



PI6ULS5V9306

## **Dual Bidirectional I2C-Bus and SMBus Voltage-Level Translator**

## **Features**

- 2-bit Bidirectional Translator for SDA and SCL Lines in • Mixed-Mode I2C-Bus Applications
- Standard-Mode, Fast-Mode, and Fast-Mode Plus I2C-Bus and **SMBus** Compatible
- Less than 1.5ns Maximum Propagation Delay to Accommodate Standard Mode and Fast Mode I2C-Bus Devices and Multiple Masters
- Allows Voltage Level Translation Between:
  - 0.9V VREF1 and 1.8V, 2.5V, 3.3V, or 5V VREF2
  - 1.2V VREF1 and 1.8V, 2.5V, 3.3V, or 5V VREF2
  - 1.5V VREF1 and 2.5V, 3.3V, or 5V VREF2
  - 1.8V VREF1 and 3.3V or 5V VREF2
  - 2.5V VREF1 and 5V VREF2
  - 3.3V VREF1 and 5V VREF2
- Provides Bidirectional Voltage Translation with no Direction Pin
- Low 3.5Ω ON-State Connection Between Input and Output Ports Provides Less Signal Distortion
- Open-Drain I2C-Bus I/O Ports (SCL1, SDA1, SCL2, and SDA2)
- 5V Tolerant I2C-Bus I/O Ports to Support Mixed-Mode Signal Operation
- High-Impedance SCL1, SDA1, SCL2, and SDA2 Pins for EN = LOW
- Lock-up Free Operation for Isolation when EN = LOW
- Flow through Pinout for Ease of Printed-Circuit Board Trace Routing
- ESD Protection Exceeds 4KV HBM per JESD22-A114
- Totally Lead-Free & Fully RoHS Compliant (Notes 1 & 2)
- Halogen and Antimony Free. "Green" Device (Note 3)
- For automotive applications requiring specific change control (i.e. parts gualified to AEC-Q100/101/104/200, PPAP capable, and manufactured in IATF 16949 certified facilities), please contact us or your local Diodes representative. https://www.diodes.com/quality/product-definitions/
- Package: TDFN2x3-8(ZE), MSOP-8(U), SOIC-8(W) •

## Description

The DIODES™ PI6ULS5V9306 is a dual bidirectional I<sup>2</sup>C-bus and SMBus voltage-level translator with an enable (EN) input. It is operational from 0.9V to 3.3V (VREF1) and 1.8V to 5V (VREF2).

The PI6ULS5V9306 allows bidirectional voltage translations between 1.0V and 5V without the use of a direction pin. The low ON-state resistance (Ron) of the switch allows connections to be made with minimal propagation delay. When EN is HIGH, the translator switch is on, and the SCL1 and SDA1 I/O are connected to the SCL2 and SDA2 I/O respectively, allowing bidirectional data flow between ports. When EN is LOW, the translator switch is off, and a high-impedance state exists between ports.

## **Block Diagram**



#### **Function Table**

EN	Function
Н	SCL1 = SCL2; SDA1 = SDA2
L	disabled

#### Notes:

2. See https://www.diodes.com/quality/lead-free/ for more information about Diodes Incorporated's definitions of Halogen- and Antimony-free, "Green" and Lead-free. 3. Halogen- and Antimony-free "Green" products are defined as those which contain <900ppm bromine. <900ppm chlorine (<1500ppm total Br + Cl) and <1000ppm antimony compounds.

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<sup>1.</sup> No purposely added lead. Fully EU Directive 2002/95/EC (RoHS), 2011/65/EU (RoHS 2) & 2015/863/EU (RoHS 3) compliant.





