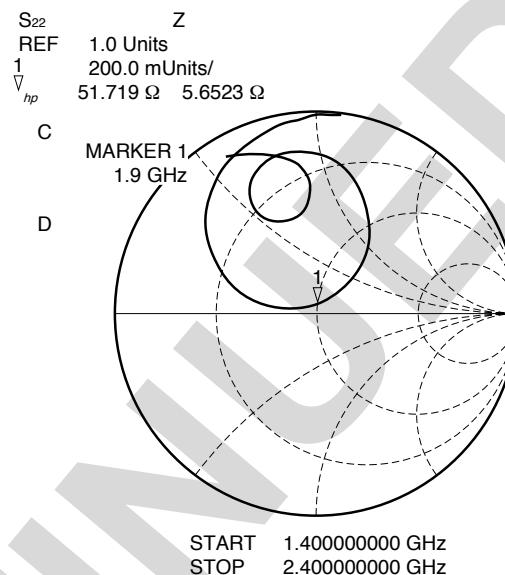
LO port**RF port (without matching)****IF port**

- ★ 13. S-PARAMETERS FOR MATCHED RF OUTPUT ($V_{CC} = V_{RFout} = 2.8$ V)
- ON EVALUATION BOARD - (S₂₂ data are monitored at RF connector on board)

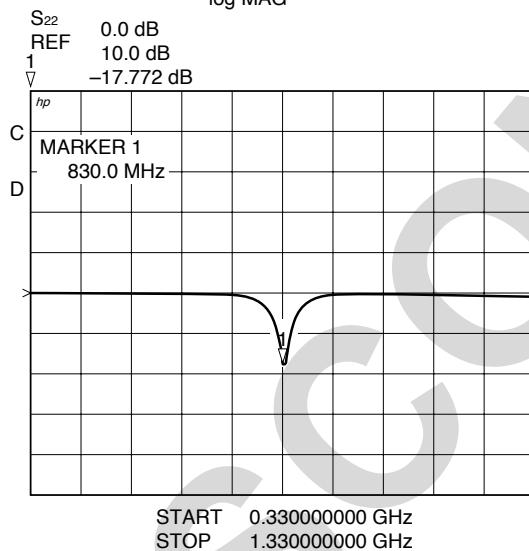
0.83 GHz (matched in test circuit 1)



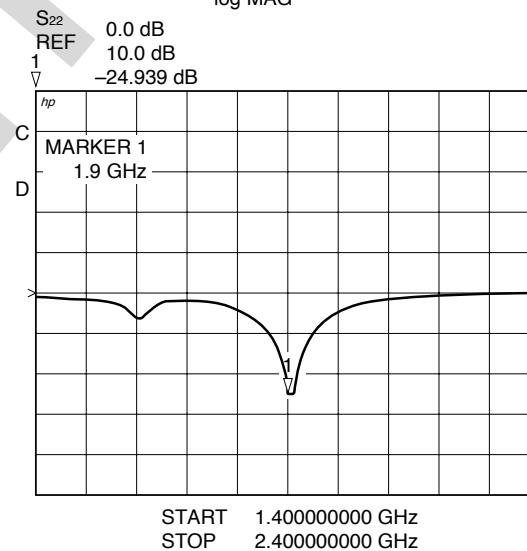
1.9 GHz (matched in test circuit 2)

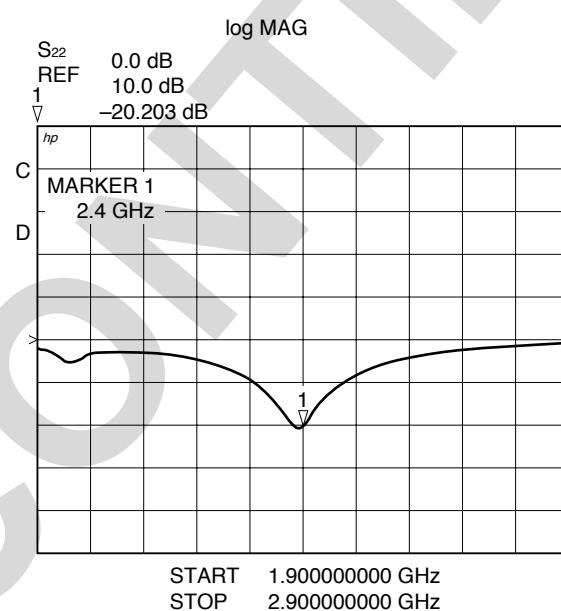
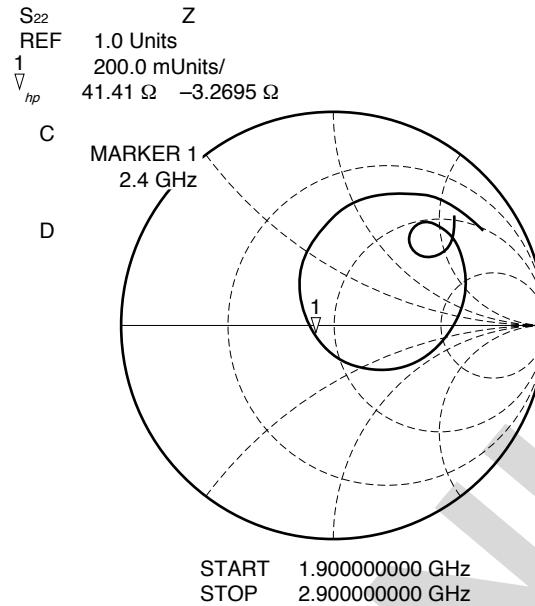


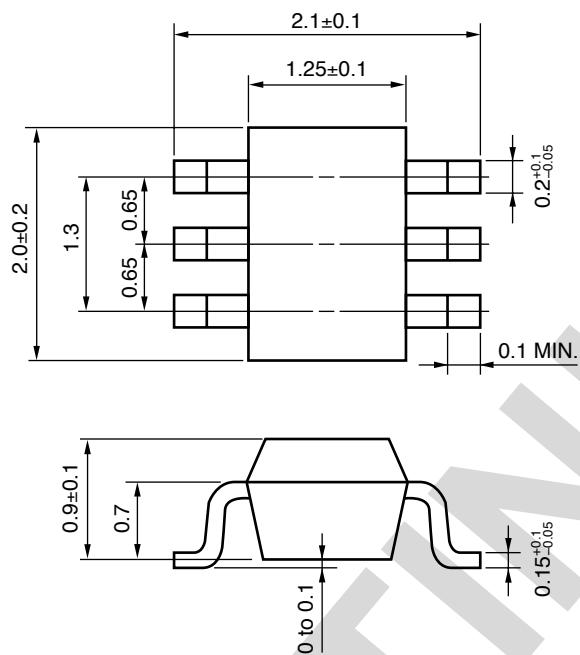
log MAG



log MAG



2.4 GHz (matched in test circuit 3)

14. PACKAGE DIMENSIONS**6-PIN SUPER MINIMOLD (UNIT: mm)**

15. NOTE ON CORRECT USE

- (1) Observe precautions for handling because of electrostatic sensitive devices.
- (2) Form a ground pattern as wide as possible to minimize ground impedance (to prevent undesired oscillation).
- (3) Connect a bypass capacitor to the V_{cc} pin.
- (4) Connect a matching circuit to the RF output pin.
- (5) The DC cut capacitor must be each attached to the input and output pins.

16. RECOMMENDED SOLDERING CONDITIONS

This product should be soldered under the following recommended conditions. For soldering methods and conditions other than those recommended below, contact your NEC sales representative.

| Soldering Method | Soldering Conditions | Recommended Condition Symbol |
|------------------|---|------------------------------|
| Infrared Reflow | Package peak temperature: 235°C or below Time: 30 seconds or less (at 210°C) Count: 3, Exposure limit: None ^{Note} | IR35-00-3 |
| VPS | Package peak temperature: 215°C or below Time: 40 seconds or less (at 200°C) Count: 3, Exposure limit: None ^{Note} | VP15-00-3 |
| Wave Soldering | Soldering bath temperature: 260°C or below Time: 10 seconds or less Count: 1, Exposure limit: None ^{Note} | WS60-00-1 |
| Partial Heating | Pin temperature: 300°C Time: 3 seconds or less (per side of device) Exposure limit: None ^{Note} | - |

Note After opening the dry pack, keep it in a place below 25°C and 65% RH for the allowable storage period.

Caution Do not use different soldering methods together (except for partial heating).

For details of recommended soldering conditions for surface mounting, refer to information document SEMICONDUCTOR DEVICE MOUNTING TECHNOLOGY MANUAL (C10535E).