

RF LDMOS Wideband Integrated Power Amplifiers

The MD7IC2050N wideband integrated circuit is designed with on-chip matching that makes it usable from 1750- 2050 MHz. This multi-stage structure is rated for 26 to 32 Volt operation and covers all typical cellular base station modulation formats.

- Typical Doherty Single-Carrier W-CDMA Performance: $V_{DD} = 28$ Volts, $I_{DQ1A} = I_{DQ1B} = 30$ mA, $I_{DQ2A} = 230$ mA, $V_{GS2B} = 1.4$ Vdc, $P_{out} = 10$ Watts Avg., IQ Magnitude Clipping, Channel Bandwidth = 3.84 MHz, Input Signal PAR = 9.9 dB @ 0.01% Probability on CCDF.

Frequency	G_{ps} (dB)	PAE (%)	Output PAR (dB)	ACPR (dBc)
2025 MHz	30.5	34.7	8.7	-37.4

- Capable of Handling 5:1 VSWR, @ 32 Vdc, 2017.5 MHz, 79 Watts CW Output Power (3 dB Input Overdrive from Rated P_{out})
- Stable into a 5:1 VSWR. All Spurs Below -60 dBc @ 20 Watts to 80 Watts CW P_{out}
- Typical P_{out} @ 3 dB Compression Point = 74 Watts CW

1880 MHz

- Typical Doherty Single-Carrier W-CDMA Performance: $V_{DD} = 28$ Volts, $I_{DQ1A} = I_{DQ1B} = 30$ mA, $I_{DQ2A} = 230$ mA, $V_{GS2B} = 1.4$ Vdc, $P_{out} = 10$ Watts Avg., IQ Magnitude Clipping, Channel Bandwidth = 3.84 MHz, Input Signal PAR = 9.9 dB @ 0.01% Probability on CCDF.

Frequency	G_{ps} (dB)	PAE (%)	Output PAR (dB)	ACPR (dBc)
1880 MHz	30.3	35.2	8.6	-34.9
1900 MHz	30.2	34.9	8.6	-36.3
1920 MHz	30.1	34.8	8.7	-36.9

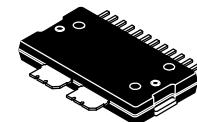
Features

- 100% PAR Tested for Guaranteed Output Power Capability
- Production Tested in a Symmetrical Doherty Configuration
- Characterized with Series Equivalent Large-Signal Impedance Parameters and Common Source S-Parameters
- On-Chip Matching (50 Ohm Input, DC Blocked)
- Integrated Quiescent Current Temperature Compensation with Enable/Disable Function (1)
- Integrated ESD Protection
- 225°C Capable Plastic Package
- RoHS Compliant
- In Tape and Reel. R1 Suffix = 500 Units per 44 mm, 13 inch Reel

MD7IC2050NR1
MD7IC2050GNR1
MD7IC2050NBR1

1880-2100 MHz, 10 W AVG., 28 V SINGLE W-CDMA RF LDMOS WIDEBAND INTEGRATED POWER AMPLIFIERS

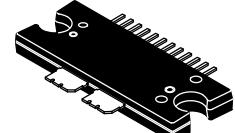
CASE 1618-02
TO-270 WB-14
PLASTIC
MD7IC2050NR1



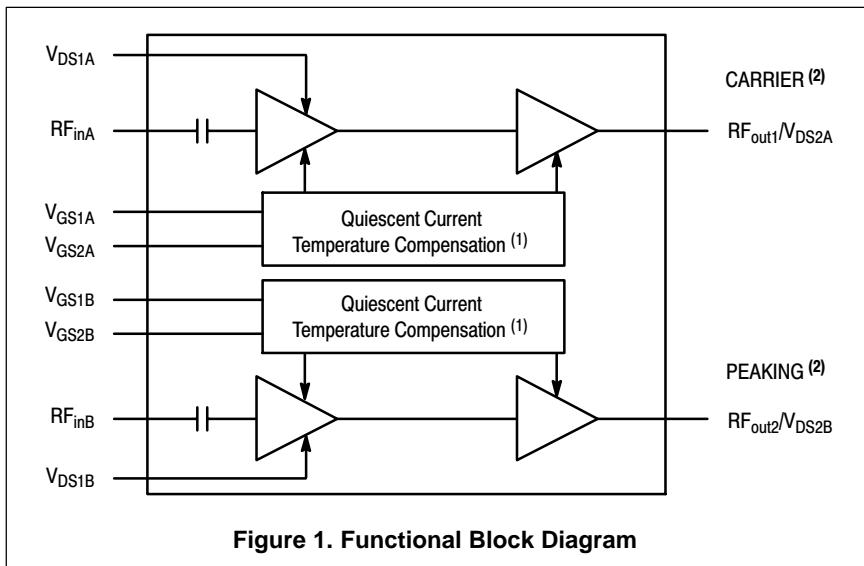
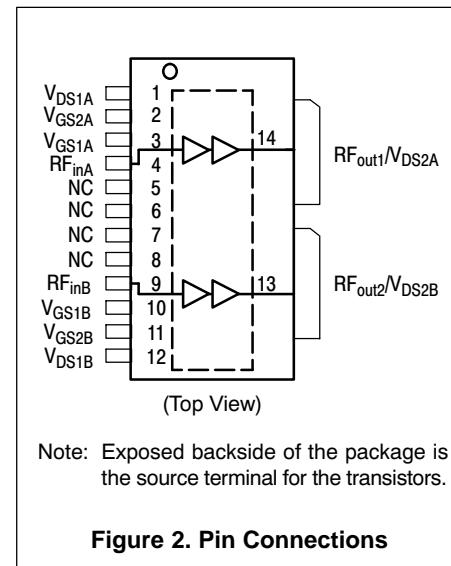
CASE 1621-02
TO-270 WB-14 GULL PLASTIC
MD7IC2050GNR1



CASE 1617-02
TO-272 WB-14
PLASTIC
MD7IC2050NBR1



1. Refer to AN1977, *Quiescent Current Thermal Tracking Circuit in the RF Integrated Circuit Family* and to AN1987, *Quiescent Current Control for the RF Integrated Circuit Device Family*. Go to <http://www.freescale.com/rf>. Select Documentation/Application Notes - AN1977 or AN1987.

**Figure 1. Functional Block Diagram**

Note: Exposed backside of the package is the source terminal for the transistors.

Figure 2. Pin Connections

- Refer to AN1977, *Quiescent Current Thermal Tracking Circuit in the RF Integrated Circuit Family* and to AN1987, *Quiescent Current Control for the RF Integrated Circuit Device Family*. Go to <http://www.freescale.com/rf>. Select Documentation/Application Notes - AN1977 or AN1987.
- Peaking and Carrier orientation is determined by the test fixture design.

Table 1. Maximum Ratings

Rating	Symbol	Value	Unit
Drain-Source Voltage	V _{DSS}	-0.5, +65	Vdc
Gate-Source Voltage	V _{GS}	-0.5, +10	Vdc
Operating Voltage	V _{DD}	32, +0	Vdc
Storage Temperature Range	T _{stg}	-65 to +150	°C
Case Operating Temperature	T _C	150	°C
Operating Junction Temperature (1,2)	T _J	225	°C
Input Power	P _{in}	28	dBm

Table 2. Thermal Characteristics

Characteristic	Symbol	Value (2,3)	Unit
Final Doherty Application			
Thermal Resistance, Junction to Case Case Temperature 81°C, P _{out} = 50 W CW Stage 1A, 28 Vdc, I _{DQ1A} = 30 mA Stage 1B, 28 Vdc, I _{DQ1B} = 30 mA Stage 2A, 28 Vdc, I _{DQ2A} = 230 mA Stage 2B, 28 Vdc, V _{GS2B} = 1.4 Vdc	R _{θJC}	8.2 8.2 1.8 1.8	°C/W
Case Temperature 73°C, P _{out} = 10 W CW Stage 1A, 28 Vdc, I _{DQ1A} = 30 mA Stage 1B, 28 Vdc, I _{DQ1B} = 30 mA Stage 2A, 28 Vdc, I _{DQ2A} = 230 mA Stage 2B, 28 Vdc, V _{GS2B} = 1.4 Vdc		8.3 8.3 1.9 1.9	

- Continuous use at maximum temperature will affect MTTF.
- MTTF calculator available at <http://www.freescale.com/rf>. Select Software & Tools/Development Tools/Calculators to access MTTF calculators by product.
- Refer to AN1955, *Thermal Measurement Methodology of RF Power Amplifiers*. Go to <http://www.freescale.com/rf>. Select Documentation/Application Notes - AN1955.