# **Pin Configuration**



# **Pin Description**

Pin#	Name	Description
1	GND	Ground (0V)
2	VREF1	Low-voltage side reference supply voltage for SCL1 and SDA1
3	SCL1	Serial clock, low-voltage side; connect to VREF1 through a pullup resistor
4	SDA1	Serial data, low-voltage side; connect to VREF1 through a pullup resistor
5	SDA2	Serial data, high-voltage side; connect to VREF2 through a pullup resistor
6	SCL2	Serial clock, high-voltage side; connect to VREF2 through a pullup resistor
7	VREF2	High-voltage side reference supply voltage for SCL2 and SDA2
8	EN	Switch enable input; connect to VREF2 and pullup through a high resistor





## **Maximum Ratings**

Storage Temperature	65°C to +150°C
Reference Voltage <sup>(2)</sup>	
Reference Bias Voltage	-0.5V to+6.0V
DC Input Voltage	-0.5V to +6.0V
Control Input Voltage (EN)	-0.5V to+6.0V
Channel Current (DC)	128mA
Input Clamping Current	50mA
ESD: HBM Mode	4000V
Junction Temperature under Bias (T <sub>J</sub> )	125°C

#### Note:

Stresses greater than those listed under MAXIMUM RATINGS may cause permanent damage to the device. This is a stress rating only and functional operation of the device at these or any other conditions above those indicated in the operational sections of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect reliability.

### **Recommended Operation Conditions**

Vcc = 2.7V to 5.5V;  $\hat{GND} = 0V$ ;  $T_A = -40^{\circ}C$  to  $+85^{\circ}C$ ; unless otherwise specified.

Symbol	Parameter	Test Conditions	Min.	Тур.	Max.	Unit
V <sub>I/O</sub>	Voltage on an Input/Output Pin	SCL1, SDA1, SCL2, SDA2	0		5	V
V <sub>REF1</sub>	Reference Voltage <sup>(1)</sup>	VREF1	0.9	_	3.3	V
V <sub>REF2</sub>	Reference Bias Voltage <sup>(2)</sup>	VREF2	1.8		5	V
V <sub>I(EN)</sub>	Input Voltage on Pin EN	_	0	_	5	V
I <sub>(pass)</sub>	Pass Switch Current	—	_		64	mA
T <sub>A</sub>	Ambient Temperature	—	-40	_	85	°C

### **DC Electrical Characteristics**

 $T_A = -40^{\circ}C$  to  $+85^{\circ}C$ ; unless otherwise specified.

Parameter	Description	Test Conditions <sup>(1)</sup>		Min	Typ. <sup>(2)</sup>	Max	Unit
Input and O	Dutput SDAB and SCLB						
V <sub>IK</sub>	Input Clamping Voltage	$I_{I}$ = -18mA; $V_{I}$	(EN) = 0V	_		-1.2	V
I <sub>IH</sub>	HIGH-Level Input Current	$V_I = 5V; V_{I(EN)}$	= 0V	_	_	5	μΑ
C <sub>i(EN)</sub>	Input Capacitance on pin EN	$V_I = 3V \text{ or } 0V$			11		pF
Cio(off)	Off-State Input/Output Capacitance (SCLn, SDAn)	$V_0 = 3V \text{ or } 0V; V_{I(EN)} = 0V$			4	_	pF
Cio(on)	On-State Input/Output Capacitance (SCLn, SDAn)	$V_0 = 3V \text{ or } 0V; V_{I(EN)} = 3V$			10.5		pF
			$V_{I(EN)} = 4.5V$	—	3.5	5.5	Ω
		$V_{I} = 0V;$	$V_{I(EN)} = 3V$	_	4.7	7.0	Ω
		$I_0 = 64 \text{mA}$	$V_{I(EN)} = 2.3V$		6.3	9.5	Ω
Ron	ON-State Resistance <sup><math>(2)</math></sup>		$V_{I(EN)} = 1.5V$		60	140	Ω
	(SCLn, SDAn)	$V_{I} = 2.4V;$	$V_{I(EN)} = 4.5V$	1	6	15	Ω
		$I_0 = 15 \text{mÅ}$	$V_{I(EN)} = 3V$	20	60	140	Ω
		$V_{I} = 1.7V;$ $I_{O} = 15mA$	$V_{I(EN)} = 2.3V$	20	60	140	Ω

1. All typical values are at  $T_A = 25^{\circ}C$ .

2. Measured by the voltage drop between the SCL1 and SCL2 or SDA1 and SDA2 terminals at the indicated current through the switch. ON-state resistance is determined by the lowest voltage of the two terminals.





## **Dynamic Characteristics**

 $T_A = -40^{\circ}$ C to +85°C; unless otherwise specified. Values guaranteed by design.

