

**Application Manual** 

# Programmable Crystal Oscillator SG-8506CA

SEIKO EPSON CORPORATION

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## 1. Overview

Programmable crystal oscillator: SG-8506CA is a low jitter programmable XO at any frequency. Its output frequency is programmable from 50 MHz to 800 MHz with almost 2 ppb resolution.

SG-8506CA consists of XO, PLL and LVPECL output buffer.

XO supplies stable reference clock to PLL with fundamental tone crystal.

PLL consists of a low jitter fractional-N PLL technology. The components for loop filter are embedded into IC, so no external filter component is needed.

- Programmable clock output frequency from 50 MHz to 800 MHz
- Frequency setting resolution is around 2 ppb
- Low jitter and high reliability clock source from the fundamental tone internal crystal
- Low jitter and low noise PLL
- One factory preset power-up default frequency
- Programmable one preset power-up default frequency (Only the blank Sample can be programmed one time with SG-Writer II)
- Factory preset device options
  - OE polarity
  - Output standby type: Hi-Z or OUT = "L", OUTN = "H"
  - I<sup>2</sup>C interface slave address
- Embedded resistors and capacitors for oscillator and loop filter for PLL
- I<sup>2</sup>C interface
- LVPECL output
- 8-pin ceramic 5 x 7 mm package
- 2.5 V or 3.3 V supply voltage modes
- -40 ℃ ~ +85 ℃ ambient operating temperature
- Pb-free / RoHS-compliant

#### [Evaluation Kit]

The evaluation kit of SG-8506CA is available. Please ask us for more information.

VG7050EAN I2C Slave Address (Hex)	7 Search Addr	R	egister Re	ad Table		
		3	Addr	Value	^	
VG7050ECN Reference Frequency 114.1444	44 [MHz]	•	00	00		xoute xoute
SG-8506CA			01	00		
			02	00		
uency Setting			03	00		
			04	00		VDD PVSS R4 0 0
equency Setting No. 0			05	00		
			06	00		R50
Output Frequency 156.250000	MHz]		07	00		
Register Register Register			08	00		GND
	Divider		09	00		
Address Mane Value (nex)	Sivider		0A	00		PVSS GND VC
0x10 ODIV 09 20	-		OB	00		
0x11 NINT 1B	27		OC	00		
	21 -		OD	00	*	
0x12,0x13,0x14 NFRAC 60A9DC (	.37759185			Read		
0	Write and	Port	10			
Write	Verify	FSE	1	OE		
			00b	• H		
γ						S *2.8 2031
PLL Oscillator Register Writer, Version 1.0.0.0	*		01b	ΘL		
			105			
			11b			
						•

# 2. Part Number

• Standard (Factory preset start-up frequency product)



• Blank (one-time programmable start-up frequency product)



# 3. Block Diagram



\* If OE pin is configured as active low, OE pin is pulled down to GND with internal pull down resistor.

Figure 3.1. SG-8506CA Block Diagram

# 4. Pin Assignments

# 4.1. Pin Assignments



# 4.2. Pin Descriptions

Table 4.1Pin Descriptions

No.	Pin Name	Туре	)	Function					
1	NC	-	-	No Connect Connect to GND/V <sub>CC</sub> or open					
2	OE	Input	Pull-up/	Output Enable (Active High)					
				OE InputOUT, OUTN pin status"H" or OpenOutputs are enabled."L"High-impedance state or OUT = "L", OUTN = "H"					
			Pull-down	Output Enable (Active Low)					
				OE InputOUT, OUTN pin status"H"High-impedance state or OUT = "L", OUTN = "H""L" or OpenOutputs are enabled.					
3	GND	Power	-	Negative Power Supply					
4	OUT	Output	-	Differential clock output. LVPECL interface levels.					
5	OUTN	Output	-						
6	V <sub>cc</sub>	Power	-	Positive Power Supply					
7	SDA <sup>*1</sup>	Input/Output	-	I <sup>2</sup> C Data Input/Output Input: LVCMOS interface levels, Output: Open drain					
8	SCL <sup>*1</sup>	Input	-	I <sup>2</sup> C Clock Input					
		Pull-down" refers ull-up resistor to		CA internal input resistors. sary.					