Table 3. ESD Protection Characteristics

Test Methodology	Class
Human Body Model (per JESD22-A114)	0 (Minimum)
Machine Model (per EIA/JESD22-A115)	A (Minimum)
Charge Device Model (per JESD22-C101)	II (Minimum)

Table 4. Moisture Sensitivity Level

Test Methodology	Rating	Package Peak Temperature	Unit
Per JESD22-A113, IPC/JEDEC J-STD-020	3	260	°C

Table 5. Electrical Characteristics ($T_A = 25^\circ\text{C}$ unless otherwise noted)

Characteristic	Symbol	Min	Typ	Max	Unit
Stage 1 - Off Characteristics (1)					
Zero Gate Voltage Drain Leakage Current ($V_{DS} = 65 \text{ Vdc}$, $V_{GS} = 0 \text{ Vdc}$)	I_{DSS}	—	—	10	μAdc
Zero Gate Voltage Drain Leakage Current ($V_{DS} = 28 \text{ Vdc}$, $V_{GS} = 0 \text{ Vdc}$)	I_{DSS}	—	—	1	μAdc
Gate-Source Leakage Current ($V_{GS} = 1.5 \text{ Vdc}$, $V_{DS} = 0 \text{ Vdc}$)	I_{GSS}	—	—	1	μAdc
Stage 1 - On Characteristics (1)					
Gate Threshold Voltage ($V_{DS} = 10 \text{ Vdc}$, $I_D = 23 \mu\text{Adc}$)	$V_{GS(\text{th})}$	1.2	1.9	2.7	Vdc
Gate Quiescent Voltage ($V_{DS} = 28 \text{ Vdc}$, $I_{DQ1A} = I_{DQ1B} = 30 \text{ mA}$)	$V_{GS(Q)}$	—	3	—	Vdc
Fixture Gate Quiescent Voltage ($V_{DD} = 28 \text{ Vdc}$, $I_{DQ1A} = I_{DQ1B} = 30 \text{ mA}$, Measured in Functional Test)	$V_{GG(Q)}$	4.1	5.5	7.1	Vdc
Stage 2 - Off Characteristics (1)					
Zero Gate Voltage Drain Leakage Current ($V_{DS} = 65 \text{ Vdc}$, $V_{GS} = 0 \text{ Vdc}$)	I_{DSS}	—	—	10	μAdc
Zero Gate Voltage Drain Leakage Current ($V_{DS} = 28 \text{ Vdc}$, $V_{GS} = 0 \text{ Vdc}$)	I_{DSS}	—	—	1	μAdc
Gate-Source Leakage Current ($V_{GS} = 1.5 \text{ Vdc}$, $V_{DS} = 0 \text{ Vdc}$)	I_{GSS}	—	—	1	μAdc
Stage 2 - On Characteristics (1)					
Gate Threshold Voltage ($V_{DS} = 10 \text{ Vdc}$, $I_D = 150 \mu\text{Adc}$)	$V_{GS(\text{th})}$	1.2	2	2.7	Vdc
Gate Quiescent Voltage ($V_{DS} = 28 \text{ Vdc}$, $I_{DQ2A} = 230 \text{ mA}$)	$V_{GS(Q)}$	—	3	—	Vdc
Fixture Gate Quiescent Voltage ($V_{DD} = 28 \text{ Vdc}$, $I_{DQ2A} = 230 \text{ mA}$, Measured in Functional Test)	$V_{GG(Q)}$	4.1	5.5	7.1	Vdc
Drain-Source On-Voltage ($V_{GS} = 10 \text{ Vdc}$, $I_D = 1 \text{ Adc}$)	$V_{DS(\text{on})}$	0.1	0.3	1.2	Vdc

Functional Tests (2,3,4) (In Freescale Doherty Test Fixture, 50 ohm system) $V_{DD} = 28 \text{ Vdc}$, $I_{DQ1A} = I_{DQ1B} = 30 \text{ mA}$, $I_{DQ2A} = 230 \text{ mA}$, $V_{GS2B} = 1.4 \text{ Vdc}$, $P_{out} = 10 \text{ W Avg.}$, $f = 2025 \text{ MHz}$, Single-Carrier W-CDMA, IQ Magnitude Clipping, Input Signal PAR = 9.9 dB @ 0.01% Probability on CCDF. ACPR measured in 3.84 MHz Channel Bandwidth @ $\pm 5 \text{ MHz}$ Offset.

Power Gain	G_{ps}	28.5	30.5	33.0	dB
Power Added Efficiency	PAE	32.0	34.7	—	%
Output Peak-to-Average Ratio @ 0.01% Probability on CCDF	PAR	8.0	8.7	—	dB
Adjacent Channel Power Ratio	ACPR	—	-37.4	-34.0	dBc

1. Each side of device measured separately.
2. Part internally matched both on input and output.
3. Measurement made with device in a Symmetrical Doherty configuration.
4. Measurement made with device in straight lead configuration before any lead forming operation is applied.

(continued)

MD7IC2050NR1 MD7IC2050GNR1 MD7IC2050NBR1

Table 5. Electrical Characteristics ($T_A = 25^\circ\text{C}$ unless otherwise noted) **(continued)**

Characteristic	Symbol	Min	Typ	Max	Unit
Typical Performances (1) (In Freescale Doherty Test Fixture, 50 ohm system) $V_{DD} = 28 \text{ Vdc}$, $I_{DQ1A} = I_{DQ1B} = 30 \text{ mA}$, $I_{DQ2A} = 230 \text{ mA}$, $V_{GS2B} = 1.4 \text{ Vdc}$, 2010-2025 MHz Bandwidth					
P_{out} @ 1 dB Compression Point, CW	$P_{1\text{dB}}$	—	60	—	W
P_{out} @ 3 dB Compression Point, CW	$P_{3\text{dB}}$	—	74	—	W
IMD Symmetry @ 30 W PEP, P_{out} where IMD Third Order Intermodulation $\equiv 30 \text{ dBc}$ (Delta IMD Third Order Intermodulation between Upper and Lower Sidebands > 2 dB)	IMD_{sym}	—	55	—	MHz
VBW Resonance Point (IMD Third Order Intermodulation Inflection Point)	VBW_{res}	—	70	—	MHz
Quiescent Current Accuracy over Temperature with 4.7 k Ω Gate Feed Resistors (-30 to 85°C) (2)	ΔI_{QT}	—	2.64	—	%
Gain Flatness in 15 MHz Bandwidth @ $P_{out} = 10 \text{ W Avg.}$	G_F	—	0.1	—	dB
Gain Variation over Temperature (-30 °C to +85°C)	ΔG	—	0.033	—	dB/°C
Output Power Variation over Temperature (-30 °C to +85°C)	$\Delta P_{1\text{dB}}$	—	0.008	—	dBm/°C

Typical W-CDMA Broadband Performance — 1880 MHz (In Freescale Test Fixture, 50 ohm system) $V_{DD} = 28 \text{ Vdc}$, $I_{DQ1A} = I_{DQ1B} = 30 \text{ mA}$, $I_{DQ2A} = 230 \text{ mA}$, $V_{GS2B} = 1.4 \text{ Vdc}$, $P_{out} = 10 \text{ W Avg.}$, Single-Carrier W-CDMA, IQ Magnitude Clipping, Input Signal PAR = 9.9 dB @ 0.01% Probability on CCDF. ACPR measured in 3.84 MHz Channel Bandwidth @ $\pm 5 \text{ MHz}$ Offset.

Frequency	G_{ps} (dB)	PAE (%)	Output PAR (dB)	ACPR (dBc)	IRL (dB)
1880 MHz	30.3	35.2	8.6	-34.9	-21
1900 MHz	30.2	34.9	8.6	-36.3	-21
1920 MHz	30.1	34.8	8.7	-36.9	-22

1. Measurement made with device in a Symmetrical Doherty configuration.
2. Refer to AN1977, *Quiescent Current Thermal Tracking Circuit in the RF Integrated Circuit Family* and to AN1987, *Quiescent Current Control for the RF Integrated Circuit Device Family*. Go to <http://www.freescale.com/rf>. Select Documentation/Application Notes - AN1977 or AN1987.