C	Description	Carlttere	$C_L = 50 pF$		CL =	30pF	$C_L = 15 pF$		Unit
Symbol Parameter		Conditions	Min Max		Min	Max	Min	Max	Unit
Dynamic	Characteristics (Trans	slating Down)							
$V_{I(EN)} = 3$	$.3V; V_{IH} = 3.3V; V_{IL} =$	$= 0V; V_{\rm M} = 1.15V$							
t <sub>PLH</sub>	LOW-to-HIGH Propagation Delay	From (Input) SCL2 or SDA2 to (Output) SCL1 or SDA1	0	0.8	0	0.6	0	0.3	ns
t <sub>PHL</sub>	HIGH-to-LOW Propagation Delay	From (Input) SCL2 or SDA2 to (Output) SCL1 or SDA1	0	1.2	0	1	0	0.5	ns
$V_{I(EN)} = 2$	$.5V; V_{IH} = 2.5V; V_{IL} =$	$= 0V; V_{\rm M} = 0.75V$							
t <sub>PLH</sub>	LOW-to-HIGH Propagation Delay	From (Input) SCL2 or SDA2 to (Output) SCL1 or SDA1	0	1	0	0.7	0	0.4	ns
t <sub>PHL</sub>	HIGH-to-LOW Propagation Delay	From (Input) SCL2 or SDA2 to (Output) SCL1 or SDA1	0	1.3	0	1	0	0.6	ns
Dynam	ic Characteristics (Tra	nslating up)							
$V_{I(EN)} = 3$	$.3V; V_{IH} = 2.3V; V_{IL} =$	= 0V; $V_T$ = 3.3V; $V_M$ = 1.15V; $R_L$	= 300Ω						
t <sub>PLH</sub>	LOW-to-HIGH Propagation Delay	From (Input) SCL1 orSDA1 to (output) SCL2 or SDA2	0	0.9	0	0.6	0	0.4	ns
t <sub>PHL</sub>	HIGH-to-LOW Propagation Delay	From (Input) SCL1 or SDA1 to (Output) SCL2 or SDA2	0	1.4	0	1.1	0	0.7	ns
$V_{I(EN)} = 2$	$.5V; V_{IH} = 1.5V; V_{IL} =$	= 0V; $V_T = 2.5V$ ; $V_M = 0.75V$ ; $R_L$	= 300Ω						
t <sub>PLH</sub>	LOW-to-HIGH Propagation Delay	From (Input) SCL1 orSDA1 to (Output) SCL2 or SDA2	0	1	0	0.6	0	0.4	ns
t <sub>PHL</sub>	HIGH-to-LOW Propagation Delay	From (Input) SCL1 or SDA1 to (Output) SCL2 or SDA2	0	1.3	0	1.3	0	0.8	ns







NOTES: A. CL includes probe and jig capacitance.

B. All input pulses are supplied by generators having the following characteristics: PRR  $\leq$  10 MHz, Z<sub>O</sub> = 50 Ω, t<sub>f</sub>  $\leq$  2 ns, t<sub>f</sub>  $\leq$  2 ns. C. The outputs are measured one at a time, with one transition per measurement.

Figure 1. Load Circuit for Outputs

## **Functional Description**

The PI6ULS5V9306 can also be used to run two buses—one at 400kHz operating frequency and the other at 100kHz operating frequency. If the two buses are operating at different frequencies, the 100kHz bus must be isolated when the 400kHz operation of the other bus is required. If the master is running at 400kHz, the maximum system operating frequency may be less than 400kHz because of the delays added by the translator.

As with the standard I<sup>2</sup>C-bus system, pullup resistors are required to provide the logic HIGH levels on the translator's bus. The PI6ULS5V9306 has a standard open-collector configuration of the I<sup>2</sup>C-bus. Each side of the translator must have a pullup resistor though the size of these pullup resistors depends on the system. The device is designed to work with standard mode, fast mode, and fast mode plus I<sup>2</sup>C-bus devices in addition to SMBus devices.

When the SDA1 or SDA2 port is LOW, the clamp is in the ON-state and a low-resistance connection exists between the SDA1 and SDA2 ports. When the higher voltage is on the SDA2 port, and the SDA2 port is HIGH, the voltage on the SDA1 port is limited to the voltage set by VREF1. When the SDA1 port is HIGH, the SDA2 port is pulled to the drain pullup supply voltage (VDPU) by the pullup resistors. This functionality allows a seamless translation between higher and lower voltages selected by the user without requiring directional control. The SCL1/SCL2 channel also functions as the SDA1/SDA2 channel.