# 5. Electrical Characteristics

# 5.1. Absolute Maximum Ratings

Item Symbol		Condition	Min.	Тур.	Max.	Units		
Supply voltage, V <sub>CC</sub>	V <sub>cc</sub>	GND = 0 V	-0.3	-	4.0	V		
Pull-up voltage	V <sub>PU</sub>	SDA, SCL	-0.3	-	4.0	V		
Input voltage 1	V <sub>in1</sub>	GND = 0 V, Input pins except to SDA and SCL	GND - 0.3	-	V <sub>CC</sub> + 0.3	V		
Input voltage 2	V <sub>in2</sub>	GND = 0 V, SDA, SCL	GND - 0.3	-	4.0	V		
Storage temperature	Tstg	Store as bare product	-55	-	+125	C		
ESD sensitivity	ESD	НВМ	2000	-	-	V		
		MM	200	-	-			
Note: Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only. Functional operation of the device at these or any other conditions beyond those listed in the "DC characteristics" or "AC characteristics" is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.								

## 5.2. **DC Characteristics**

Table 5.1.	Power Supply, Operating Temperature
------------	-------------------------------------

			GND	= 0 V, Ta	= -40 ~ +	-85 °C		
Item	Symbol	Conditions	Min.	Тур.	Max.	Units		
Positive supply voltage	V <sub>cc</sub>	-	2.375	2.5/3.3	3.630	V		
Positive supply current <sup>1</sup> Output enable mode	Icc	OE = Enable, Outputs terminated with 50 $\Omega$ to V <sub>CC</sub> – 2.0 V	-	-	90	mA		
Positive supply current <sup>*1</sup> I_dis OE = Disable, Output standby type: Dutput disable mode Hi-Z		-	-	40	mA			
		OE = Disable, Output standby type: Fix (OUT = "L", OUTN = "H")	-	-	70	mA		
Operating temperature	Та	-	-40	-	+85	C		
Note 1: Guaranteed by design, characterization, and/or simulation only and not production tested.								

	V <sub>CC</sub>	$= 2.5 \text{ V} - 5\% \sim 3.3 \text{ V} + 10$	)%, GND =	= 0 V, Ta	= -40 ~ +8	35 °C
Item	Symbol	Conditions	Min.	Тур.	Max.	Units
Pull-up voltage	V <sub>PU</sub>	SDA, SCL	V <sub>CC</sub> x 0.7	-	3.630	V
High level input voltage 1	V <sub>IH1</sub>	OE	V <sub>CC</sub> x 0.7	-	V <sub>CC</sub> + 0.3	V
High level input voltage 2	V <sub>IH2</sub>	SDA, SCL, Pull Up Voltage = $V_{PU}$	V <sub>CC</sub> x 0.7	-	3.630	V
Low level input voltage	VIL	SDA, SCL, OE	-0.3	-	V <sub>CC</sub> x 0.3	V
High level input current 1	I <sub>IH1</sub>	SDA, SCL, OE (Active High)	-	-	2	μA
High level input current 2	I <sub>IH2</sub>	OE (Active Low)	-	-	170	μA
Low level input current 1	I <sub>IL1</sub>	SDA, SCL	-2	-	-	μA
Low level input current 2 IIL2 C		OE (Active High)	-70	-	-	μA
Low level output voltage	V <sub>OL</sub>	SDA, at 3 mA sink current	0	-	0.4	V
Low level output current	I <sub>OL</sub>	SDA, $V_{OL} = 0.4 V$	3	-	-	mA
Pull-up resistor	R <sub>UP</sub>	OE (Active High)	-	85	-	kΩ
	R <sub>DOWN</sub>	OE (Active Low)	-	35	-	
Input Capacitance <sup>*1</sup>	CIN	OE, SDA, SCL	-	5	-	pF
Note 1: Guaranteed by design	gn, charact	terization, and/or simulation only and	d not producti	on tested.		