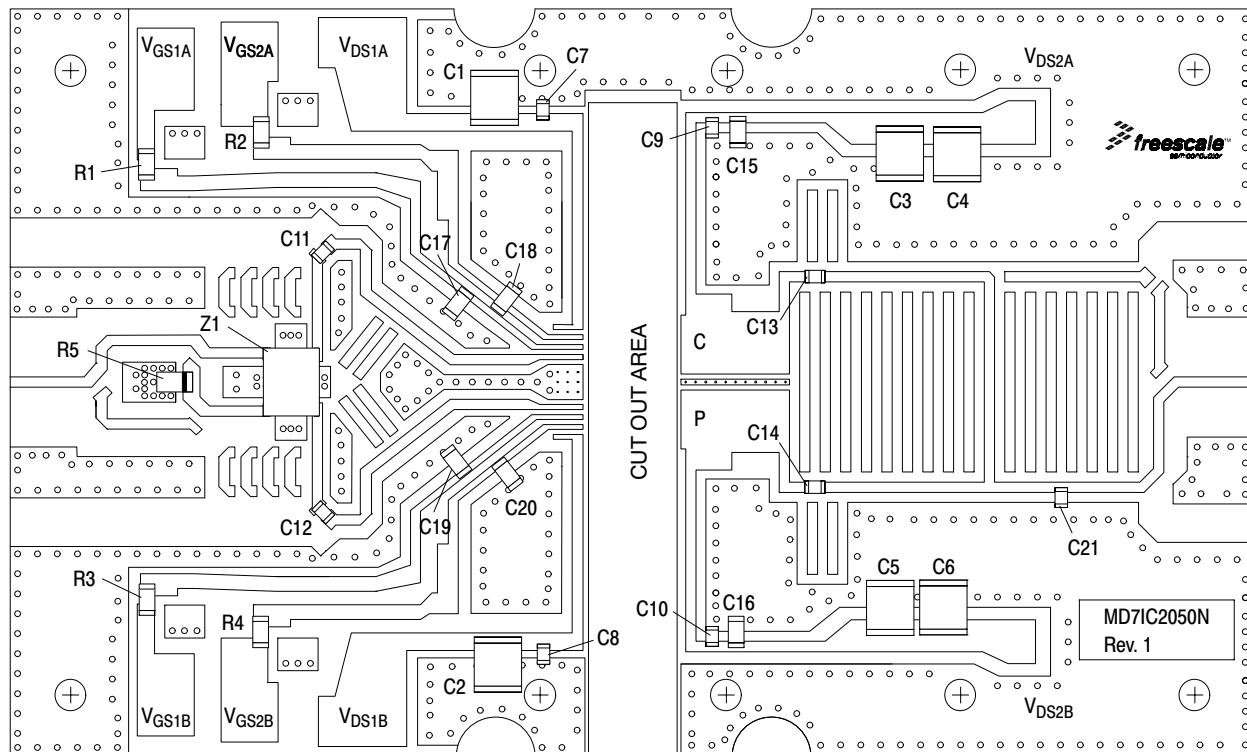


Figure 3. MD7IC2050NR1(GNR1)(NBR1) Test Circuit Component Layout

Table 6. MD7IC2050NR1(GNR1)(NBR1) Test Circuit Component Designations and Values

Part	Description	Part Number	Manufacturer
C1, C2, C3, C4, C5, C6	10 μ F, 50 V Chip Capacitors	GRM55DR61H106KA88L	Murata
C7, C8	4.7 pF Chip Capacitors	ATC600F4R7BT250XT	ATC
C9, C10	5.6 pF Chip Capacitors	ATC600F5R6BT250XT	ATC
C11, C12, C13, C14	39 pF Chip Capacitors	ATC600F390JT250XT	ATC
C15, C16, C17, C18, C19, C20	4.7 μ F, 50 V Chip Capacitors	GRM31CR71H475KA12L	Murata
C21	1.0 pF Chip Capacitor	ATC600F1R0BT250XT	ATC
R1, R2, R3, R4	4.7 k Ω , 1/4 W Chip Resistors	CRCW12064701KEA	Vishay
R5	50 Ω , 1/4 W Thick Film Chip Resistor	RK73B2BTTD510J	KOA Speer
Z1	1900 MHz Band 90°, 3 dB Hybrid Coupler	GSC351-HYB1900	Soshin
PCB	0.020", $\epsilon_r = 3.5$	RF-35	Taconic

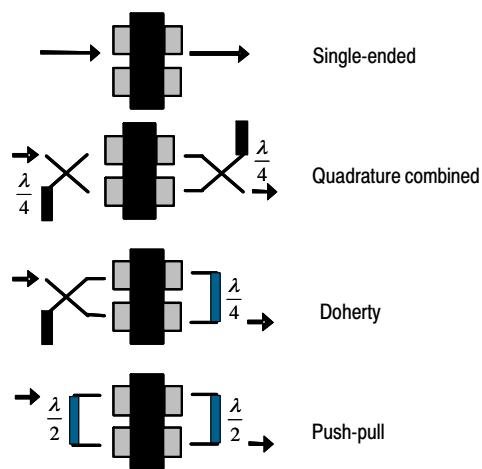


Figure 4. Possible Circuit Topologies

TYPICAL CHARACTERISTICS

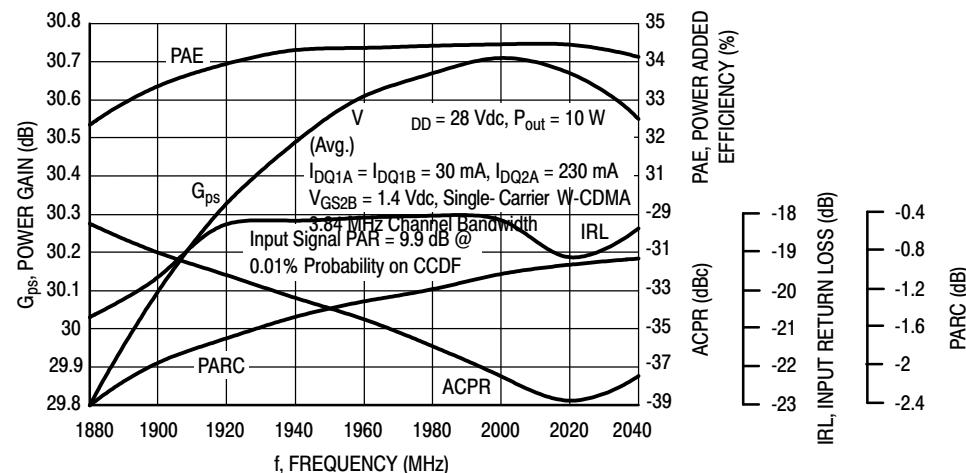


Figure 5. Output Peak-to-Average Ratio Compression (PARC)
Broadband Performance @ $P_{out} = 10$ Watts Avg.

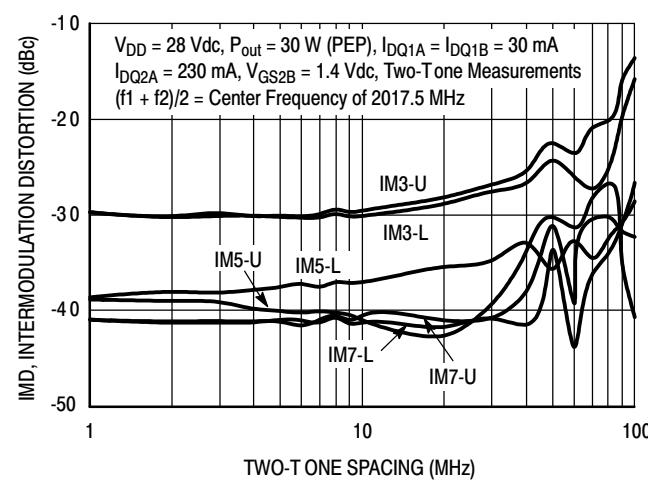


Figure 6. Intermodulation Distortion Products
versus Two-Tone Spacing

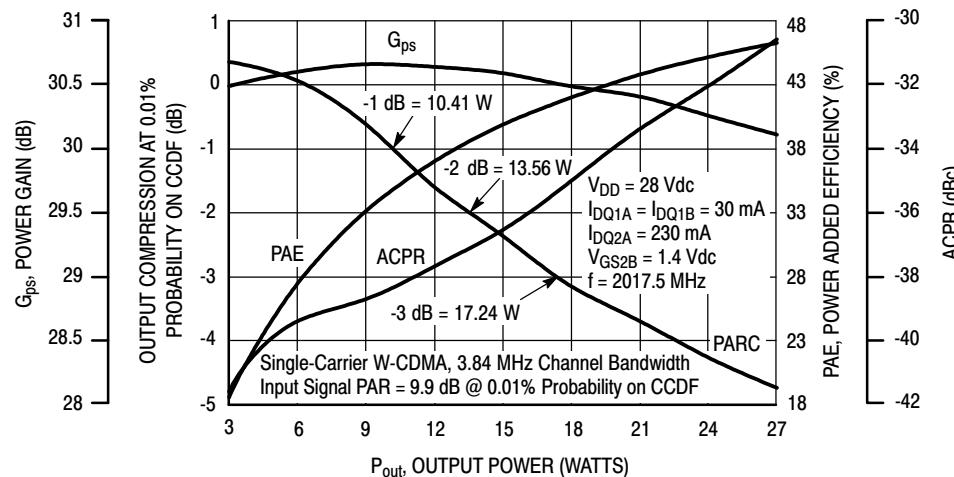
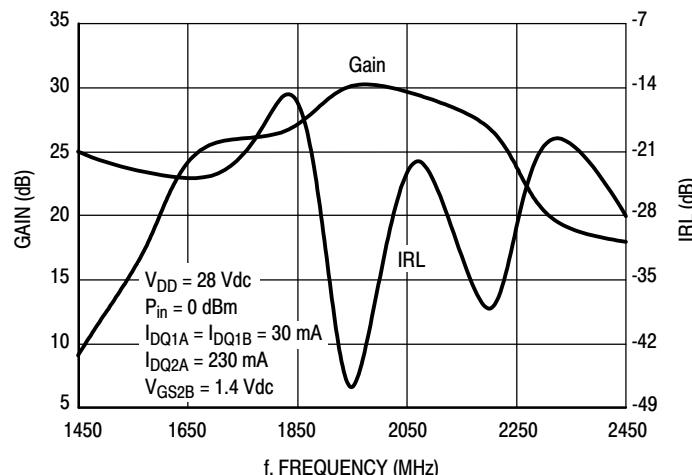
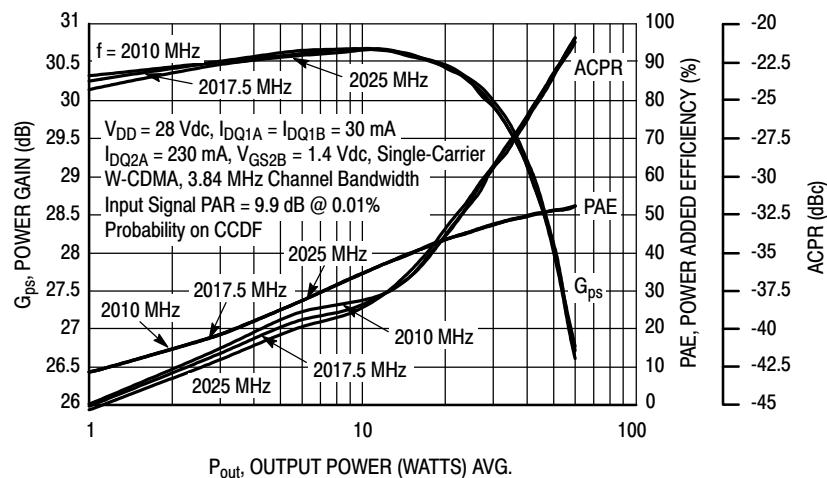


