

# 20 GHz to 42 GHz, Wideband I/Q Mixer

### **Data Sheet**

# HMC8192LG

#### FEATURES

Passive, wideband I/Q mixer RF and LO range: 20 GHz to 42 GHz Wide IF bandwidth of dc to 5 GHz Single-ended RF, LO, and IF Conversion loss: 9 dB typical, 20 GHz to 32 GHz Image rejection: 25 dBc typical, 20 GHz to 32 GHz Noise figure: 12 dB typical Input IP3 (downconverter): 24 dBm typical, 20 GHz to 32 GHz Input P1dB (downconverter) compression: 17 dBm typical, 20 GHz to 32 GHz Input IP2: 55 dBm typical, 20 GHz to 32 GHz LO to RF isolation: 42 dB, 20 GHz to 32 GHz LO to IFx isolation: 45 dB, 20 GHz to 32 GHz RF to IF isolation: 35 dB, 20 GHz to 32 GHz Amplitude balance: ±1 dB typical Phase balance (downconverter): ±8° typical RF return loss: 12 dB typical LO return loss: 10 dB typical IFx return loss: 20 dB typical Exposed pad, 4.00 mm × 4.00 mm, 25-terminal LGA\_CAV package

#### **APPLICATIONS**

Test and measurement instrumentation Military, radar, aerospace, and defense applications Microwave point to point base stations

#### **GENERAL DESCRIPTION**

The HMC8192LG is a passive, wideband, inphase/quadrature (I/Q), monolithic microwave integrated circuit (MMIC) mixer that can be used either as an image rejection mixer for receiver operations or as a single-sideband upconverter for transmitter operations. With a radio frequency (RF) and local oscillator (LO) range of 20 GHz to 42 GHz, and an intermediate frequency (IF) bandwidth of dc to 5 GHz, the HMC8192LG is ideal for applications requiring a wide frequency range, excellent RF performance, and a simple design with fewer components and a small printed circuit board (PCB) footprint. A single HMC8192LG can replace multiple narrow-band mixers in a design.

The inherent I/Q architecture of the HMC8192LG offers excellent image rejection, eliminating the need for expensive filtering for unwanted sidebands. The mixer also provides

#### FUNCTIONAL BLOCK DIAGRAM



excellent LO to RF and LO to IF isolation and reduces the effect of LO leakage to ensure signal integrity.

As a passive mixer, the HMC8192LG does not require any dc power sources. The HMC8192LG offers a lower noise figure compared to an active mixer, ensuring superior dynamic range for high performance and precision applications.

The HMC8192LG is fabricated on a gallium arsenide (GaAs), metal semiconductor field effect transistor (MESFET) process and uses Analog Devices, Inc., mixer cells and a 90° hybrid. The HMC8192LG is available in a compact, 4.00 mm × 4.00 mm, 25-terminal land grid array cavity (LGA\_CAV) package and operates over a  $-40^{\circ}$ C to  $+85^{\circ}$ C temperature range. The evaluation board for the HMC8192LG, EV1HMC8192LG, is also available on the Analog Devices website.

Rev. 0

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### **SPECIFICATIONS**

#### 20 GHz TO 32 GHz

 $T_A = 25^{\circ}$ C, IF = 100 MHz, LO drive = 18 dBm, all measurements performed as downconverter with upper sideband selected, external 90° hybrid at the IFx ports, and LO amplifier in line with lab bench LO source, unless otherwise noted.

Table 1.									
Parameter	Symbol	Min	Тур	Max	Unit				
FREQUENCY									
Radio	RF	20		32	GHz				
LO	f <sub>LO</sub>	20		32	GHz				
Intermediate	IF	dc		5	GHz				
LO DRIVE LEVEL		16	18	20	dBm				
RF PERFORMANCE AS DOWNCONVERTER									
Conversion Loss			9	10	dB				
Image Rejection		15	25		dBc				
Single-Sideband Noise Figure	SSB NF		12		dB				
Input Third-Order Intercept	IP3	22	24		dBm				
Input 1 dB Compression Point	P1dB		17		dBm				
Input Second-Order Intercept	IP2		55		dBm				
Amplitude Balance <sup>1</sup>			±1		dB				
Phase Balance <sup>1</sup>			±8		Degrees				
RF PERFORMANCE AS UPCONVERTER									
Conversion Loss			9		dB				
Sideband Rejection			18		dBc				
Input Third-Order Intercept	IP3		21		dBm				
Input 1 dB Compression Point	P1dB		14		dBm				
ISOLATION PERFORMANCE									
LO to RF		36	42		dB				
LO to IFx <sup>1</sup>			45		dB				
RF to IF <sup>1</sup>			35		dB				
RETURN LOSS PERFORMANCE <sup>1</sup>									
RF			12		dB				
LO			10		dB				
IFx			20		dB				

<sup>1</sup> Measurements taken without 90° hybrid at the IFx ports.

#### 32 GHz TO 42 GHz

 $T_A = 25^{\circ}$ C, IF = 100 MHz, LO drive = 18 dBm, all measurements performed as downconverter with upper sideband selected, external 90° hybrid at the IFx ports, and LO amplifier in line with lab bench LO source, unless otherwise noted.

Parameter	Symbol	Min	Тур	Max	Unit
FREQUENCY					
Radio	RF	32		42	GHz
LO	f <sub>LO</sub>	32		42	GHz
Intermediate	IF	dc		5	GHz
LO DRIVE LEVEL		16	18	20	dBm
RF PERFORMANCE AS DOWNCONVERTER					
Conversion Loss			11	15	dB
Image Rejection		15	20		dBc
Single-Sideband Noise Figure	SSB NF		12		dB
Input Third-Order Intercept	IP3	17	23		dBm
Input 1 dB Compression Point	P1dB		18		dBm
Input Second-Order Intercept	IP2		50		dBm
Amplitude Balance <sup>1</sup>			±1		dB
Phase Balance <sup>1</sup>			±8		Degrees
RF PERFORMANCE AS UPCONVERTER					
Conversion Loss			10		dB
Sideband Rejection			18		dBc
Input Third-Order Intercept	IP3		20		dBm
Input 1 dB Compression Point	P1dB		14		dBm
ISOLATION PERFORMANCE					
LO to RF		24	42		dB
LO to IFx <sup>1</sup>			43		dB
RF to IF <sup>1</sup>			40		dB
RETURN LOSS PERFORMANCE <sup>1</sup>					
RF			12		dB
LO			10		dB
IFx			20		dB

<sup>1</sup> Measurements taken without 90° hybrid at the IFx ports.