### 5.3. AC Characteristics

Table 5.3.	<b>Output Frequency Characteristics</b>
------------	---

	V <sub>CC</sub>	<u> </u>	2.5 V - 5%	~ 3.3 V + 10	0%, GND	<u>= 0 V, Ta</u>	= -40 ~	+85 ℃
Item	Symbol		Conditi	ons	Min.	Тур.	Max.	Units
Output frequency	f <sub>O</sub>	οι	JT, OUTN		50	-	800	MHz
Internal crystal frequency	<b>f</b> <sub>XTAL</sub>		-	-	114.144	-	MHz	
Frequency reprogramming resolution	M <sub>RES</sub>		-		2.2	-	2.8	ppb
Frequency tolerance <sup>*1</sup>	f_tol	fre su	is parameter inc quency toleranc pply voltage vari ars aging <sup>*2</sup> at 25	-50	-	+50	10 <sup>-6</sup>	
Delta frequency for continuous output <sup>*1</sup>	-		om Center Frequ fined by setting		-500	-	+500	10 <sup>-6</sup>
Setting time for large frequency change <sup>1</sup>	t <sub>SET1</sub>		om setting NEW		-	-	1.5	ms
Setting time for small frequency change <sup>*1</sup>	t <sub>SET2</sub>	fre	±500 ppm from c equency that is d tting NEW_FRE	-	-	100	μs	
SSB phase noise <sup>*1</sup>	F <sub>CN</sub>	fo	= 622.08 MHz, f	rom carrier				
			$V_{CC} = 3.3 V^{*3}$	100 Hz	-	-76.5	-	dBc/Hz
				1 kHz	-	-103.1	-	
				10 kHz	-	-119.4	-	
				100 kHz	-	-121.3	-	
				1 MHz	-	-129.1	-	
				10 MHz	-	-146.8	-	
			$V_{CC} = 2.5 V^{*4}$	100 Hz	-	-75.5	-	
				1 kHz	-	-101.1	-	
				10 kHz	-	-118.9	-	
				100 kHz	-	-121.3	-	
				1 MHz	-	-129.0	-	-
				10 MHz	-	-146.7	-	
RMS phase jitter <sup>*1, *4</sup>	t <sub>PJ</sub>	fo	= 622.08 MHz, I	ntegration range	: 12 kHz – 2	0 MHz (OC-4	.8)	1
			$V_{CC} = 3.3 V^{*3}$		-	0.3	-	ps
			$V_{CC} = 2.5 V^{*4}$		-	0.3	-	ps
		$f_0 = 622.08$ MHz, Integration range: 20 kHz – 50 MHz					1	
			$V_{CC} = 3.3 V^{*3}$		-	0.3	-	ps
			$V_{\rm CC} = 2.5 \ V^{*4}$		-	0.3	-	ps
		fo		ntegration range	: 50 kHz – 8	0 MHz (OC-1	92)	
			$V_{CC} = 3.3 V^{3}$	-	-	0.3	-	ps
			$V_{CC} = 2.5 V^{*4}$		-	0.3	-	ps

Note 1: Guaranteed by design, characterization, and/or simulation only and not production tested.

Note 2: The aging in the frequency tolerance is from environmental tests results to the expectation of the amount of the frequency variation. This doesn't guarantee the product life cycle.

Note 3:  $f_{XTAL} = 114.144$  MHz, Ta = +25 °C, V <sub>CC</sub> = 3.3 V. Note 4:  $f_{XTAL} = 114.144$  MHz, Ta = +25 °C, V <sub>CC</sub> = 2.5 V.

Note 5: The output clock may contain spurious that depends on the settings of fo, f<sub>XTAL</sub>, PLL and output divider. The RMS jitter may be worse, if the spurious is in integration range of RMS jitter. For more information, please contact us.



Frequency Change Time



Phase Noise Test Circuit

Table	5.4.	Serial	Interface

$V_{\rm CC} = 2.$	5 V - 5%	~ 3.3 V +	10%, GND	= 0 V, Ta	= -40 ~ +	85 C
Item	Symbol	Conditions	Min.	Тур.	Max.	Units
SCL clock frequency	f <sub>SCL</sub>	-	-	-	400	kHz
Hold time (repeated) START condition, After this period, the first clock pulse is generated.	t <sub>HD;STA</sub>	-	0.6	-	-	μs
Low period of the SCL clock	t <sub>LOW</sub>	-	1.3	-	-	μs
High period of the SCL clock	t <sub>HIGH</sub>	-	0.6	-	-	μs
Set up time for a repeated START condition	t <sub>SU;STA</sub>	-	0.6	-	-	μs
Input data hold time	t <sub>HD;DAT</sub>	-	0	-	-	μs
Output data set-up time	t <sub>SU;DAT</sub>	-	100	-	-	ns
Rise time of both SDA and SCL signals <sup>*1</sup>	tr	-	-	-	300	ns
Fall time of both SDA and SCL signals	t <sub>f</sub>	-	-	-	300	ns
Set up time for STOP condition	t <sub>SU;STO</sub>	-	0.6	-	-	μs
Bus free time between a STOP and START condition	t <sub>BUF</sub>	-	1.3	-	-	μs
Data valid time	t <sub>VD:DAT</sub>	-	-	-	0.9	μs
Data valid acknowledge time	t <sub>VD;ACK</sub>	-	-	-	0.9	μs
Note 1: Guaranteed by design, characterization	n, and/or sin	nulation only	and not producti	on tested.		