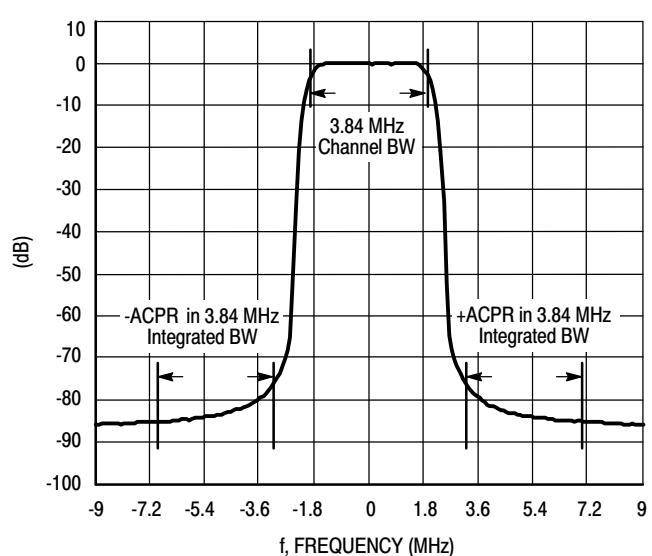
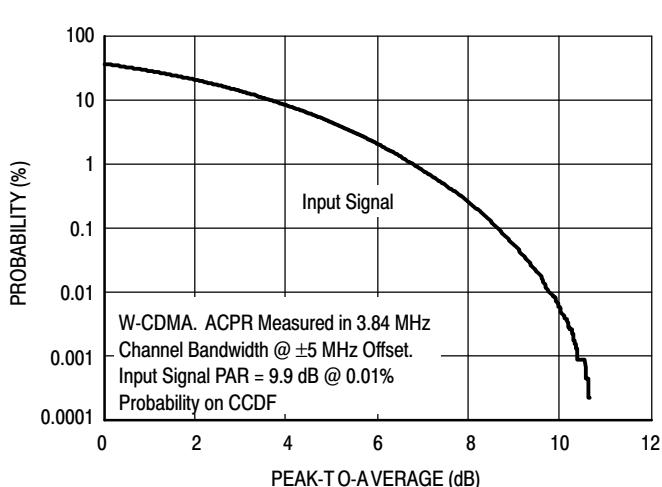
Figure 7. Output Peak-to-Average Ratio
Compression (PARC) versus Output Power

MD7IC2050NR1 MD7IC2050GNR1 MD7IC2050NBR1

TYPICAL CHARACTERISTICS



W-CDMA TEST SIGNAL



$V_{DD} = 28 \text{ Vdc}$, $I_{DQ1A} = I_{DQB} = 30 \text{ mA}$, $I_{DQ2A} = 230 \text{ mA}$, $V_{GS2B} = 1.4 \text{ Vdc}$, $P_{out} = 10 \text{ W Avg.}$

f MHz	Z_{in} Ω	Z_{load} Ω
1995	60.12 - j33.28	11.79 - j6.72
2000	59.30 - j32.57	11.78 - j6.78
2005	58.41 - j32.06	11.78 - j6.85
2010	57.41 - j31.31	11.78 - j6.92
2015	56.31 - j30.27	11.79 - j7.00
2020	55.94 - j29.62	11.81 - j7.08
2025	55.28 - j28.90	11.81 - j7.16
2030	54.75 - j28.12	11.84 - j7.24
2035	54.39 - j27.55	11.80 - j7.33

Note: Measured with Peaking side open.

Z_{in} = Device input impedance as measured from gate to ground.

Z_{load} = Test circuit impedance as measured from drain to ground.

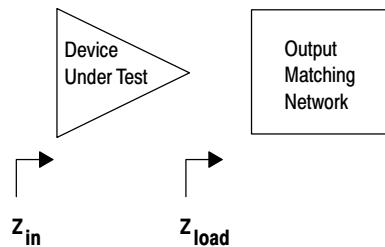


Figure 12. Series Equivalent Input and Load Impedance — Carrier Side

$V_{DD} = 28 \text{ Vdc}$, $I_{DQ1A} = I_{DQB} = 30 \text{ mA}$, $I_{DQ2A} = 230 \text{ mA}$, $V_{GS2B} = 1.4 \text{ Vdc}$, $P_{out} = 10 \text{ W Avg.}$

f MHz	Z_{in} Ω	Z_{load} Ω
1995	60.12 - j33.28	1.86 - j11.38
2000	59.30 - j32.57	1.80 - j11.24
2005	58.41 - j32.06	1.71 - j11.12
2010	57.41 - j31.31	1.64 - j11.00
2015	56.31 - j30.27	1.58 - j10.91
2020	55.94 - j29.62	1.51 - j10.78
2025	55.28 - j28.90	1.45 - j10.66
2030	54.75 - j28.12	1.38 - j10.56
2035	54.39 - j27.55	1.33 - j10.40

Note: Measured with Carrier side open.

Z_{in} = Device input impedance as measured from gate to ground.

Z_{load} = Test circuit impedance as measured from drain to ground.

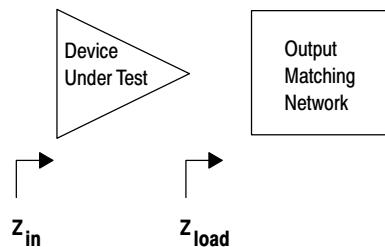
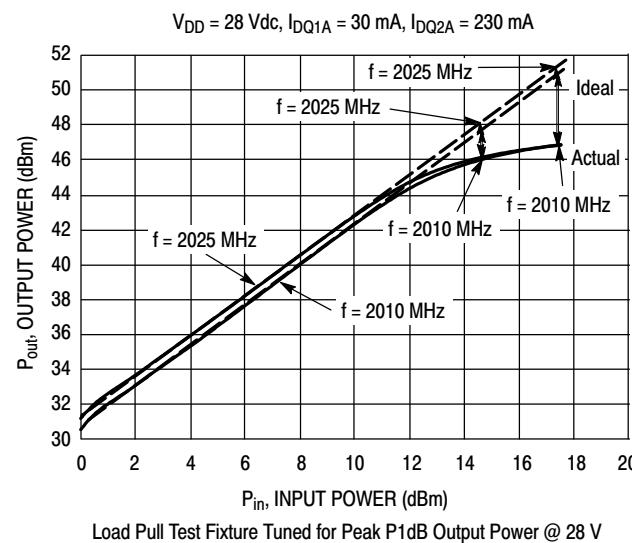


Figure 13. Series Equivalent Input and Load Impedance — Peaking Side

MD7IC2050NR1 MD7IC2050GNR1 MD7IC2050NBR1

ALTERNATIVE PEAK TUNE LOAD PULL CHARACTERISTICS



f (MHz)	P1dB		P3dB	
	Watts	dBm	Watts	dBm
2010	40	46	49	46.9
2025	38.9	45.9	47.9	46.8

Test Impedances per Compression Level

f (MHz)		Z_{source} Ω	Z_{load} Ω
2010	P1dB	$73.6 + j31.1$	$6.8 - j13.7$
2025	P1dB	$68.9 + j26.7$	$8.3 - j14.3$

Figure 14. CW Output Power versus Input Power @ 28 V

NOTE: Measurement made on the Class AB, carrier side of the device.

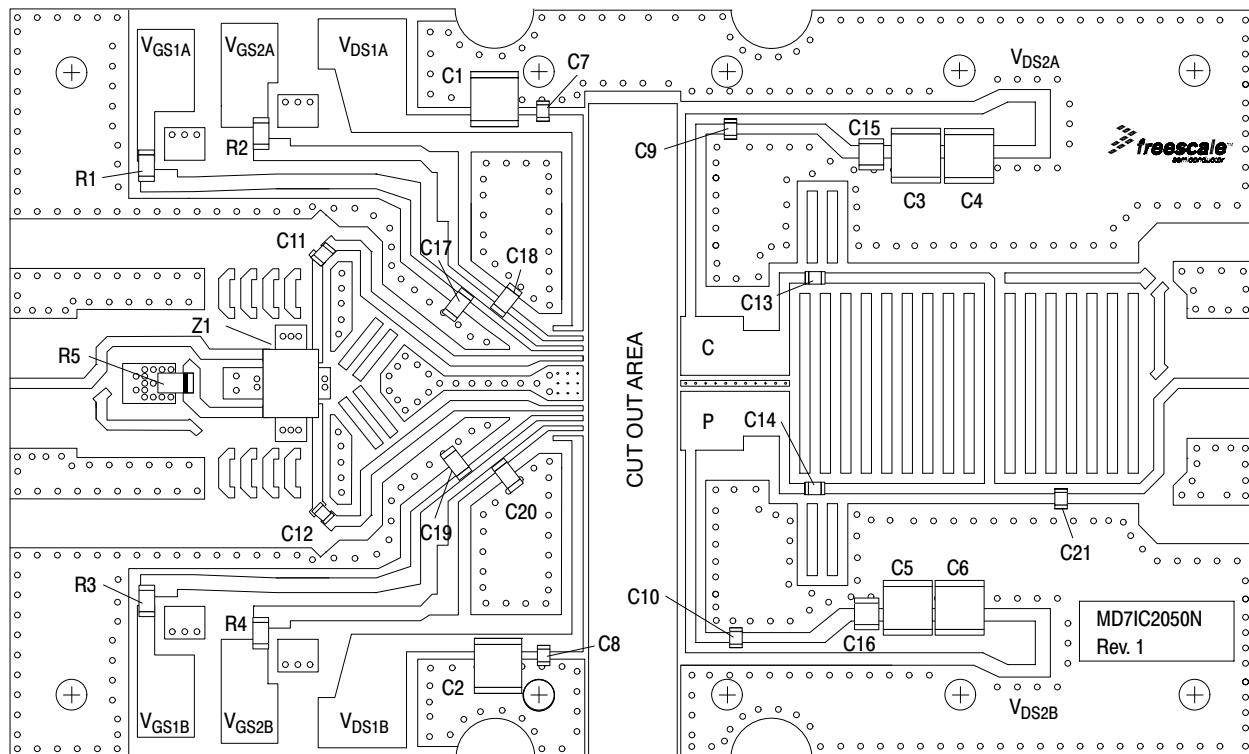


Figure 15. MD7IC2050NR1(GNR1)(NBR1) Test Circuit Component Layout — 1880 MHz

Table 7. MD7IC2050NR1(GNR1)(NBR1) Test Circuit Component Designations and Values — 1880 MHz