### **ABSOLUTE MAXIMUM RATINGS**

#### Table 3.

Parameter	Rating
RF Input Power	23 dBm
LO Input Power	23 dBm
IF Input Power	23 dBm
IF Source/Sink Current	3 mA
Continuous Power Dissipation, P <sub>DISS</sub> (T <sub>A</sub> = 85°C, Derate 7.2 mW/°C Above 85°C) <sup>1</sup>	647 mW
Maximum Junction Temperature (T <sub>J</sub> )	175°C
Lifetime at Maximum T	$>1 \times 10^6$ hours
Moisture Sensitivity Level (MSL) <sup>2</sup>	3
Maximum Peak Reflow Temperature (MLS3)	260°C
Operating Temperature Range	-40°C to +85°C
Storage Temperature Range	–65°C to +150°C
Electrostatic Discharge Sensitivity	
Human Body Model	4000 V
Field Induced Charged Device Model	1250 V

 $^{1}$  P<sub>DISS</sub> is a theoretical number calculated by (T<sub>J</sub> – 85°C)/ $\theta_{JC}$ .

<sup>2</sup> Based on IPC/JEDEC J-STD-20 MSL classifications.

Stresses at or above those listed under Absolute Maximum Ratings may cause permanent damage to the product. This is a stress rating only; functional operation of the product at these or any other conditions above those indicated in the operational section of this specification is not implied. Operation beyond the maximum operating conditions for extended periods may affect product reliability.

#### THERMAL RESISTANCE

Thermal performance is directly linked to printed circuit board (PCB) design and operating environment. Careful attention to PCB thermal design is required.

 $\theta_{JA}$  is the junction to ambient (or die to ambient) thermal resistance measured in a one cubic foot sealed enclosure, and  $\theta_{JC}$  is the junction to case (or die to package) thermal resistance.

#### **Table 4. Thermal Resistance**

Package Type	θ」Α	οις	Unit					
CE-25-11	120	139	°C/W					

 $^1$  Thermal impedance simulated values are based on a JEDEC 2S2P test board with 4  $\times$  4 thermal vias. Refer to JEDEC standard JESD51-2 for additional information.

#### **ESD CAUTION**



**ESD** (electrostatic discharge) sensitive device. Charged devices and circuit boards can discharge without detection. Although this product features patented or proprietary protection circuitry, damage may occur on devices subjected to high energy ESD. Therefore, proper ESD precautions should be taken to avoid performance degradation or loss of functionality.

### **PIN CONFIGURATION AND FUNCTION DESCRIPTIONS**



Figure 2. Pin Configuration

#### Table 5. Pin Function Descriptions

Pin No.	Mnemonic	Description
1 to 8, 10, 12 to 15, 17 to 23, 25	GND	Ground Connections. These pins and the package bottom must be connected to RF/dc ground. See Figure 3 for the interface schematic.
9, 11	IF1, IF2	First and Second Quadrature IF Input/Output Pins. These pins are dc-coupled. For applications not requiring operation to dc, use an off-chip dc blocking capacitor. For operations to dc, these pins must not source or sink more than 3 mA of current. Otherwise, the device may not function and may fail. See Figure 4 for the interface schematic.
16	RF	RF Input/Output. This pin is ac-coupled and matched to 50 $\Omega$ . See Figure 5 for the interface schematic.
24	LO	LO Input. This pin is ac-coupled and matched to 50 $\Omega$ . See Figure 6 for the interface schematic.
	EPAD	Exposed Pad. The exposed pad must be connected to RF/dc ground.

#### **INTERFACE SCHEMATICS**

GND  $\stackrel{\bullet}{=}$ Figure 3. GND Interface Schematic

Figure 4. IF1 and IF2 Interface Schematic

005 15697-Figure 5. RF Interface Schematic



Figure 6. LO Interface Schematic

### **TYPICAL PERFORMANCE CHARACTERISTICS**

DOWNCONVERTER PERFORMANCE: IF = 100 MHz, UPPER SIDEBAND (LOW-SIDE LO)



Figure 7. Conversion Gain vs. RF Frequency at Various Temperatures, LO Drive = 18 dBm



Figure 8. Image Rejection vs. RF Frequency at Various Temperatures, LO Drive = 18 dBm



ure 9. Input IP3 vs. RF Frequency at Various Tempera LO Drive = 18 dBm



Figure 10. Conversion Gain vs. RF Frequency at Various LO Drives,  $T_A = 25^{\circ}$ C



Figure 11. Image Rejection vs. RF Frequency at Various LO Drives,  $T_A = 25^{\circ}$ C



Figure 12. Input IP3 vs. RF Frequency at Various LO Drives,  $T_A = 25^{\circ}C$ 



Figure 13. Noise Figure vs. RF Frequency at Various Temperatures, LO Drive = 18 dBm



Figure 14. Input IP2 vs. RF Frequency at Various Temperatures, LO Drive = 18 dBm





Figure 16. Noise Figure vs. RF Frequency at Various LO Drives,  $T_A = 25^{\circ}C$ 



Figure 17. Input IP2 vs. RF Frequency at Various LO Drives,  $T_A = 25^{\circ}$ C

#### DOWNCONVERTER PERFORMANCE: IF = 2500 MHz, UPPER SIDEBAND (LOW-SIDE LO)

CONVERSION GAIN (dB)



Figure 18. Conversion Gain vs. RF Frequency at Various Temperatures, LO Drive = 18 dBm



Figure 19. Image Rejection vs. RF Frequency at Various Temperatures, LO Drive = 18 dBm



0 14dBm 16dBm 18dBm -2 20dBm -6 -8 -10 -12 -14 -16 -18 -20 5697-021 20 22 24 26 28 30 32 34 36 38 40 42 **RF FREQUENCY (GHz)** 

Figure 21. Conversion Gain vs. RF Frequency at Various LO Drives,  $T_A = 25^{\circ}C$ 



Figure 22. Image Rejection vs. RF Frequency at Various LO Drives,  $T_A = 25^{\circ}C$ 



Figure 23. Input IP3 vs. RF Frequency at Various LO Drives,  $T_A = 25^{\circ}C$ 



Figure 24. Noise Figure vs. RF Frequency at Various Temperatures, LO Drive = 18 dBm



Figure 25. Input IP2 vs. RF Frequency at Various Temperatures, LO Drive = 18 dBm





Figure 27. Noise Figure vs. RF Frequency at Various LO Drives,  $T_A = 25^{\circ}C$ 



Figure 28. Input IP2 vs. RF Frequency at Various LO Drives,  $T_A = 25^{\circ}$ C

#### DOWNCONVERTER PERFORMANCE: IF = 5000 MHz, UPPER SIDEBAND (LOW-SIDE LO)



Figure 29. Conversion Gain vs. RF Frequency at Various Temperatures, LO Drive = 18 dBm



Figure 30. Image Rejection vs. RF Frequency at Various Temperatures, LO Drive = 18 dBm



0 14dBm 16dBm 18dBm -2 20dBm CONVERSION GAIN (dB) -6 -8 -10 -12 -14 -16 -18 -20 5697-032 20 22 24 26 28 30 32 34 36 38 40 42 **RF FREQUENCY (GHz)** 

Figure 32. Conversion Gain vs. RF Frequency at Various LO Drives,  $T_A = 25^{\circ}C$ 