**Serial Interface** 

## SG-8506CA

## 5.4. LVPECL

#### Table 5.5. LVPECL

	Vcc	$c = 2.5 \text{ V} - 5\% \sim 3.3 \text{ V}$	+ 10%, GN	D = 0 V, Ta	a = -40 ~ +	85 °C			
Item	Symbol	Conditions	Min.	Тур.	Max	Units			
Output load condition	L_PECL	Outputs terminated with 50 $\Omega$	Dutputs terminated with 50 $\Omega$ to V <sub>CC</sub> – 2.0 V -						
Rise time <sup>*1</sup>	t <sub>R</sub>	-	-	-	400	ps			
Fall time <sup>*1</sup>	t <sub>F</sub>	-	-	-	400	ps			
Symmetry <sup>*1</sup> (duty cycle)	SYM	-	45	50	55	%			
High level output voltage	V <sub>OH</sub>	-	V <sub>CC</sub> – 1.025	V <sub>CC</sub> – 0.95	-	V			
Low level output voltage	V <sub>OL</sub>	-	-	V <sub>CC</sub> – 1.7	V <sub>CC</sub> – 1.62	V			
OE disable delay time <sup>*1</sup>	t <sub>PXZ</sub>	-	-	-	100	ns			
OE enable delay time <sup>*1</sup>	t <sub>pZX</sub>	-	-	-	10	μs			
Note: OUT and OUTN are not used as single end									

OUT and OUTN are not used as single end. ote:

Note 1: Guaranteed by design, characterization, and/or simulation only and not production tested.



Output Rise/Fall Time, Symmetry (duty cycle)



**Output AC Test Circuit** 









## 5.5. Startup

#### Table 5.6. Startup

		$V_{\rm CC} = 2.5 \ V - 5\% \sim 3.3 \ V$	′ + 10%, GN	ID = 0 V, Ta	a = -40 ~ +	·85 °C	
Item	Symbol	Conditions	Min.	Тур.	Max	Units	
V <sub>CC</sub> ramp rate <sup>*1</sup>	R <sub>VCC</sub>	$V_{CC}$ from 0 V to $V_{CCMIN}.$	5 x 10 <sup>-6</sup>	-	3	S	
Startup time <sup>*2</sup>	t_str	-	-	-	5	ms	
I <sup>2</sup> C I/F enable time <sup>*2</sup>	t <sub>I2CEN</sub>	-	-	-	5	ms	
Note 1: V <sub>CC</sub> ramp must be monotonic.							

Note 2: Guaranteed by design, characterization, and/or simulation only and not production tested.



a) Output standby type: Hi-Z



Start-Up Time

# 6. Functions

#### 6.1. Overview

The SG-8506CA has a XO, PLL and output buffer unit. The XO unit is composed of a fundamental mode crystal that generates stable reference clock for PLL. The output frequency is determined by the feedback divider and the output divider. The feedback divider can offer not only integer setting that achieves lower jitter, but also fractional setting that provides frequency in ppb resolution.

The device's default output frequency can be set at the factory and can be reprogrammed via l<sup>2</sup>C bus. Once the device is powered down, it will return to its factory-set default setting.

#### 6.2. Setting of the Output Frequency

#### 6.2.1. Calculation of the Frequency Setting

The output frequency ( $f_0$ ) is determined by the VCO frequency ( $f_{VCO}$ ) and the output divider (ODIV). This is shown:

$$f_0 = \frac{f_{VCO}}{ODIV} \tag{1}$$

The VCO frequency must be from 2.55 GHz to 3.20 GHz. Base on the relation between this limit and the formula (1), ODIV is calculated from the  $f_0$  as shown in Table 6.1.

The VCO frequency is determined by the reference frequency ( $f_{REF}$ ) from the XO and the feedback divider (N). The feedback divider (N) consists of both a 6-bit integer portion ( $N_{INT}$ ) and a 24-bit fractional portion ( $N_{FRAC}$ ) and provides the means for high-resolution frequency generation. The VCO frequency is calculated by:

$$f_{VCO} = f_{REF} \times N$$
  
=  $f_{REF} \times \left( N_{INT} + \frac{N_{FRAC}}{2^{24}} \right)$  (2)

f <sub>o</sub> [MHz]	ODIV	ODIV.ODIV register setting
50 ~ 57	56	0xF
53 ~ 67	48	0xE
64 ~ 80	40	0xD
80 ~ 100	32	0xC
91 ~ 114	28	0xB
106 ~ 133	24	0xA
128 ~ 160	20	0x9
159 ~ 200	16	0x8
182 ~ 229	14	0x7
213 ~ 267	12	0x6
255 ~ 320	10	0x5
319 ~ 400	8	0x4
364 ~ 457	7	0x3
425 ~ 533	6	0x2
510 ~ 640	5	0x1
638 ~ 800	4	0x0