All channels have the same electrical characteristics, and there is minimal deviation from one output to another in voltage or propagation delay. This is a benefit over discrete transistor voltage translation solutions because the fabrication of the switch is symmetrical. The translator provides excellent ESD protection to lower voltage devices, and at the same time protects less ESD-resistant devices





## **Application Information**





## **Open-Drain Application**

For the bidirectional clamping configuration (higher voltage to lower voltage or lower voltage to higher voltage), the EN input must be connected to VREF2 and both pins pulled to high-side VDPU through a pullup resistor (typically 200k $\Omega$ ). This allows VREF2 to regulate the EN input. A filter capacitor on VREF2 is recommended.







Figure 4. Typical Push-Pull Application Circuit (Switch Enabled Control)

## **Push-Pull Application**

If used in push-pull system, the pullup resistors on REF side are also required. The data must be unidirectional, or the outputs must be 3-stateable and controlled by some direction-control mechanism to prevent high-to-low contentions in either direction.

## **Operating Voltage**

Refer to Figure 2

Symbol	Description	Min	Typ <sup>(1)</sup>	Max	Unit
VDPU	Ref2 Side Pullup Voltage on 200kΩ	VREF1 + 0.6	2.1	5	V
EN	Enable Input Voltage	VREF1 + 0.6	2.1	5	V
VREF1	Reference Voltage	0	1.5	4.4	V
IPASS	Pass Switch Current	14			mA
IREF	EF Reference-Transistor Current			5	μΑ
ТА	Operating Free-Air Temperature	-40	_	85	°C

## The Pass-Through Current: I\_pass

I\_pass is determined by the pullup and the low voltage added on the PI6LS5V9306.

In Figure 6, I\_pass equals (V<sub>REF1</sub>-V<sub>OL1\_9306</sub>)/R<sub>PU1</sub>.

When V\_IN is 0V, the PI6ULS5V9306 can support as large as 64mA pass-through current in theory, but it is recommend to limit the I\_pass in 15mA.







Figure 5. Typical Open-Drain Application Circuit

(1) The Sink Current: I\_sink

The device sinks the total current from both pullup resistors. For example, in figure below, when the SDA2 is pulled low by the I2C device, the sink current of the I2C device is  $I_sink = Ipass+I_2=I_1+I_2$ . The same thing happens when I2C master pulls low the I2C bus. The I\_sink should not be larger than the tolerance of the I2C devices.

## (2) $V_{\rm IL}\!,\!V_{\rm OL}$ of the External Drive and $V_{\rm OL}$ of PI6ULS5V9306

In normal application, the  $V_{IL}$  of external devices should always be larger than the  $V_{OL}$  of PI6ULS5V9306. The value of PI6ULS5V9306's  $V_{OL}$  is determined by the pass-through current and the low voltage added on the SDA, SCL pins. The  $V_{OL_{9306}} = V_{IN_{L}} + V_{UP}$  ( $V_{UP}$  is mainly determined by the I\_pass, which is always less than 0.35V).

## (3) Low VREF Application

The PI6ULS5V9306 can support very-low Vref1 application in theory, but it is recommended no lower than 0.9V. Because when VREF1 is less than 1.8V, the  $V_{OL}$  of REF1 side is a concern in system. For example, in Figure 6, if VREF1 = 0.9V, VDPU = 3.3V, the  $V_{IL}$  of the REF1 side I2C master is normally 0.3 × VREF1 = 0.25V, but the  $V_{OL}$  of REF2 side can up to 0.1 × VDPU=0.36V. The system designer must make sure this situation does not happen. A limit for the  $V_{OL}$  of REF2 side devices is required then.

The following table shows the requirements for  $V_{OL}$  of VREF2 side devices when using PI6ULS5V9306. Figure 6 shows the requirement for  $V_{OL\_DEVICE}$ .

The V <sub>OL</sub> Requirement of V <sub>REF2</sub> Side External Devices (Temp = 25°C, