1N5823, 1N5824, 1N5825 (SILICON)



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NOTE 1: DETERMINING MAXIMUM RATINGS

Reverse power dissipation and the possibility of thermal runaway must be considered when operating this rectifier at reverse voltages above 0.1 $V_{\ensuremath{\mathsf{RWM}}\xspace}$. Proper derating may be accomplished by use of equation (1)

 $T_{A(max)} = T_{J(max)} - R_{\theta JA} P_{F(AV)} - R_{\theta JA} P_{R(AV)}$ (1)where

T_{A(max)} = Maximum allowable ambient temperature

T_{J(max)} = Maximum allowable junction temperature (125°C or the temperature at which thermal runaway occurs, whichever is lowest).

P_{F(AV)} = Average forward power dissipation

PR(AV) = Average reverse power dissipation

 $R_{\theta JA}$ = Junction-to-ambient thermal resistance

Figures 1, 2 and 3 permit easier use of equation (1) by taking reverse power dissipation and thermal runaway into consideration. The figures solve for a reference temperature as determined by equation (2): (2)

 $T_R = T_{J(max)} - R_{\theta JA} P_{R(AV)}$

Substituting equation (2) into equation (1) yields:

 $T_{A(max)} = T_{R} - R_{\theta JA} P_{F(AV)}$ (3)Inspection of equations (2) and (3) reveals that TR is the ambient temperature at which thermal runaway occurs or where $T_J = 125^{\circ}C$, when forward power is zero. The transition from one boundary condition to the other is evident on the curves of Figures 1, 2 and

3 as a difference in the rate of change of the slope in the vicinity of 115°C. The data of Figures 1, 2 and 3 is based upon dc conditions. For use in common rectifier circuits, Table I indicates suggested factors for an equivalent dc voltage to use for conservative design; i.e.

(4) VR(equiv) = VIN(PK) × F

The Factor F is derived by considering the properties of the various rectifier circuits and the reverse characteristics of Schottky diodes.

Example: Find TA(max) for 1N5825 operated in a 12-Volt dc supply using a bridge circuit with capacitive filter such that I_{DC} = 10 A (IF(AV) = 5 A), I(PK)/I(AV) = 10, Input Voltage = 10 V(rms), $R_{\theta}JA = 10^{\circ}C/W$.

- Find VR(equiv). Read F = 0.65 from Table I \therefore Step 1: VR(equiv) = (1.41)(10)(0.65) = 9.2 V
- Find T_R from Figure 3. Read T_R = 118^oC @ V_R = Step 2: 9.2 V & $R_{\theta JA} = 10^{\circ}$ C/W.
- Find $P_{F(AV)}$ from Figure 4. Thead $P_{F(AV)} = 5.5$ W Step 3: $a = \frac{|(PK)|}{|F(\Delta V)|} = 5 A$

Find $T_{A(max)}$ from equation (3). $T_{A(max)} = 118-(10)$ Step 4: $(5.5) = 63^{\circ}C.$

tValues given are for the 1N5825. Power is slightly lower for the other units because of their lower forward voltage.

TABLE I - VALUES FOR FACTOR F

Half Wave		Full Wave, Bridge		Full Wave, Center Tapped ^{(1),(2)}	
Resistive	Capacitive (1)	Resistive	Capacitive	Resistive	Capacitive
0.5	1.3	0.5	0.65	1.0	1.3
0.75	1.5	0.75	0.75	1.5	1.5
	Resistive	ResistiveCapacitive (1)0.51.3	ResistiveCapacitive (1)Resistive0.51.30.5	Resistive Capacitive (1) Resistive Capacitive 0.5 1.3 0.5 0.65	Half Wave Full Wave, Bridge Center Ta Resistive Capacitive (1) Resistive Capacitive Resistive 0.5 1.3 0.5 0.65 1.0

(1) Note that VR(PK)≈2 Vin(PK) (2)Use line to center tap voltage for Vin.





FIGURE 3 - MAXIMUM REFERENCE TEMPERATURE - 1N5825











THERMAL CHARACTERISTICS





FIGURE 6 - APPROXIMATE THERMAL CIRCUIT MODEL



Use of the above model permits calculation of average junction temperature for any mounting situation. Lowest values of thermal resistance will occur when the cathode lead is brought as close as possible to a heat dissipator; as heat conduction through the anode lead is small. Terms in the model are defined as follows:

Case temperature reference

is at cathode end.

TEMPERATURES

T_A = Ambient

- T_{AA} = Anode Heat Sink Ambient
- T_{AK} = Cathode Heat Sink Ambient
- T_{LA} = Anode Lead
- TLK = Cathode Lead
- $T_{j} = Junction$

THERMAL RESISTANCES

 $R_{\theta CA} = Case to Ambient$

- $R_{\theta SA}$ = Anode Lead Heat Sink to Ambient
- $R_{\theta SK}$ = Cathode Lead Heat Sink to Ambient
- $R_{\theta LA} = Anode Lead$

- $R_{\theta}LK = Cathode Lead$ $R_{\theta}CL = Case to Cathode Lead$
- $R_{\theta JC}$ = Junction to Case
- $R_{\theta JA}$ = Junction to Anode Lead (S bend)

NOTE 3 - MOUNTING DATA



FIGURE 10 - CAPACITANCE 2500 2000 Tj = 25°C 1500 C. CAPACITANCE (pF) 1000 ++ 1N5823 700 1N5824 500 1N5823 – 20 V 1N5824 – 30 V 1N5825 – 40 V 1N5825 400 300 250 0.04 0.06 0.1 0.2 0.4 0.6 1.0 2.0 4.0 6.0 10 20 40 VR. REVERSE VOLTAGE (VOLTS)

FIGURE 8 - MAXIMUM SURGE CAPABILITY







NOTE 4 - HIGH FREQUENCY OPERATION

Since current flow in a Schottky rectifier is the result of majority carrier conduction, it is not subject to junction diode forward and reverse recovery transients due to minority carrier injection and stored charge. Satisfactory circuit analysis work may be performed by using a model consisting of an ideal diode in parallel with a variable capacitance. (See Figure 10).

Rectification efficiency measurements show that operation will be satisfactory up to several megahertz. For example, relative waveform rectification efficiency is approximately 70 per cent at 2.0 MHz, e.g., the ratio of dc power to RMS power in the load is 0.28 at this frequency, whereas perfect rectification would yield 0.406 for sine wave inputs. However, in contrast to ordinary junction diodes, the loss in waveform efficiency is not indicative of power loss; it is simply a result of reverse current flow through the diode capacitance, which lowers the dc output voltage.