Table 6.1. fo and ODIV

The output frequency (f<sub>O</sub>) is shown:

$$f_{O} = \frac{f_{VCO}}{ODIV}$$
$$= f_{REF} \frac{\left(N_{INT} + \frac{N_{FRAC}}{2^{24}}\right)}{ODIV}$$

(3)

(4)

(5)

For example if the reference frequency ( $f_{REF}$ ) is 114.144444 MHz and the output frequency is 120.0 MHz, ODIV is fixed to "24" from the Table 6.1. The setting of N, N<sub>INT</sub>, N<sub>FRAC</sub> is calculated:

$$N = N_{INT} + \frac{N_{FRAC}}{2^{24}} = \frac{f_{OUT} \times ODIV}{f_{REF}} = \frac{120.0 \times 10^6 \times 24}{114.1444444 \times 10^6} = 25.231188535690308$$

$$N_{INT} = floor(N) = floor(25.231188535690308) = 25$$

$$N_{FRAC} = (N - N_{int}) \times 2^{24} = (25.231188535690308 - 25) \times 2^{24}$$
  
= 0.231188535690308 \times 2^{24}  
\approx 3878700 = 0x3B2F2C (6)

Depending on the fo, the ODIV may become two values.

For example if the fo is 380 MHz, ODIV can be 7 or 8. Even if either of the ODIV values is selected, the same fo can be gained by setting NINT and NFRAC but phase noise included in the output signal become different. Please evaluate the performances fully in your actual usage environment and select the ODIV.

 $N_{FRAC}$  is a 24-bit value. By setting 6 bit of  $N_{INT}$  and 20 bit of  $N_{INT}$  frequency resolution is 10 ppb order. The lower 4 bit of the rest of the  $N_{FRAC}$  corresponds to the setting of the frequency in 1ppb order. By setting these values, the output frequency is changed very small, but the spurious of the output signal may change significantly. Please evaluate the performances fully in your actual usage environment and fix the lower 4 bit of the  $N_{FRAC}$ .

#### 6.2.2. Reconfiguring Frequency Setting

The SG-8506CA has a "user register" and a "PLL register". The user register stores ODIV, NINT and NFRAC. It can be reprogrammed at any time when I<sup>2</sup>C bus is available. The PLL register is connected directly to the PLL. When the device is powered on, the default value programmed in the non-volatile memory is automatically fetched to the user register, and the PLL register is updated with it.

The PLL register is also updated with the user register, by writing PLL\_CTRL.NEW\_FREQ or PLL\_CTRL.SML\_CHG register. This flow is shown in Figure 6.1.



Figure 6.1. Reconfiguring Frequency Setting

First, ODIV, NINT, and NFRAC in the user register need to be changed. For details of the user register please refer to the chapter 6.

Next by writing 1 in the PLL\_CTRL.NEW\_FREQ register or PLL\_CTRL.SML\_CHG register, frequency setting can be forwarded from the user register to the PLL register. As a result, output signal of frequency ( $f_0$ ) is updated. Difference between the PLL\_NEW\_FREQ register and the PLL\_SML\_CHG is shown in Table 6.2

Table 6.2.	Updating tl	he frequency	setting
------------	-------------	--------------	---------

No	Register	PLL calibration	Output signal	Frequency pull range
1	PLL_CTRL.NEW_FREQ		Momentarily stopped and start over after PLL is optimized	50 MHz to 800 MHz
2	PLL_CTRL.SML_CHG	N	Continuous output	Within the ±500 ppm window

As 1 is written in the PLL\_CTRL.NEW\_FREQ register, the output clock is momentarily stops and PLL is calibrated to new output frequency. After the calibration, output clock starts at any arbitrary point during a clock cycle. This method has no limitation in frequency change range and provides lower jitter. This also establishes a new center frequency. Circuitry receiving a clock from the SG-8506CA that is sensitive to glitches or runt pulses may have to be reset once this process is complete.

## SG-8506CA

For output clock frequency changes less than ±500 ppm from the center frequency configuration, PLL\_CTRL.SML\_CHG register is available. By writing this register as 1, NINT and NFRAC in the user register are transferred to PLL register and the output frequency is updated without interruption to the output clock. Since the PLL is not calibrated, jitter might be increased. It is not guaranteed that the output frequency is in the frequency range defined by the old and new output frequency.