Part	Description	Part Number	Manufacturer
C1, C2, C3, C4, C5, C6	10 μF , 50 V Chip Capacitors	GRM55DR61H106KA88L	Murata
C7, C8	6.8 pF Chip Capacitors	ATC600F6R8BT250XT	ATC
C9, C10	15 pF Chip Capacitors	ATC600F150JT250XT	ATC
C11, C12, C13, C14	33 pF Chip Capacitors	ATC600F330JT250XT	ATC
C15, C16	6.8 μF , 50 V Chip Capacitors	GRM32CF51H685ZA01L	Murata
C17, C18, C19, C20	2.2 μF , Chip Capacitors	GRM31CR61H225KA88L	Murata
C21	0.9 pF Chip Capacitor	ATC600F0R9BT250XT	ATC
R1, R2, R3, R4	4.7 k Ω , 1/4 W Chip Resistors	CRCW12064701FKEA	Vishay
R5	50 Ω , 1/4 W Thick Film Chip Resistor	RK73B2BTBD510J	KOA Speer
Z1	1900 MHz Band 90°, 3 dB Hybrid Coupler	GSC351-HYB1900	Soshin
PCB	0.020", $\epsilon_r = 3.5$	RO4350B	Rogers

TYPICAL CHARACTERISTICS

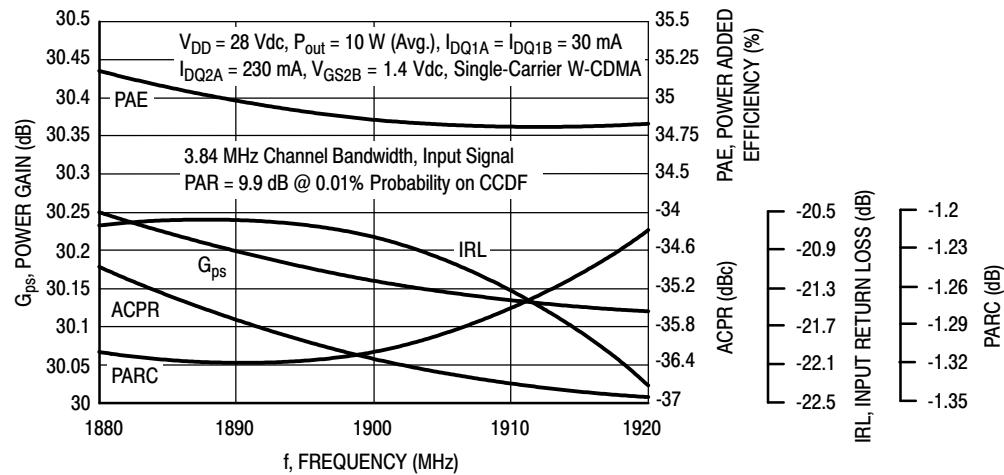


Figure 16. Output Peak-to-Average Ratio Compression (PARC)
Broadband Performance @ $P_{out} = 10$ Watts Avg. — 1880 MHz

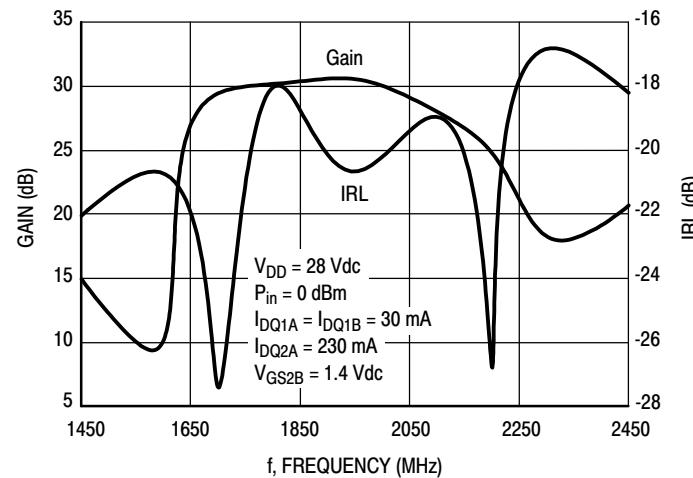


Figure 17. Broadband Frequency Response — 1880 MHz

$V_{DD} = 28$ Vdc, $I_{DQ1A} = I_{DQB} = 30$ mA, $I_{DQ2A} = 230$ mA,
 $V_{GS2B} = 1.4$ Vdc, $P_{out} = 10$ W Avg.

f MHz	Z_{in} Ω	Z_{load} Ω
1725	$24.58 + j28.09$	$13.68 - j7.83$
1750	$30.62 + j35.84$	$14.09 - j7.95$
1775	$39.80 + j43.59$	$14.42 - j8.13$
1800	$53.16 + j51.72$	$14.72 - j8.33$
1825	$75.48 + j54.32$	$15.02 - j8.57$
1850	$101.49 + j44.03$	$15.26 - j8.91$
1875	$127.43 + j11.39$	$15.47 - j9.29$
1900	$113.52 - j23.46$	$15.59 - j9.67$
1925	$92.03 - j36.95$	$15.66 - j10.15$
1950	$74.95 - j38.10$	$15.64 - j10.65$
1975	$64.95 - j35.67$	$15.59 - j11.22$
2000	$59.30 - j32.57$	$15.41 - j11.76$
2025	$55.28 - j28.90$	$15.20 - j12.36$
2050	$52.85 - j26.07$	$14.84 - j12.97$
2075	$51.34 - j23.91$	$14.42 - j13.56$

Note: Measured with Peaking side open.

Z_{in} = Device input impedance as measured from gate to ground.

Z_{load} = Test circuit impedance as measured from drain to ground.

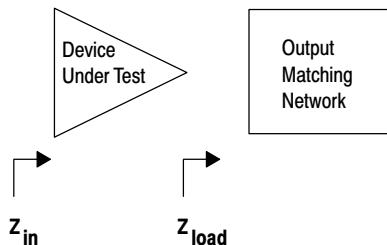


Figure 18. Series Equivalent Input and Load Impedance — Carrier Side — 1880 MHz

$V_{DD} = 28$ Vdc, $I_{DQ1A} = I_{DQB} = 30$ mA, $I_{DQ2A} = 230$ mA,
 $V_{GS2B} = 1.4$ Vdc, $P_{out} = 10$ W Avg.

f MHz	Z_{in} Ω	Z_{load} Ω
1725	$24.58 + j28.09$	$4.10 - j18.22$
1750	$30.62 + j35.84$	$3.61 - j17.55$
1775	$39.80 + j43.59$	$3.09 - j16.79$
1800	$53.16 + j51.72$	$2.61 - j16.00$
1825	$75.48 + j54.32$	$2.31 - j15.22$
1850	$101.49 + j44.03$	$1.99 - j14.46$
1875	$127.43 + j11.39$	$1.71 - j13.71$
1900	$113.52 - j23.46$	$1.47 - j12.96$
1925	$92.03 - j36.95$	$1.27 - j12.19$
1950	$74.95 - j38.10$	$1.15 - j11.44$
1975	$64.95 - j35.67$	$1.04 - j10.70$
2000	$59.30 - j32.57$	$1.00 - j9.97$
2025	$55.28 - j28.90$	$0.98 - j9.28$
2050	$52.85 - j26.07$	$1.05 - j8.57$
2075	$51.34 - j23.91$	$1.16 - j7.91$

Note: Measured with Carrier side open.

Z_{in} = Device input impedance as measured from gate to ground.

Z_{load} = Test circuit impedance as measured from drain to ground.

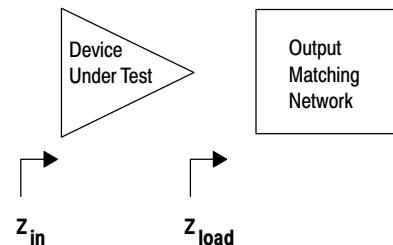
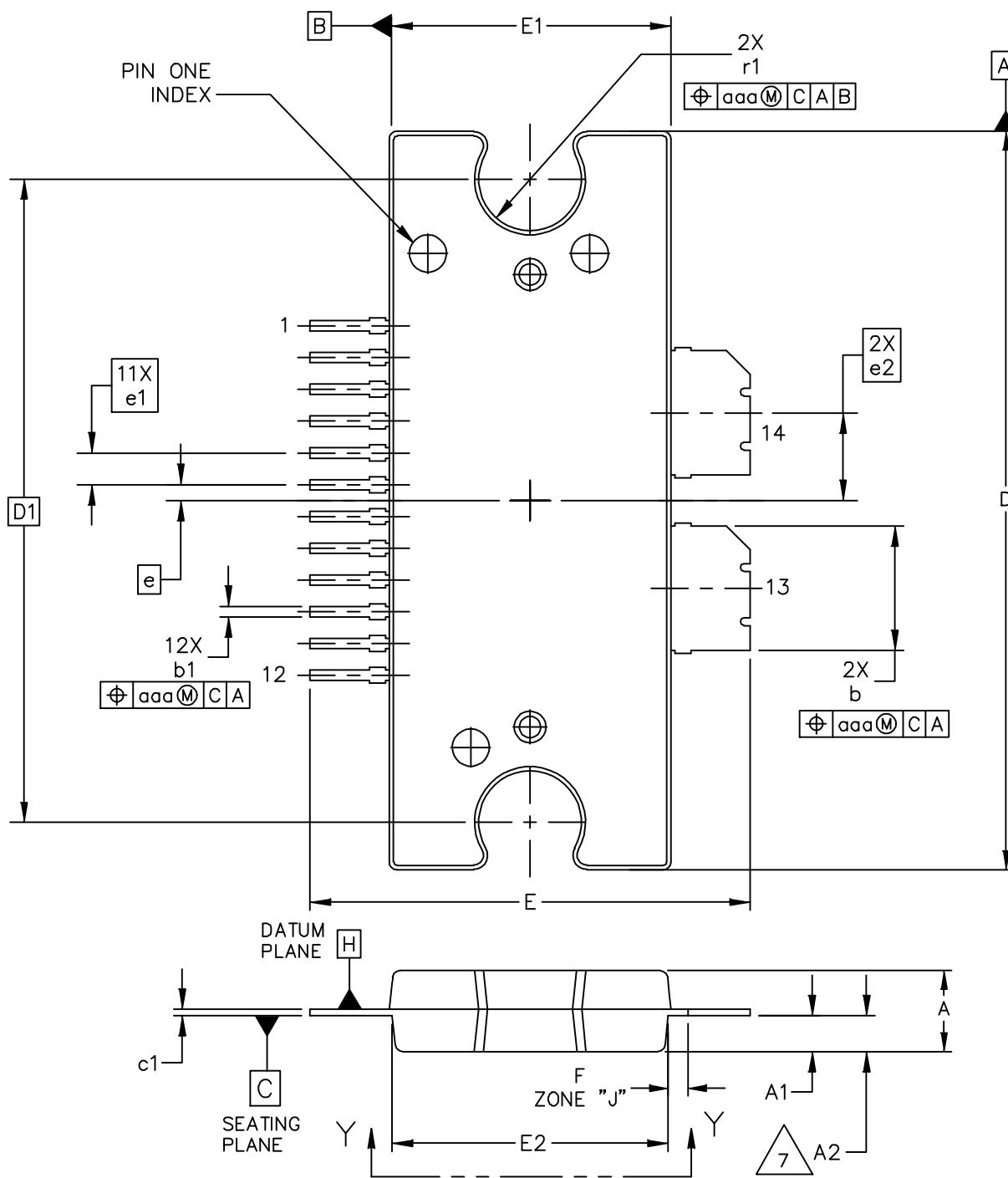


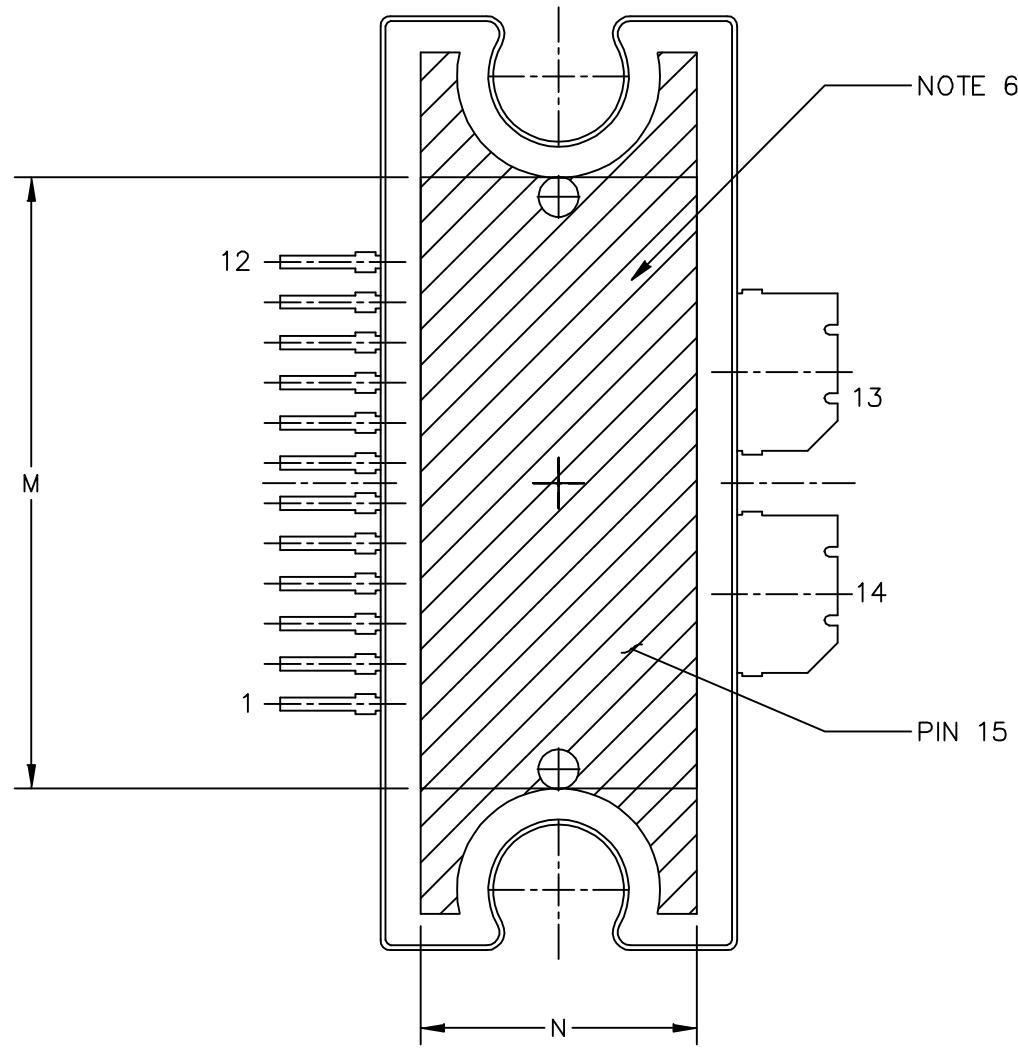
Figure 19. Series Equivalent Input and Load Impedance — Peaking Side — 1880 MHz

PACKAGE DIMENSIONS



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VIEW Y-Y

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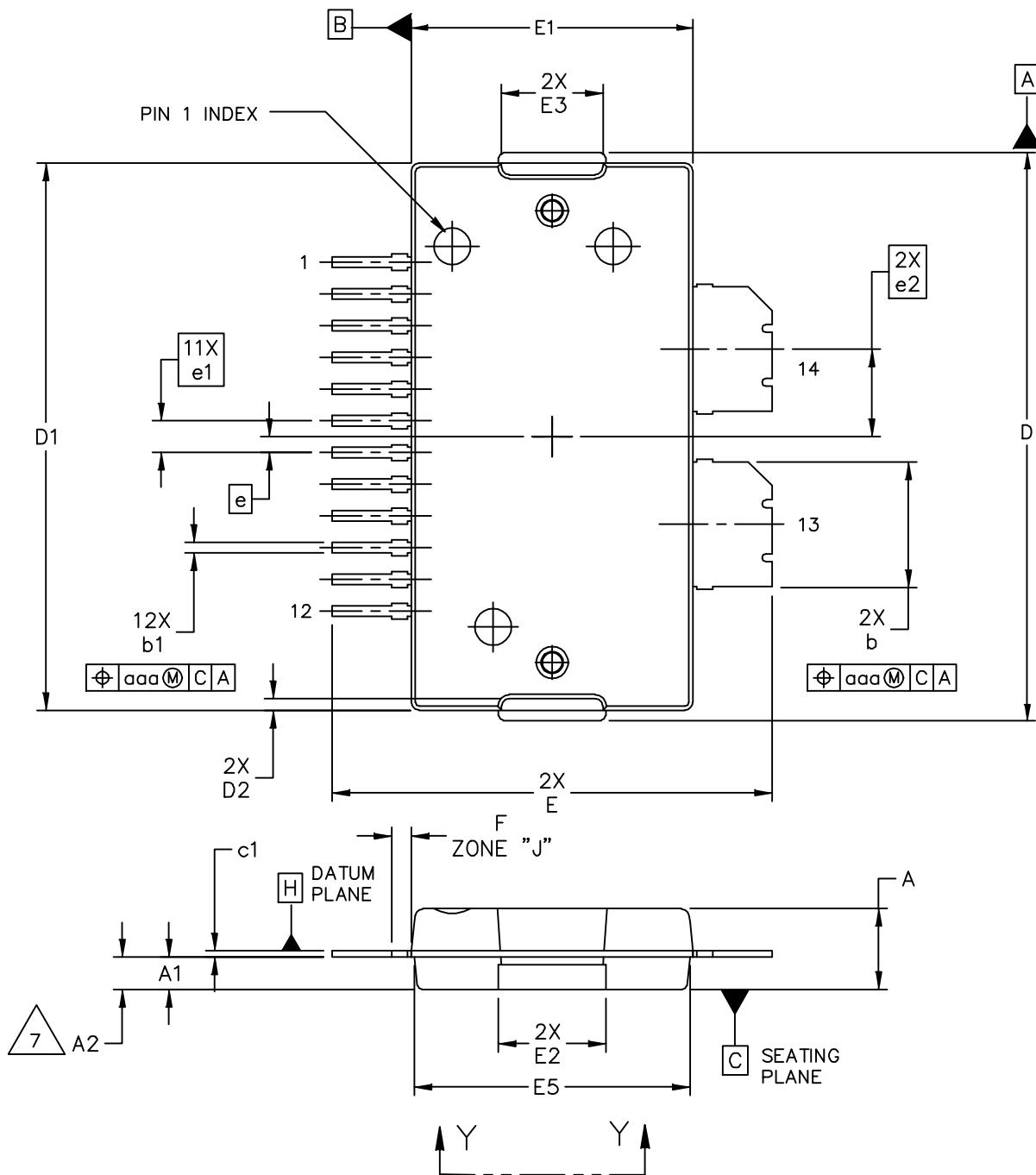
NOTES:

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3. DATUM PLANE -H- IS LOCATED AT THE TOP OF LEAD AND IS COINCIDENT WITH THE LEAD WHERE THE LEAD EXITS THE PLASTIC BODY AT THE TOP OF THE PARTING LINE.
4. DIMENSIONS "D" AND "E1" DO NOT INCLUDE MOLD PROTRUSION. ALLOWABLE PROTRUSION IS .006 (.15) PER SIDE. DIMENSIONS "D" AND "E1" DO INCLUDE MOLD MISMATCH AND ARE DETERMINED AT DATUM PLANE -H-.
5. DIMENSIONS "b" AND "b1" DO NOT INCLUDE DAMBAR PROTRUSION. ALLOWABLE DAMBAR PROTRUSION SHALL BE .005 (.13) TOTAL IN EXCESS OF THE "b" AND "b1" DIMENSIONS AT MAXIMUM MATERIAL CONDITION.
6. HATCHING REPRESENTS THE EXPOSED AREA OF THE HEAT SLUG.
7. DIM A2 APPLIES WITHIN ZONE "J" ONLY.

DIM	INCH		MILLIMETER		DIM	INCH		MILLIMETER	
	MIN	MAX	MIN	MAX		MIN	MAX	MIN	MAX
A	.100	.104	2.54	2.64	b	.154	.160	3.91	4.06
A1	.039	.043	0.99	1.09	b1	.010	.016	0.25	0.41
A2	.040	.042	1.02	1.07	c1	.007	.011	0.18	0.28
D	.928	.932	23.57	23.67	e	.020 BSC		0.51 BSC	
D1	.810 BSC		20.57	BSC	e1	.040 BSC		1.02 BSC	
E	.551	.559	14.00	14.20	e2	.1105 BSC		2.807 BSC	
E1	.353	.357	8.97	9.07	r1	.063	.068	1.6	1.73
E2	.346	.350	8.79	8.89	aaa		.004		0.10
F	.025 BSC		0.64	BSC					
M	.600	-----	15.24	-----					
N	.270	-----	6.86	-----					

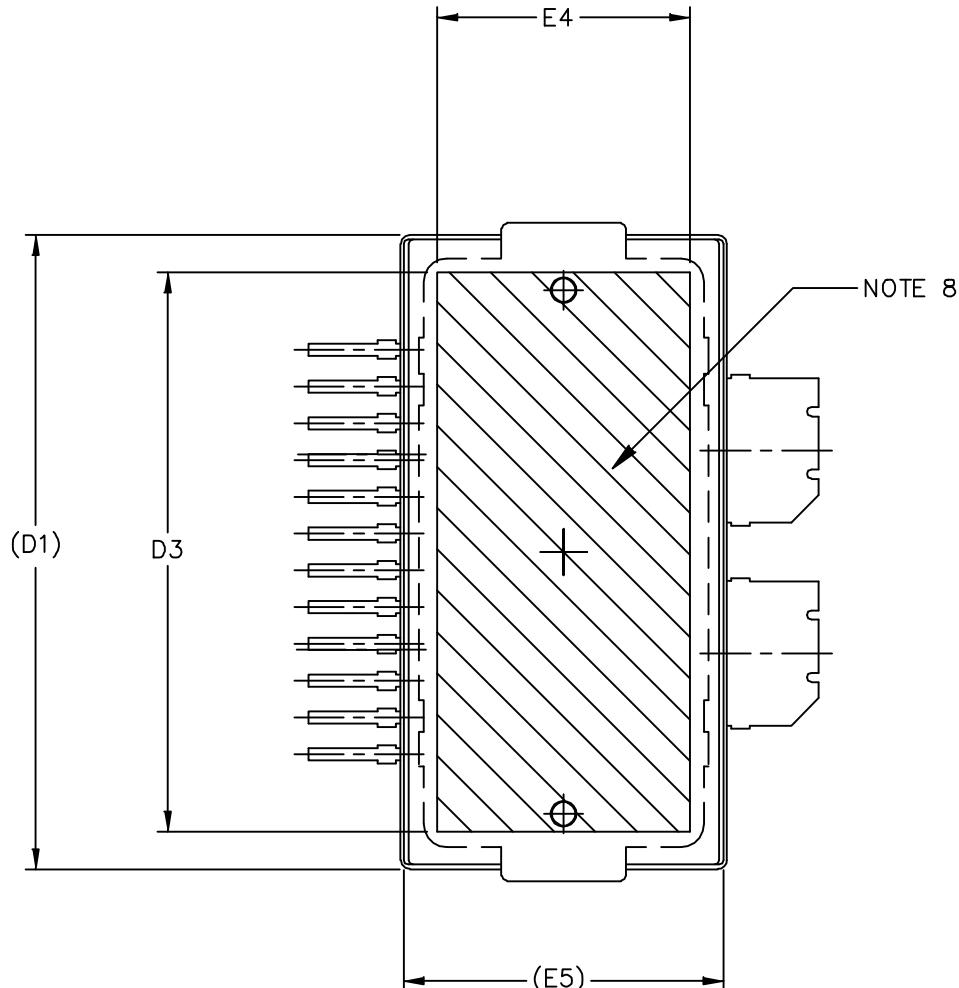
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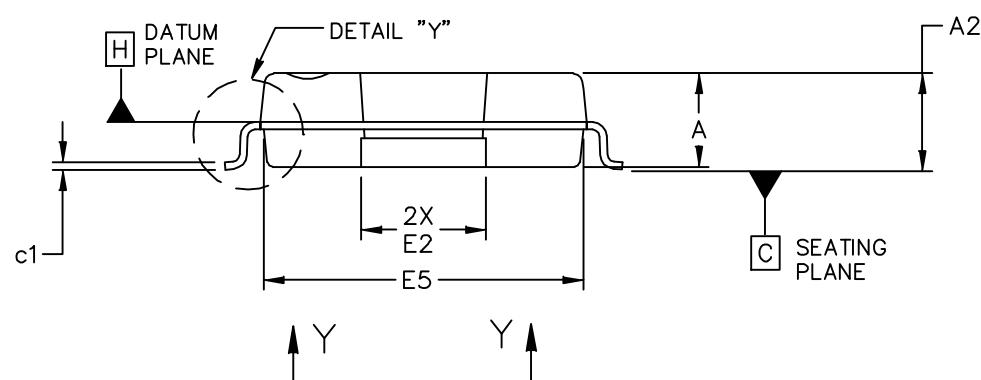
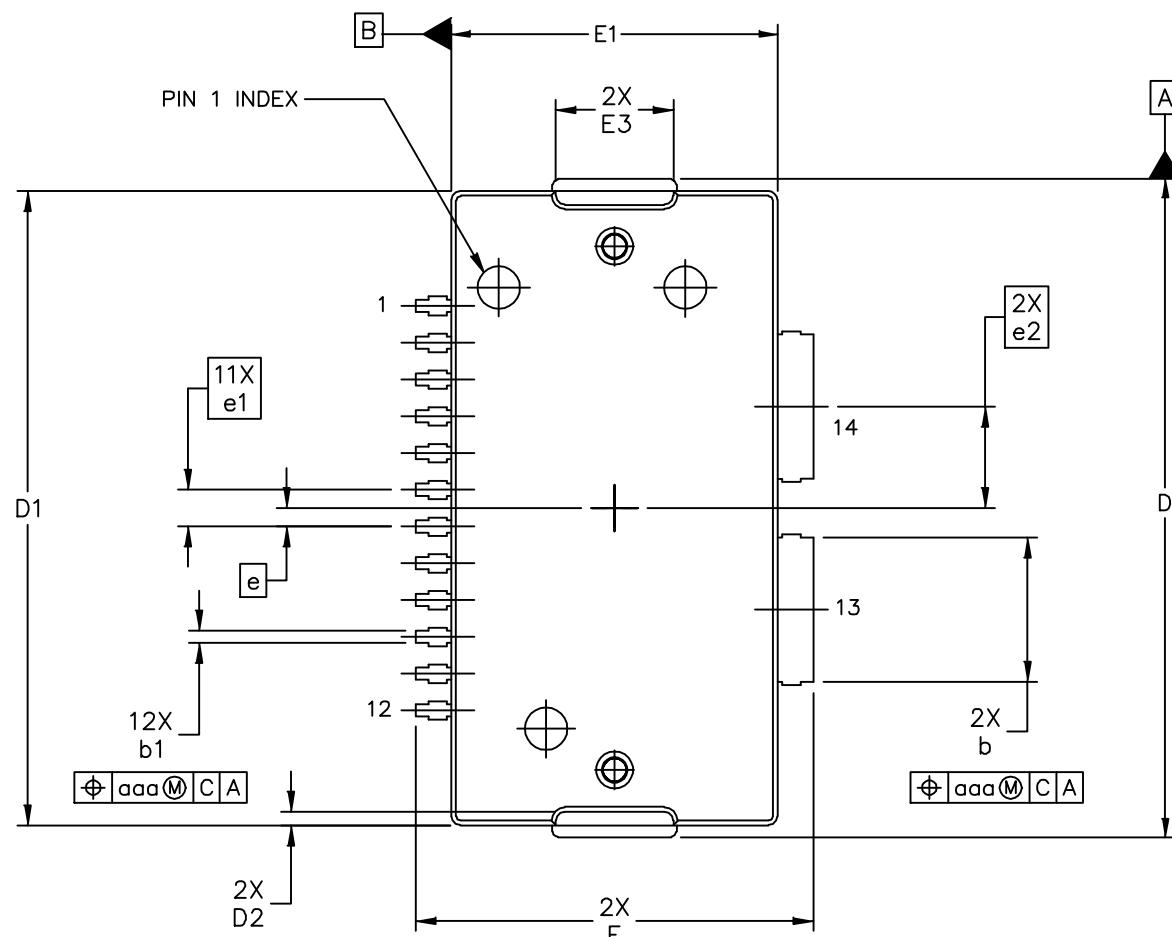
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6. DATUMS -A- AND -B- TO BE DETERMINED AT DATUM PLANE -H-.
7. DIMENSION A2 APPLIES WITHIN ZONE "J" ONLY.
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DIM	INCH		MILLIMETER		DIM	INCH		MILLIMETER	
	MIN	MAX	MIN	MAX		MIN	MAX	MIN	MAX
A	.100	.104	2.54	2.64	F	.025	BSC	0.64	BSC
A1	.039	.043	0.99	1.09	b	.154	.160	3.91	4.06
A2	.040	.042	1.02	1.07	b1	.010	.016	0.25	0.41
D	.712	.720	18.08	18.29	c1	.007	.011	.18	.28
D1	.688	.692	17.48	17.58	e	.020	BSC	0.51	BSC
D2	.011	.019	0.28	0.48	e1	.040	BSC	1.02	BSC
D3	.600	---	15.24	---	e2	.1105	BSC	2.807	BSC
E	.551	.559	14	14.2					
E1	.353	.357	8.97	9.07	aaa	.004			.10
E2	.132	.140	3.35	3.56					
E3	.124	.132	3.15	3.35					
E4	.270	---	6.86	---					
E5	.346	.350	8.79	8.89					

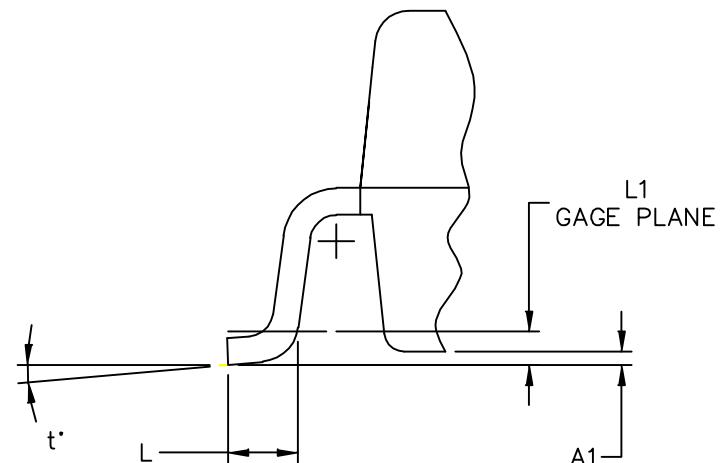
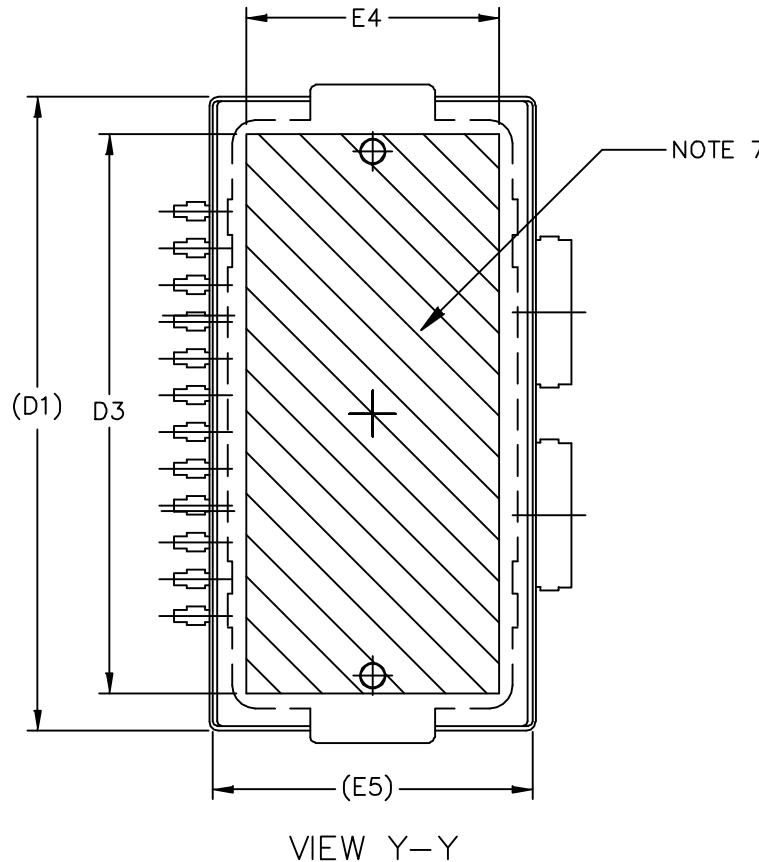
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DETAIL "Y"

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6. DATUMS -A- AND -B- TO BE DETERMINED AT DATUM PLANE -H-.
7. HATCHING REPRESENTS THE EXPOSED AREA OF THE HEAT SLUG.

DIM	INCH		MILLIMETER		DIM	INCH		MILLIMETER	
	MIN	MAX	MIN	MAX		MIN	MAX	MIN	MAX
A	.100	.104	2.54	2.64	L	.018	.024	0.46	0.61
A1	.001	.004	0.02	0.10	L1	.010	BSC	0.25	BSC
A2	.099	.110	2.51	2.79	b	.154	.160	3.91	4.06
D	.712	.720	18.08	18.29	b1	.010	.016	0.25	0.41
D1	.688	.692	17.48	17.58	c1	.007	.011	.18	.28
D2	.011	.019	0.28	0.48	e	.020	BSC	0.51	BSC
D3	.600	---	15.24	---	e1	.040	BSC	1.02	BSC
E	.429	.437	10.9	11.1	e2	.1105	BSC	2.807	BSC
E1	.353	.357	8.97	9.07	t	2"	8"	2"	8"
E2	.132	.140	3.35	3.56	aaa	.004		.10	
E3	.124	.132	3.15	3.35					
E4	.270	---	6.86	---					
E5	.346	.350	8.79	8.89					

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PRODUCT DOCUMENTATION AND SOFTWARE

Refer to the following documents to aid your design process.

Application Notes

- AN1907: Solder Reflow Attach Method for High Power RF Devices in Plastic Packages
- AN1955: Thermal Measurement Methodology of RF Power Amplifiers
- AN1977: Quiescent Current Thermal Tracking Circuit in the RF Integrated Circuit Family
- AN1987: Quiescent Current Control for the RF Integrated Circuit Device Family
- AN3263: Bolt Down Mounting Method for High Power RF Transistors and RFICs in Over-Molded Plastic Packages
- AN3789: Clamping of High Power RF Transistors and RFICs in Over-Molded Plastic Packages

Engineering Bulletins

- EB212: Using Data Sheet Impedances for RF LDMOS Devices

Software

- Electromigration MTTF Calculator
- RF High Power Model
- .s2p File

For Software, do a Part Number search at <http://www.freescale.com>, and select the “Part Number” link. Go to the Software & Tools tab on the part’s Product Summary page to download the respective tool.

REVISION HISTORY

The following table summarizes revisions to this document.

Revision	Date	Description
0	Aug. 2009	<ul style="list-style-type: none"> • Initial Release of Data Sheet
1	May 2010	<ul style="list-style-type: none"> • Corrected Thermal Characteristics table values for thermal resistance as follows: 50 W CW I_{DQ1B} changed from 6.1 to 8.2 °C/W and V_{GS2B} changed from 1.4 to 1.8 °C/W; 10 W CW I_{DQ1B} changed from 3.6 to 8.3 °C/W and V_{GS2B} changed from *Stage 2B is turned off to 1.9 °C/W. Thermal values now reflect the symmetrical Doherty nature of the device, p. 2 • Changed ESD Human Body Model rating from Class 1B to Class 0 to reflect recent ESD test results of the device, p. 3 • Added RF High Power Model availability to Product Software, p. 23

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