Figure 33. Image Rejection vs. RF Frequency at Various LO Drives,  $T_A = 25^{\circ}C$ 



Figure 34. Input IP3 vs. RF Frequency at Various LO Drives,  $T_A = 25^{\circ}C$ 



Figure 35. Noise Figure vs. RF Frequency at Various Temperatures, LO Drive = 18 dBm



Figure 36. Input IP2 vs. RF Frequency at Various Temperatures, LO Drive = 18 dBm



Figure 37. Input P1dB vs. RF Frequency at Various LO Drives,  $T_A = 25^{\circ}C$ 



Figure 38. Noise Figure vs. RF Frequency at Various LO Drives,  $T_A = 25^{\circ}C$ 



Figure 39. Input IP2 vs. RF Frequency at Various LO Drives,  $T_A = 25^{\circ}C$ 

#### DOWNCONVERTER PERFORMANCE: IF = 2500 MHz, LOWER SIDEBAND (HIGH-SIDE LO)



Figure 40. Conversion Gain vs. RF Frequency at Various Temperatures, LO Drive = 18 dBm



Figure 41. Image Rejection vs. RF Frequency at Various Temperatures, LO Drive = 18 dBm



0 14dBm 16dBm 18dBm -2 20dBm CONVERSION GAIN (dB) -6 -8 -10 -12 -14 -16 -18 -20 5697-043 20 22 24 26 28 30 32 34 36 38 40 **RF FREQUENCY (GHz)** 

Figure 43. Conversion Gain vs. RF Frequency at Various LO Drives,  $T_A = 25^{\circ}C$ 



Figure 44. Image Rejection vs. RF Frequency at Various LO Drives,  $T_A = 25^{\circ}$ C



Figure 45. Input IP3 vs. RF Frequency at Various LO Drives,  $T_A = 25^{\circ}C$ 



Figure 46. Noise Figure vs. RF Frequency at Various Temperatures, LO Drive = 18 dBm











Figure 49. Noise Figure vs. RF Frequency at Various LO Drives,  $T_A = 25^{\circ}C$ 



Figure 50. Input IP2 vs. RF Frequency at Various LO Drives,  $T_A = 25^{\circ}$ C

#### DOWNCONVERTER PERFORMANCE: IF = 5000 MHz, LOWER SIDEBAND (HIGH-SIDE LO)



Figure 51. Conversion Gain vs. RF Frequency at Various Temperatures, LO Drive = 18 dBm



Figure 52. Image Rejection vs. RF Frequency at Various Temperatures, LO Drive = 18 dBm







Figure 54. Conversion Gain vs. RF Frequency at Various LO Drives,  $T_A = 25^{\circ}C$ 



Figure 55. Image Rejection vs. RF Frequency at Various LO Drives,  $T_A = 25^{\circ}C$ 



Figure 56. Input IP3 vs. RF Frequency at Various LO Drives,  $T_A = 25^{\circ}C$ 







Figure 58. Input IP2 vs. RF Frequency at Various Temperatures, LO Drive = 18 dBm



LO Drive = 18 dBm



Figure 60. Noise Figure vs. RF Frequency at Various LO Drives,  $T_A = 25^{\circ}C$ 



Figure 61. Input IP2 vs. RF Frequency at Various LO Drives,  $T_A = 25^{\circ}$ C

#### UPCONVERTER PERFORMANCE: IF = 100 MHz, UPPER SIDEBAND



Figure 62. Conversion Gain vs. RF Frequency at Various Temperatures, LO Drive = 18 dBm



Figure 63. Sideband Rejection vs. RF Frequency at Various Temperatures, LO Drive = 18 dBm



LO Drive = 18 dBm



Figure 65. Conversion Gain vs. RF Frequency at Various LO Drives,  $T_A = 25^{\circ}C$ 



Figure 66. Sideband Rejection vs. RF Frequency at Various LO Drives,  $T_A = 25^{\circ}C$ 



Figure 67. Input IP3 vs. RF Frequency at Various LO Drives,  $T_A = 25^{\circ}C$ 



#### UPCONVERTER PERFORMANCE: IF = 2500 MHz, UPPER SIDEBAND



Figure 69. Conversion Gain vs. RF Frequency at Various Temperatures, LO Drive = 18 dBm



Figure 70. Sideband Rejection vs. RF Frequency at Various Temperatures, LO Drive = 18 dBm



LO Drive = 18 dBm



Figure 72. Conversion Gain vs. RF Frequency at Various LO Drives,  $T_A = 25^{\circ}C$ 



Figure 73. Sideband Rejection vs. RF Frequency at Various LO Drives,  $T_A = 25^{\circ}C$ 



Figure 74. Input IP3 vs. RF Frequency at Various LO Drives,  $T_A = 25^{\circ}C$ 



#### UPCONVERTER PERFORMANCE: IF = 5000 MHz, UPPER SIDEBAND



Figure 76. Conversion Gain vs. RF Frequency at Various Temperatures, LO Drive = 18 dBm



Figure 77. Sideband Rejection vs. RF Frequency at Various Temperatures, LO Drive = 18 dBm





Figure 79. Conversion Gain vs. RF Frequency at Various LO Drives,  $T_A = 25^{\circ}$ C



Figure 80. Sideband Rejection vs. RF Frequency at Various LO Drives,  $T_A = 25^{\circ}C$ 





#### UPCONVERTER PERFORMANCE: IF = 2500 MHz, LOWER SIDEBAND



Figure 83. Conversion Gain vs. RF Frequency at Various Temperatures, LO Drive = 18 dBm



Figure 84. Sideband Rejection vs. RF Frequency at Various Temperatures, LO Drive = 18 dBm





Figure 86. Conversion Gain vs. RF Frequency at Various LO Drives,  $T_A = 25^{\circ}C$ 



Figure 87. Sideband Rejection vs. RF Frequency at Various LO Drives,  $T_A = 25^{\circ}C$ 



Figure 88. Input IP3 vs. RF Frequency at Various LO Drives,  $T_A = 25^{\circ}C$ 



#### UPCONVERTER PERFORMANCE: IF = 5000 MHz, LOWER SIDEBAND



Figure 90. Conversion Gain vs. RF Frequency at Various Temperatures, LO Drive = 18 dBm



Figure 91. Sideband Rejection vs. RF Frequency at Various Temperatures, LO Drive = 18 dBm



LO Drive = 18 dBm



Figure 93. Conversion Gain vs. RF Frequency at Various LO Drives,  $T_A = 25^{\circ}C$ 



Figure 94. Sideband Rejection vs. RF Frequency at Various LO Drives,  $T_A = 25^{\circ}C$ 



Figure 95. Input IP3 vs. RF Frequency at Various LO Drives,  $T_A = 25^{\circ}C$ 



#### ISOLATION AND RETURN LOSS WITHOUT EXTERNAL 90° HYBRID AT THE IFx PORTS



Figure 97. LO to IF Isolation vs. LO Frequency at Various Temperatures, IF = 100 MHz, LO Drive = 18 dBm