Figure 6.2. VCO frequency range

## 6.3. I<sup>2</sup>C interface

#### 6.3.1. Connection of I<sup>2</sup>C Bus

The SG-8506CA can be used as a slave device of  $I^2C$  bus. The  $I^2C$  bus is composed of serial data line (SDA) and serial clock (SCL). The lines need to be both pulled up by external resistors. Electric level of the pull up resistor need to be above the Vcc so these are recommended to be pulled up to the Vcc. Also slave address of the slave devices on the  $I^2C$  bus must be unique.



The address of slave devices must be unique

Figure 6.3. Connection of I<sup>2</sup>C bus

#### 6.3.2. I<sup>2</sup>C Bus Protocols Supported by the SG-8506CA

I2C bus protocols that can be supported by the SG-8506CA are shown in the below Table 6.3.

Feature	SG-8506CA
START condition	✓
STOP condition	$\checkmark$
Acknowledge	$\checkmark$
Clock stretching	n/a
7-bit slave address	$\checkmark$
10-bit slave address	n/a
General Call address	n/a
Software Reset	n/a
Device ID	n/a
	n/a = not applicable

Table 6.3.	I <sup>2</sup> C bus protocols supported by the SG-8506CA
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#### 6.3.3. START Condition and STOP Condition

Data communication on the  $I^2C$  bus starts by START condition (S). The START condition means that SDA changes from "H" to "L" when SCL is at "H". When the START condition occurs,  $I^2C$  bus becomes busy.

Data communication on the  $I^2C$  bus can be terminated by STOP condition (P). The STOP condition means that SDA changes from "L" to "H" when SCL is at "H". When the STOP condition occurs,  $I^2C$  bus becomes free.

When  $I^2C$  bus is busy, instead of STOP condition START condition can be generated, which is called repeated START condition (Sr). The  $I^2C$  bus maintains busy status. If the START or repeated START condition is received,  $I^2C$  interface circuit of the SG-8506CA is always reset, even if these START conditions are not positioned according to the proper format.



Figure 6.4. START and STOP condition

#### 6.3.4. Byte Format and ACK/NACK

Data transmission and reception on  $I^2C$  is done in a unit of 8 bit = 1 byte. Each byte is followed by acknowledge bit. Data is transmitted by MSB first. Including acknowledge bit all SCL pulses are generated by Master.

The Acknowledge signal (ACK: A) is defined as follows: the transmitter (master transmitter or slave transmitter) releases the SDA line during the acknowledge clock pulse so the receiver can pull the SDA line "L" and it remains stable "L" during the "H" period of this clock pulse. When SDA remains "H" during this ninth clock pulse, this is defined as the Not Acknowledge signal (NACK:  $\overline{A}$ ).

#### 6.3.5. Read/Write to Register

Procedure of Read/Write to register is shown in the below Figure 6.5. The SG-8506CA can Read/Write single or multi byte data. The SG-8506CA slave address is able to be specified by the customer. It will be programmed to non-volatile memory at our factory.



#### Figure 6.5. Read/Write from/to register by I<sup>2</sup>C bus

# 7. Registers

# 7.1. List of registers

A al alva a a	Register				В	Sit			
Address	name	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
0x00	P_CODE0				0x46 (Ascii 'F	', Read Only	)		
0x01	P_CODE1				0x06 (Re	ead Only)			
0x02	REV				0x01 (Re	ead Only)			
0x03	ID_CODE0				0x01 (Re	ead Only)			
0x04	ID_CODE1	-	- ID (Read Only)						
0x10	ODIV	-	ODIV						
0x11	NINT	-	NINT						
0x12	NFRAC_H		NFRAC_H						
0x13	NFRAC_M		NFRAC_M						
0x14	NFRAC_L				NFR	AC_L			
0x15	PLL_CTRL0	OE_REG	-	-	-	-	NEW_FRE Q	SML_CHG	NVM_RES TORE
0x50	PLL_CTRL1	OE_REG	-	-	-	-	NEW_FRE Q	SML_CHG	NVM_RES TORE

Note: Please do not write values in the addresses that are not mentioned in this list. Please write 0 in the bit that is not defined.

## 7.2. Product Code 0 Register

Addroop	Register				В	Bit			
Address	name	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
0x00	P_CODE0				P_C	ODE			
-	Туре	R/O							
D	efault	0	1	0	0	0	1	1	0

Bit	Name	Function					
7:0	P_CODE	Product code (0x46) Ascii Code 'F'					

## 7.3. Product Code 1 Register

Address	Register	Bit							
Address	name	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
0x01	P_CODE1		P_CODE						
-	Туре		R/O						
Default		0	0	0	0	0	1	1	0

Bit	Name	Function
7:0	P_CODE	Product code (0x41) 0x06

# 7.4. Revision Code Register

Addroop	Register		Bit						
Address	name	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
0x02	REV		REV						
-	Туре				R	0			
D	efault	0	0	0	0	0	0	0	1

Bit	Name	Function
7:0	REV	Revision code 0x01

# 7.5. ID Code 0 Register

Addroop	Register				B	lit			
Address	name	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
0x03	ID_CODE0		ID						
Type R/O									
Default		0	0	0	0	0	0	0	1

	Bit	Name	Function
ſ	7:0	ID	ID code 0x01

# 7.6. ID Code 1 Register

Addroop	Register		Bit							
Address	name	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	
0x04	ID_CODE1	-	ID							
-	Type -			R/O						
D	efault	- Depend on the product								

Bit	Name	Function					
7	Reserved	Always read as 0.					
6:0	ID	ID code					

# 7.7. ODIV Register

Addroop	Register				В	Bit			
Address	name	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
0x10	ODIV	-	-	-	-	ODIV			
-	Туре	-	-	-	-	R/W			
D	efault	-	-	-	-	NVM			

Bit	Name		Function						
7:4	Reserved	Please write 0 at	Please write 0 at all the times.						
3:0	ODIV	Division ratio of	Division ratio of output divider						
		0x0: 4	0x4: 8	0x8: 16	0xC: 32				
		0x1: 5	0x5: 10	0x9: 20	0xD: 40				
		0x2: 6	0x2: 6 0x6: 12 0xA: 24 0xE: 48						
		0x3: 7	0x7: 14	0xB: 28	0xF: 56				

# 7.8. NINT Register

Address	Register				В	lit			
Address	name	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
0x11	NINT	-	-	NINT					
	Туре	-	R/W						
D	Default	-	-	NVM					

Bit	Name		Function					
7:6	Reserved	Please write 0 at all th	Please write 0 at all the times.					
5:0	NINT	Integer portion of the	nteger portion of the feedback divider (N <sub>INT</sub> )					
		Set	tting	Description				
		0x00 ~ 0x11,	0d ~ 17d	This setting shall not be configured				
		0x12	18d	N <sub>INT</sub> = 18				
		0x20	0x20 32d N <sub>INT</sub> = 32					
		0x21 ~ 0x3F	33d ~ 63d	This setting shall not be configured				

# 7.9. NFRAC Register

م <u>ما ما برم م</u>	Register	Bit							
Address	Name	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
0x12	NFRAC_H		NFRAC[23:16]						
0x13	NFRAC_M		NFRAC[15:8]						
0x14	NFRAC_L		NFRAC[7:0]						
	Type R/W								
C	Default	NVM							

Bit	Name	Function
7:0	NFRAC[23:16] NFRAC[15:8] NFRAC[7:0]	Fractional portion of the feedback divider (N <sub>FRAC</sub> ) E.g. Setting in case N <sub>FRAC</sub> is $0x123456$ NFRAC_H = $0x12$ NFRAC_M = $0x34$ NFRAC_L = $0x56$

# 7.10. PLL Control Register

Address	Register	Bit									
Address	Name	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0		
0x15 0x50	PLL_CTRL0 PLL_CTRL1	OE_REG	-	-	-	-	NEW_FR EQ	SML_CH G	NVM_RE STORE		
Туре		R/W	-	-	-	-	R/W	R/W	R/W		
Default		0	-	-	-	-	0	0	0		

PLL\_CTRL0 and PLL\_CTRL1 is an address shared register.

Bit	Name	Function					
7	OE_REG	Output enable register function LVPECL output buffer is enable when OE pin or this register is set as 1/High as shown below table.					
		LVPECL output buffer					
		OE pin (Active High) status OE pin (Active Low) status					
				H or Open	Ĺ	H )	L or Open
		OE_REG	1	Enable	Enable	Enable	Enable
		value	0	Enable	Disable	Disable	Enable
6:3	Reserved	Please write 0 at all the times.					
2	NEW_FREQ	and output freq change of the c	eque uenc outpu	ncy setting config cy is updated acc ut frequency and	gured in user reg cordingly. This bit PLL calibration is for details of freq	is automatically completed.	cleared once
1	SML_CHG	<ul> <li>New frequency applied (small change in frequency)</li> <li>By writing 1, frequency setting configured in user register is forwarded to PLL register and output frequency is updated accordingly. This bit is automatically cleared once change of the output frequency is done.</li> <li>Note: Please refer to the item 6.2.2 for details of frequency change by this bit.</li> </ul>					
0	NVM_RESTORE	Restore user register from NVM         By writing 1, default value of user register is restored from non-volatile memory (NVM). This bit is automatically cleared once the register restore is done.         Note:       PLL register is not updated only by writing to this bit. In order to initialize the user register and the PLL register (= output frequency) at the same time, please write 0x05 to PLL_CTRL register (NEW_FREQ bit and NVM_RESTORE bit is written as 1).					