Figure 98. LO to RF Isolation vs. LO Frequency at Various Temperatures, IF = 100 MHz, LO Drive = 18 dBm



Figure 99. RF to IF Isolation vs. RF Frequency at Various Temperatures, LO Drive = 18 dBm



Figure 100. LO to IF Isolation vs. LO Frequency at Various LO Drives, IF = 100 MHz,  $T_A = 25$  °C



Figure 101. LO to RF Isolation vs. LO Frequency at Various LO Drives,  $IF = 100 \text{ MHz}, T_A = 25 ^{\circ}\text{C}$ 

















Figure 106. LO Return Loss vs. LO Frequency at Various LO Drives



Figure 107. RF Return Loss vs. RF Frequency at Various LO Drives, LO Frequency = 32 GHz



Figure 108. IF1/IF2 Return Loss vs. IF Frequency at Various LO Drives, LO Frequency = 32 GHz

#### IF BANDWIDTH PERFORMANCE: DOWNCONVERTER, UPPER SIDEBAND (LOW-SIDE LO)

Data over the IF frequency was taken without an external 90° hybrid at the IFx ports.



Figure 109. Conversion Gain vs. IF Frequency at Various Temperatures, LO Drive = 18 dBm at 32 GHz



Igure 110. Input 193 vs. IF Frequency at Various Temperatu LO Drive = 18 dBm at 32 GHz



Figure 111. Conversion Gain vs. IF Frequency at Various LO Drives, LO Frequency = 32 GHz,  $T_A = 25^{\circ}\text{C}$ 



LO Frequency = 32 GHz,  $T_A = 25^{\circ}C$ 

#### IF BANDWIDTH PERFORMANCE: DOWNCONVERTER, LOWER SIDEBAND (HIGH-SIDE LO)

Data over the IF frequency was taken without an external 90° hybrid at the IFx ports.



Figure 113. Conversion Gain vs. IF Frequency at Various Temperatures, LO Drive = 18 dBm at 32 GHz



LO Drive = 18 dBm at 32 GHz



Figure 115. Conversion Gain vs. IF Frequency at Various LO Drives, LO Frequency = 32 GHz,  $T_A = 25^{\circ}C$ 



igure 116. Input IP3 vs. IF Frequency at Various LO Drives, LO Frequency = 32 GHz,  $T_A = 25^{\circ}$ C

#### AMPLITUDE AND PHASE IMBALANCE PERFORMANCE: DOWNCONVERTER, UPPER SIDEBAND (LOW-SIDE LO)



Figure 117. Amplitude Imbalance vs. RF Frequency at Various Temperatures, LO Drive = 18 dBm, IF = 100 MHz



Figure 118. Phase Imbalance vs. RF Frequency at Various Temperatures, LO Drive = 18 dBm, IF = 100 MHz



Figure 119. Amplitude Imbalance vs. RF Frequency at Various LO Drives, IF = 100 MHz,  $T_A = 25^{\circ}$ C



Figure 120. Phase Imbalance vs. RF Frequency at Various LO Drives,  $IF = 100 \text{ MHz}, T_A = 25 ^{\circ}\text{C}$ 

#### SPURIOUS AND HARMONICS PERFORMANCE

Data was taken without an IF hybrid at the IFx ports. N/A means not applicable.

#### Downconverter M × N Spurious Outputs

Mixer spurious products are measured in dBc from the IF output power level, unless otherwise specified. Spur values are (M  $\times$  RF) – (N  $\times$  LO).

IF = 100 MHz, RF = 20,000 MHz, LO = 19,900 MHz, RF power = -10 dBm, LO power = 18 dBm, and T<sub>A</sub> = 25°C.

			N × LO					
		0	1	2	3	4	5	
	0	N/A	10	37	N/A	N/A	N/A	
	1	14	0	45	62	N/A	N/A	
M×RF	2	66	76	66	74	66	N/A	
	3	N/A	66	75	67	74	66	
	4	N/A	N/A	66	74	88	76	
	5	N/A	N/A	N/A	63	75	88	

IF = 100 MHz, RF = 30,000 MHz, LO = 29,900 MHz, RF power = -10 dBm, LO power = 18 dBm, and T<sub>A</sub> = 25°C.

			N × LO						
		0	1	2	3	4	5		
	0	N/A	1	N/A	N/A	N/A	N/A		
	1	19	0	49	N/A	N/A	N/A		
	2	N/A	72	79	71	N/A	N/A		
M×RF	3	N/A	N/A	70	83	72	N/A		
	4	N/A	N/A	N/A	70	88	73		
	5	N/A	N/A	N/A	N/A	69	89		

IF = 100 MHz, RF = 40,000 MHz, LO = 39,900 MHz, RF power = -10 dBm, LO power = 18 dBm, and  $T_{\rm A}$  = 25°C.

		N × LO						
		0	1	2	3	4	5	
M×RF	0	N/A	-4	N/A	N/A	N/A	N/A	
	1	+17	N/A	N/A	N/A	N/A	N/A	
	2	N/A	N/A	+56	+61	N/A	N/A	
	3	N/A	N/A	+62	+79	+61	N/A	
	4	N/A	N/A	N/A	+62	+81	N/A	
	5	N/A	N/A	N/A	N/A	+58	+86	

IF = 2500 MHz, RF = 20,000 MHz, LO = 17,500 MHz, RF power = -10 dBm, LO power = 18 dBm, and  $T_A = 25^{\circ}C$ .

			N × LO						
		0	1	2	3	4	5		
M×RF	0	N/A	12	32	N/A	N/A	N/A		
	1	14	0	26	47	59	N/A		
	2	64	73	72	63	71	60		
	3	N/A	63	72	80	79	73		
	4	N/A	N/A	64	73	77	80		
	5	N/A	N/A	N/A	60	71	64		

IF = 2500 MHz, RF = 30,000 MHz, LO = 27,500 MHz, RF power = -10 dBm, LO power = 18 dBm, and  $T_A = 25^{\circ}$ C.

			N × LO							
		0	1	2	3	4	5			
M×RF	0	N/A	8	N/A	N/A	N/A	N/A			
	1	19	0	52	N/A	N/A	N/A			
	2	N/A	69	66	70	60	N/A			
	3	N/A	N/A	68	80	75	57			
	4	N/A	N/A	N/A	67	76	76			
	5	N/A	N/A	N/A	N/A	65	80			

IF = 2500 MHz, RF = 40,000 MHz, LO = 37,500 MHz, RF power = -10 dBm, LO power = 18 dBm, and  $T_A = 25^{\circ}$ C.

			N × LO							
		0	1	2	3	4	5			
-	0	N/A	-8	N/A	N/A	N/A	N/A			
	1	+17	0	+43	N/A	N/A	N/A			
M×RF	2	N/A	+59	+59	+66	N/A	N/A			
M × KF	3	N/A	N/A	+57	+76	+65	N/A			
	4	N/A	N/A	N/A	+55	+77	+66			
	5	N/A	N/A	N/A	N/A	+57	+77			

IF = 5000 MHz, RF = 20,000 MHz, LO = 15,000 MHz, RF power = -10 dBm, LO power = 18 dBm, and  $T_A = 25^{\circ}$ C.

			N × LO							
		0	1	2	3	4	5			
	0	N/A	-4	+26	+30	N/A	N/A			
	1	+13	0	+24	+35	+46	N/A			
M×RF	2	+45	+35	+24	N/A	+13	+53			
М X КГ	3	N/A	+30	+26	-4	N/A	-4			
	4	N/A	N/A	+54	+53	+13	N/A			
	5	N/A	N/A	N/A	N/A	+40	+35			

IF = 5000 MHz, RF = 30,000 MHz, LO = 25,000 MHz, RF power = -10 dBm, LO power = 18 dBm, and  $T_A = 25^{\circ}$ C.

			N × LO						
		0	1	2	3	4	5		
0	0	N/A	6	34	N/A	N/A	N/A		
	1	18	0	37	61	N/A	N/A		
	2	N/A	63	72	73	65	N/A		
M×RF	3	N/A	N/A	65	73	68	65		
	4	N/A	N/A	N/A	61	37	N/A		
	5	N/A	N/A	N/A	NA	34	6		

### **Data Sheet**

### HMC8192LG

IF = 5000 MHz, RF = 40,000 MHz, LO = 35,000 MHz, RF power =
-10 dBm, LO power = 18 dBm, and T <sub>A</sub> = 25°C.

			N × LO						
		0	1	2	3	4	5		
-	0	N/A	4	N/A	N/A	N/A	N/A		
	1	17	0	44	N/A	N/A	N/A		
M×RF	2	N/A	57	65	69	N/A	N/A		
	3	N/A	N/A	56	73	69	N/A		
	4	N/A	N/A	N/A	N/A	72	74		
	5	N/A	N/A	N/A	N/A	N/A	70		

IF = 100 MHz, RF = 20,000 MHz, LO = 20,100 MHz, RF power = -10 dBm, LO power = 18 dBm, and T<sub>A</sub> = 25°C.

			N×LO						
		0	1	2	3	4	5		
	0	N/A	11	36	N/A	N/A	N/A		
	1	14	0	53	63	N/A	N/A		
M×RF	2	69	77	67	75	67	N/A		
	3	N/A	66	77	67	76	66		
	4	N/A	N/A	66	75	88	77		
	5	N/A	N/A	N/A	65	76	89		

IF = 100 MHz, RF = 30,000 MHz, LO = 29,900 MHz, RF power = -10 dBm, LO power = 18 dBm, and  $T_{\rm A}$  = 25°C.

			N×LO						
		0	1	2	3	4	5		
	0	N/A	2	N/A	N/A	N/A	N/A		
	1	19	0	49	N/A	N/A	N/A		
M×RF	2	N/A	73	71	71	N/A	N/A		
	3	N/A	N/A	71	83	71	N/A		
	4	N/A	N/A	N/A	72	90	73		
	5	N/A	N/A	N/A	N/A	74	89		

IF = 100 MHz, RF = 40,000 MHz, LO = 39,900 MHz, RF power = -10 dBm, LO power = 18 dBm, and T<sub>A</sub> = 25°C.

				N×	LO		
		0	1	2	3	4	5
	0	N/A	-3	N/A	N/A	N/A	N/A
	1	+17	0	+46	N/A	N/A	N/A
M×RF	2	N/A	+61	+56	+61	N/A	N/A
	3	N/A	N/A	+61	+83	+61	N/A
	4	N/A	N/A	N/A	+63	+83	+62
	5	N/A	N/A	N/A	N/A	+64	+85

IF = 2500 MHz, RF = 20,000 MHz, LO = 22,500 MHz, RF power = -10 dBm, LO power = 18 dBm, and  $T_{\rm A}$  = 25°C.

				N×	LO		
		0	1	2	3	4	5
M × RF	0	N/A	3	32	N/A	N/A	N/A
	1	15	0	41	59	N/A	N/A
	2	66	77	73	72	61	N/A
	3	N/A	64	78	77	72	N/A
	4	N/A	N/A	67	78	81	72
	5	N/A	N/A	N/A	70	81	80

IF = 2500 MHz, RF = 30,000 MHz, LO = 32,500 MHz, RF power = -10 dBm, LO power = 18 dBm, and  $T_{\rm A}$  = 25°C.

				N×	LO		
		0	1	2	3	4	5
	0	N/A	-2	N/A	N/A	N/A	N/A
	1	+18	0	+41	N/A	N/A	+18
M×RF	2	N/A	+71	+63	+64	N/A	N/A
M × KF	3	N/A	N/A	+72	+79	+65	N/A
	4	N/A	N/A	N/A	+73	+79	N/A
	5	N/A	N/A	N/A	N/A	+76	N/A

IF = 100 MHz, RF = 20,000 MHz, LO = 25,000 MHz, RF power = -10 dBm, LO power = 18 dBm, and T<sub>A</sub> = 25°C.

				N	× LO		
		0	1	2	3	4	5
	0	N/A	7	34	N/A	N/A	N/A
	1	15	0	64	N/A	N/A	N/A
	2	65	80	75	67	N/A	N/A
M×RF	3	N/A	67	77	78	67	N/A
	4	N/A	N/A	67	N/A	15	40
	5	N/A	N/A	40	7	N/A	7

IF = 2500 MHz, RF = 30,000 MHz, LO = 35,000 MHz, RF power = -10 dBm, LO power = 18 dBm, and  $T_{\rm A}$  = 25°C.

			N × LO							
		0	1	2	3	4	5			
	0	N/A	5	N/A	N/A	N/A	N/A			
	1	18	0	48	N/A	N/A	N/A			
M×RF	2	N/A	71	76	60	N/A	N/A			
	3	N/A	N/A	75	75	59	N/A			
	4	N/A	N/A	58	76	74	N/A			
	5	N/A	N/A	N/A	59	76	71			

#### Upconverter M × N Spurious Outputs

Mixer spurious products are measured in dBc from the RF output power level, unless otherwise specified. Hybrid loss is not de-embedded.

IF = 100 MHz, RF = 20,000 MHz, LO = 19,900 MHz, IF power =
$-10$ dBm, LO power = 18 dBm, and $T_A = 25^{\circ}C$ .

				N×	LO		
		0	1	2	3	4	5
	-5	90	79	70	N/A	N/A	N/A
	-4	92	78	70	N/A	N/A	N/A
	-3	93	78	69	N/A	N/A	N/A
	-2	91	67	69	N/A	N/A	N/A
	-1	43	20	31	N/A	N/A	N/A
M×IF	0	N/A	7	24	N/A	N/A	N/A
	+1	43	0	42	N/A	N/A	N/A
	+2	93	60	69	N/A	N/A	N/A
	+3	90	81	70	N/A	N/A	N/A
	+4	92	79	71	N/A	N/A	N/A
	+5	N/A	81	71	N/A	N/A	N/A

IF = 100 MHz, RF = 30,000 MHz, LO = 29,900 MHz, IF power = -10 dBm, LO power = 18 dBm, and  $T_{\rm A}$  = 25°C.

				N×	LO		
		0	1	2	3	4	5
	-5	91	74	N/A	N/A	N/A	N/A
	-4	91	75	N/A	N/A	N/A	N/A
	-3	91	69	N/A	N/A	N/A	N/A
	-2	92	69	N/A	N/A	N/A	N/A
	-1	43	19	N/A	N/A	N/A	N/A
M×IF	0	N/A	N/A	N/A	N/A	N/A	N/A
	+1	43	0	N/A	N/A	N/A	N/A
	+2	91	58	N/A	N/A	N/A	N/A
	+3	92	73	N/A	N/A	N/A	N/A
	+4	91	74	N/A	N/A	N/A	N/A
	+5	N/A	72	N/A	N/A	N/A	N/A

IF = 100 MHz, RF = 40,000 MHz, LO = 39,900 MHz, IF power = -10 dBm, LO power = 18 dBm, and  $T_{\rm A}$  = 25°C.

				N×	LO		
		0	1	2	3	4	5
	-5	+90	+68	N/A	N/A	N/A	N/A
	-4	+89	+66	N/A	N/A	N/A	N/A
	-3	+90	+65	N/A	N/A	N/A	N/A
	-2	+65	+60	N/A	N/A	N/A	N/A
	-1	+41	+18	N/A	N/A	N/A	N/A
M×IF	0	N/A	-8	N/A	N/A	N/A	N/A
	+1	+41	N/A	N/A	N/A	N/A	N/A
	+2	+90	+58	N/A	N/A	N/A	N/A
	+3	+90	+67	N/A	N/A	N/A	N/A
	+4	+90	+66	N/A	N/A	N/A	N/A
	+5	N/A	+67	N/A	N/A	N/A	N/A

IF = 2500 MHz, RF = 20,000 MHz, LO = 17,500 MHz, IF power =
-10 dBm, LO power = 18 dBm, and T <sub>A</sub> = 25°C.

				N×	LO		
		0	1	2	3	4	5
	-5	81	81	75	67	N/A	N/A
	-4	81	82	75	66	N/A	N/A
	-3	81	83	75	64	N/A	N/A
	-2	82	80	64	63	N/A	N/A
	-1	17	23	22	38	N/A	N/A
M×IF	0	N/A	7	18	N/A	N/A	N/A
	+1	17	0	31	N/A	N/A	N/A
	+2	81	75	66	N/A	N/A	N/A
	+3	82	75	66	N/A	N/A	N/A
	+4	80	75	65	N/A	N/A	N/A
	+5	78	64	62	N/A	N/A	N/A

IF = 2500 MHz, RF = 30,000 MHz, LO = 27,500 MHz, IF power = -10 dBm, LO power = 18 dBm, and T<sub>A</sub> = 25°C.

				N×	LO		
		0	1	2	3	4	5
	-5	79	77	65	N/A	N/A	N/A
	-4	80	77	62	N/A	N/A	N/A
	-3	81	74	61	N/A	N/A	N/A
	-2	80	66	61	N/A	N/A	N/A
	-1	14	16	N/A	N/A	N/A	N/A
M×IF	0	N/A	N/A	N/A	N/A	N/A	N/A
	+1	14	0	N/A	N/A	N/A	N/A
	+2	79	71	N/A	N/A	N/A	N/A
	+3	77	66	N/A	N/A	N/A	N/A
	+4	81	64	N/A	N/A	N/A	N/A
	+5	79	67	N/A	N/A	N/A	N/A

IF = 2500 MHz, RF = 40,000 MHz, LO = 37,500 MHz, IF power = -10 dBm, LO power = 18 dBm, and T<sub>A</sub> = 25°C.

				Ν	× LO		
		0	1	2	3	4	5
	-5	+80	+72	N/A	N/A	N/A	N/A
	-4	+79	+71	N/A	N/A	N/A	N/A
	-3	+80	+70	N/A	N/A	N/A	N/A
	-2	+78	+65	N/A	N/A	N/A	N/A
	-1	+15	+18	N/A	N/A	N/A	N/A
M×IF	0	N/A	-4	N/A	N/A	N/A	N/A
	+1	+15	0	N/A	N/A	N/A	N/A
	+2	+80	+65	N/A	N/A	N/A	N/A
	+3	+80	+61	N/A	N/A	N/A	N/A
	+4	+76	+60	N/A	N/A	N/A	N/A
	+5	+77	+59	N/A	N/A	N/A	N/A

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### **Data Sheet**

### HMC8192LG

IF = 5000 MHz, RF = 20,000 MHz, LO = 15,000 MHz, IF power = -10 dBm, LO power = 18 dBm, and  $\rm T_{A}$  = 25°C.

				Ν	× LO		
		0	1	2	3	4	5
	-5	21	32	12	N/A	29	45
	-4	N/A	12	32	22	22	N/A
	-3	7	N/A	7	7	15	N/A
	-2	31	12	N/A	29	45	N/A
	-1	12	32	22	22	N/A	N/A
M×IF	0	N/A	7	7	14	N/A	N/A
	+1	12	0	29	45	N/A	N/A
	+2	32	22	22	N/A	N/A	N/A
	+3	7	7	14	N/A	N/A	N/A
	+4	N/A	29	45	N/A	N/A	N/A
	+5	22	22	N/A	N/A	N/A	N/A

IF = 5000 MHz, RF = 30,000 MHz, LO = 25,000 MHz, IF power = -10 dBm, LO power = 18 dBm, and T<sub>A</sub> = 25°C.

				N	× LO		
		0	1	2	3	4	5
	-5	1	N/A	1	17		
	-4	13	9	N/A	N/A	N/A	N/A
	-3	70	79	66	N/A	N/A	N/A
	-2	80	70	64	N/A	N/A	N/A
	-1	9	13	38	N/A	N/A	N/A
M×IF	0	N/A	1	18	N/A	N/A	N/A
	+1	9	0	N/A	N/A	N/A	N/A
	+2	78	64	N/A	N/A	N/A	N/A
	+3	71	67	N/A	N/A	N/A	N/A
	+4	13	36	N/A	N/A	N/A	N/A
	+5	1	18	N/A	N/A	N/A	N/A

IF = 5000 MHz, RF = 40,000 MHz, LO = 35,000 MHz, IF power = -10 dBm, LO power = 18 dBm, and T<sub>A</sub> = 25°C.

				Ν	× LO		
		0	1	2	3	4	5
	-5	+72	+79	+60	+72	N/A	N/A
	-4	+74	+77	+60	+74	N/A	N/A
	-3	+76	+71	N/A	N/A	N/A	N/A
	-2	+55	+72	N/A	N/A	N/A	N/A
	-1	+8	+15	N/A	N/A	N/A	N/A
M×IF	0	N/A	-2	N/A	N/A	N/A	N/A
	+1	+8	0	+8	N/A	N/A	N/A
	+2	+76	+58	+76	N/A	N/A	N/A
	+3	+76	+58	+76	N/A	N/A	N/A
	+4	+76	N/A	+76	N/A	N/A	N/A
	+5	+71	N/A	+71	N/A	N/A	N/A

IF = 100 MHz, RF = 20,000 MHz, LO = 20,100 MHz, IF power = -10 dBm, LO power = 18 dBm, and  $T_{\rm A}$  = 25°C.

				Ν	× LO		
		0	1	2	3	4	5
	-5	92	80	69	92	N/A	N/A
	-4	92	79	68	92	N/A	N/A
	-3	92	80	71	92	N/A	N/A
	-2	93	57	68	N/A	N/A	N/A
	-1	43	0	38	N/A	N/A	N/A
M×IF	0	N/A	6	27	N/A	N/A	N/A
	+1	43	21	31	N/A	N/A	N/A
	+2	94	71	65	N/A	N/A	N/A
	+3	94	68	69	N/A	N/A	N/A
	+4	92	79	70	N/A	N/A	N/A
	+5	91	79	70	N/A	N/A	N/A

IF = 100 MHz, RF = 30,000 MHz, LO = 30,100 MHz, IF power = -10 dBm, LO power = 18 dBm, and  $T_A = 25^{\circ}$ C.

			N × LO							
		0	1	2	3	4	5			
	-5	92	73	N/A	N/A	N/A	N/A			
	-4	92	73	N/A	N/A	N/A	N/A			
	-3	90	72	N/A	N/A	N/A	N/A			
	-2	92	61	N/A	N/A	N/A	N/A			
	-1	43	0	N/A	N/A	N/A	N/A			
M×IF	0	N/A	2	N/A	N/A	N/A	N/A			
	+1	43	19	N/A	N/A	N/A	N/A			
	+2	93	68	N/A	N/A	N/A	N/A			
	+3	93	73	N/A	N/A	N/A	N/A			
	+4	91	72	N/A	N/A	N/A	N/A			
	+5	91	74	N/A	N/A	N/A	N/A			

IF = 100 MHz, RF = 40,000 MHz, LO = 40,100 MHz, IF power = -10 dBm, LO power = 18 dBm, and T<sub>A</sub> = 25°C.

			N × LO						
		0	1	2	3	4	5		
	-5	+90	+64	N/A	N/A	N/A	N/A		
	-4	+90	+66	N/A	N/A	N/A	N/A		
	-3	+92	+68	N/A	N/A	N/A	N/A		
	-2	+91	+53	N/A	N/A	N/A	N/A		
	-1	+41	0	N/A	N/A	N/A	N/A		
M×IF	0	N/A	-8	N/A	N/A	N/A	N/A		
	+1	+41	+18	N/A	N/A	N/A	N/A		
	+2	+65	+64	N/A	N/A	N/A	N/A		
	+3	+93	+66	N/A	N/A	N/A	N/A		
	+4	+91	+67	N/A	N/A	N/A	N/A		
	+5	+90	+67	N/A	N/A	N/A	N/A		

IF = 2500 MHz, RF = 20,000 MHz, LO = 22,500 MHz, IF power = -10 dBm, LO power = 18 dBm, and  $T_{\rm A}$  = 25°C.

		N × LO						
		0	1	2	3	4	5	
	-5	82	80	70	N/A	N/A	N/A	
	-4	81	82	67	N/A	N/A	N/A	
	-3	82	79	66	N/A	N/A	N/A	
	-2	81	79	63	N/A	N/A	N/A	
	-1	17	0	40	N/A	N/A	N/A	
M×IF	0	N/A	5	21	N/A	N/A	N/A	
	+1	17	24	42	N/A	N/A	N/A	
	+2	82	74	63	N/A	N/A	N/A	
	+3	80	72	N/A	N/A	N/A	N/A	
	+4	81	73	N/A	N/A	N/A	N/A	
	+5	79	68	N/A	N/A	N/A	N/A	

IF = 2500 MHz, RF = 30,000 MHz, LO = 32,500 MHz, IF power = -10 dBm, LO power = 18 dBm, and  $T_{\rm A}$  = 25°C.

			N × LO						
		0	1	2	3	4	5		
	-5	+81	+77	N/A	N/A	N/A	N/A		
	-4	+79	+74	N/A	N/A	N/A	N/A		
	-3	+79	+72	N/A	N/A	N/A	N/A		
	-2	+79	+62	N/A	N/A	N/A	N/A		
	-1	+15	0	N/A	N/A	N/A	N/A		
M×IF	0	N/A	-4	N/A	N/A	N/A	N/A		
	+1	+15	+19	N/A	N/A	N/A	N/A		
	+2	+78	+62	N/A	N/A	N/A	N/A		
	+3	+75	+66	N/A	N/A	N/A	N/A		
	+4	+78	+64	N/A	N/A	N/A	N/A		
	+5	+80	+64	N/A	N/A	N/A	N/A		

IF = 5000 MHz, RF = 20,000 MHz, LO = 25,000 MHz, IF power =
-10 dBm, LO power = 18 dBm, and T <sub>A</sub> = 25°C.

		N × LO						
		0	1	2	3	4	5	
	-5	1	N/A	1	18			
	-4	N/A	12	19	N/A	N/A	N/A	
	-3	71	76	68	N/A	N/A	N/A	
	-2	80	70	66	N/A	N/A	N/A	
	-1	12	0	42	N/A	N/A	N/A	
M×IF	0	N/A	1	18	N/A	N/A	N/A	
	+1	12	18	N/A	N/A	N/A	N/A	
	+2	73	67	N/A	N/A	N/A	N/A	
	+3	71	66	N/A	N/A	N/A	N/A	
	+4	N/A	42	N/A	N/A	N/A	N/A	
	+5	1	19	N/A	N/A	N/A	N/A	

IF = 5000 MHz, RF = 30,000 MHz, LO = 35,000 MHz, IF power = -10 dBm, LO power = 18 dBm, and T<sub>A</sub> = 25°C.

			N × LO						
		0	1	2	3	4	5		
	-5	+68	+78	+60	N/A	N/A	N/A		
	-4	+74	+77	+59	N/A	N/A	N/A		
	-3	+77	+74	N/A	N/A	N/A	N/A		
	-2	+74	+70	N/A	N/A	N/A	N/A		
	-1	+9	0	N/A	N/A	N/A	N/A		
M×IF	0	N/A	-6	N/A	N/A	N/A	N/A		
	+1	+9	+16	N/A	N/A	N/A	N/A		
	+2	+53	+58	N/A	N/A	N/A	N/A		
	+3	+75	+59	N/A	N/A	N/A	N/A		
	+4	+73	N/A	N/A	N/A	N/A	N/A		
	+5	+70	N/A	N/A	N/A	N/A	N/A		

### **THEORY OF OPERATION**

The HMC8192LG is a passive, wideband, I/Q MMIC mixer that can be used either as an image rejection mixer for receiver operations or as a single-sideband upconverter for transmitter operations. With an RF and LO range of 20 GHz to 42 GHz and an IF bandwidth of dc to 5 GHz, the HMC8192LG is ideal for applications requiring a wide frequency range, excellent RF performance, and a simple design with fewer components and a small PCB footprint. A single HMC8192LG can replace multiple narrow-band mixers in a design.

The inherent I/Q architecture of the HMC8192LG offers excellent image rejection, eliminating the need for expensive filtering for unwanted sidebands. The double balanced architecture of the mixer also provides excellent LO to RF isolation and LO to IF isolation, and reduces the effect of LO leakage to ensure signal integrity. Because the HMC8192LG is a passive mixer, the HMC8192LG does not require any dc power sources. The HMC8192LG offers a lower noise figure compared to an active mixer, ensuring superior dynamic range for high performance and precision applications.

The HMC8192LG is fabricated on a GaAs MESFET process and uses Analog Devices mixer cells and a 90° hybrid. The HMC8192LG is available in a compact, 4.00 mm × 4.00 mm, 25-terminal LGA\_CAV package and operates over a  $-40^{\circ}$ C to +85°C temperature range. The evaluation board for the HMC8192LG, EV1HMC8192LG, is also available on the Analog Devices website.

For both upconversion and downconversion, an external 90° hybrid is required. See the Applications Information section for details on interfacing with an external 90° hybrid.

### **APPLICATIONS INFORMATION**

Figure 121 shows the typical application circuit for the HMC8192LG. To select the appropriate sideband, an external 90° hybrid is needed. For applications not requiring operation to dc, use an off-chip dc blocking capacitor. For applications that require the LO signal at the output to be suppressed, use a bias tee or RF feed as shown in Figure 121. Ensure that the source or sink current used for LO suppression is greater than 3 mA for each IFx port to prevent damage to the device. The common-mode voltage for each IFx port is 0 V.

To select the upper sideband when using the HMC8192LG as an upconverter, connect the IF1 pin to the 90° port of the hybrid and connect the IF2 pin to the 0° port of the hybrid. To select the lower sideband, connect the IF1 pin to the 0° port of the hybrid and the IF2 pin to the 90° port of the hybrid. The input is from the sum port of the hybrid, and the difference port is 50  $\Omega$  terminated.

To select the upper sideband (low-side LO) when using the HMC8192LG as a downconverter, connect the IF1 pin to the 0° port of the hybrid and connect the IF2 pin to the 90° port of the hybrid. To select the lower sideband (high-side LO), connect the IF1 pin to the 90° port of the hybrid and the IF2 pin to the 0° port of the hybrid. The output is from the sum port of the hybrid, and the difference port is 50  $\Omega$  terminated.



### LAYOUT

Solder the exposed pad on the underside of the HMC8192LG to a low thermal and electrical impedance ground plane. This pad is typically soldered to an exposed opening in the solder mask on the evaluation board. Connect these ground vias to all other ground layers on the evaluation board to maximize heat dissipation from the device package.

### **EVALUATION BOARD INFORMATION**

The EV1HMC8192LG evaluation board PCB used in the application must use RF circuit design techniques. Signal lines must have a 50  $\Omega$  impedance and connect the package ground leads and exposed pad directly to the ground plane, similar to the setup shown in Figure 123. Use a sufficient number of via holes to connect the top and bottom ground planes.



Figure 122. Evaluation Board Layout for the HMC8192LG



Figure 123. EV1HMC8192LG Evaluation Board PCB, Top Layer

5697-121

L'valuation Doard I CD							
Quantity	Reference Designator	Description	Manufacturer				
1	Not applicable	PCB, EV1HMC8192LG <sup>2</sup>	Analog Devices				
2	J1, J2	PCB connector, end launch SMA edge mount	Southwest Microwave				
2	J3, J4	PCB connector, jack assembly, end launch, SMA	CINCH				
1	U1	Device under test, HMC8192LG	Analog Devices				

# Table 6. Bill of Materials for the EV1HMC8192LG1Evaluation Board PCB

<sup>1</sup> Reference this number when ordering the evaluation board PCB. <sup>2</sup> Circuit board material: RO4350B™ laminates.

### PERFORMANCE AT LOWER IF FREQUENCIES

The HMC8192LG can operate at low IF frequencies approaching dc. Figure 124 shows the conversion gain and image rejection performance at lower IF frequencies for upconversion. Figure 125 shows the conversion gain and image rejection performance at lower IF frequencies for downconversion. This performance is typical and is not guaranteed.



Figure 124. IF Bandwidth as Upconverter at Low IF Frequencies, LO = 25 GHz, 18 dBm, Upper Sideband



Figure 125. IF Bandwidth as Downconverter at Low IF Frequencies, LO = 25 GHz, 18 dBm, Upper Sideband

#### PERFORMANCE AT HIGHER IF FREQUENCIES

Figure 126 and Figure 127 show the IF performance above 5 GHz. The data for Figure 126 was taken as an upconverter configuration with a lower sideband at an LO frequency of 31 GHz and an LO power of 18 dBm. The data for Figure 127 was taken as a down-converter configuration with a lower sideband at an LO frequency of 31 GHz and an LO power of 18 dBm. This performance is typical and is not guaranteed.



Figure 126. IF Bandwidth at IF Frequencies Above 5 GHz, Data Taken as Upconverter, Lower Sideband, LO = 31 GHz at 18 dBm,  $T_A = 25$  °C



Figure 127. IF Bandwidth at IF Frequencies Above 5 GHz, Data Taken as Downconverter, Lower Sideband (High-Side LO), LO = 31 GHz at 18 dBm,  $T_A = 25^{\circ}$ C

### **OUTLINE DIMENSIONS**



Figure 128. 25-Terminal Chip Array Small Outline No Lead Cavity [LGA\_CAV] 4.00 mm × 4.00 mm Body and 1.55 mm Package Height (CE-25-1) Dimensions shown in millimeters

#### **ORDERING GUIDE**

Model <sup>1</sup>	Temperature Range	Package Description	MSL Rating <sup>2</sup>	Package Option
HMC8192LG	–40°C to +85°C	25-Terminal LGA_CAV	MSL3	CE-25-1
HMC8192LGTR	–40°C to +85°C	25-Terminal LGA_CAV	MSL3	CE-25-1
EV1HMC8192LG		Evaluation PCB Assembly		

<sup>1</sup> The HMC8192LG and the HMC8192LGTR are RoHS compliant parts.

<sup>2</sup> See the Absolute Maximum Ratings section.

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