# 8. Dimensions



# 9. Device Marking

• Standard (Factory preset start-up frequency product)



• Blank Sample (SG-Writer II programmable start-up frequency product)



The above marking layout shows only marking contents and their approximate position, not actual font, size and exact position.

# **10. Soldering Pattern**

Example of patterning design indicated as follows. In an actual design, please consider mounting density, the reliability of soldering, etc. and check whether performance is optimal.



# **11. Application Note**

- 1. This device contains a crystal resonator, so please do not expose to excessive shock or vibration. The internal crystal resonator might be damaged in case that too much shock or vibration is produced mechanically. Be sure to check your machine condition in advance.
- 2. This device is made with C-MOS IC. Please take necessary precautions to prevent damage due to electrostatic discharge.
- 3. We recommend to use and store under room temperature and normal humidity to secure frequency accuracy and prevent moisture.
- 4. We will announce the discontinuance and switch to our successor before six months or more.
- 5. Recommendation reflow times are less than 3 times.

When there was a soldering error, please do alteration with a soldering iron. In this case, the iron ahead is equal to or less than +350 °C and asks within 5 s.

In case that this device is reflow soldered on the back side of your circuit board, please carefully verify the device is properly secured to prevent coming detached from card.

Soldering method	Good or No good
Reflow soldering (top side)	Good
Reflow soldering (back side)	Please carefully verify the device is properly secured to prevent coming detached from card.
Solder pot (static solder pot/flow solder pot)	No good
Iron soldering	Good

#### Soldering method

- 6. Ultrasonic cleaning can be used on this product, however, since the oscillator might be damaged under some conditions, please exercise caution in advance.
- 7. Protection against periodically mechanical vibration

While there is any given shock or mechanical vibration periodically to crystal products, such as, a cooling fan, a piezo sounder, a piezo buzzer, and a speaker to crystal products, output frequency and amplitude can be changed. Especially the quality of telecommunication equipment could be affected by this phenomenon. Although Epson's crystal products are designed to minimize the effect of mechanical vibration, we recommend checking them in advance.

8. The metal part of the surface (metal cap) is connected to GND #3 pin. Please take necessary precautions to prevent short circuit to GND by contact with the metal cap.

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- 9. Side leads as shown below are connected to IC internally. Therefore be careful for short or a fall of insulation resistance.



- 10.  $V_{\text{CC}}$  and GND pattern shall be as large as possible so that high frequency impedance shall be small.
- 11. Seiko Epson doesn't recommend to power on from intermediate electric voltage or extreme fast power on. Those powering conditions may cause no oscillation or abnormal oscillation.
- 12. Please design the output lines by characteristic impedance  $50 \Omega$  and try to make the output lines as short as possible. A long output line may cause irregular output. Other high level signal lines may cause incorrect operation, so please do not place high-level signal line close to this device.
- 13. If OE (Active High), SDA or SCL pin is not used, please connect them to V<sub>CC</sub>. In order to suppress surge, resister may be used for OE pin.
- 14. If output pin is connected to the ground when supply voltage is applied to product, the internal elements can be destroyed. So please use the products that always have connection with load resistance.
- 15. As with any high speed analog circuitry, the power supply pins for SG-8506CA are vulnerable to noise. In order to achieve optimum jitter performance, the 0.1 μF and 10 μF capacitor as shown below is required. These capacitors should be placed as close to Vcc (#3 pin) as possible. It is also recommended that the capacitors are placed on the device side of the PCB. To achieve best performance, it is recommended to place the filter composing devices. Please see next page.



# SG-8506CA

#### ■ Example of SG-8506CA schematic layout

This figure shows an example of this product's application schematic.

As with any high speed analog circuitry, the power supply pins for SG-8506CA are vulnerable to noise. In order to achieve optimum jitter performance, power isolation with filter device is required for power supply pins.

In order to achieve best performance of the power isolation filter, it is recommended that the filter composing devices is placed on the device side of the PCB as close to the power pins as possible. The component value of this filter is just an example; it may have to be adjusted.







Please place them on the device side of the PCB as close to the power pins as possible

# **Application Manual**

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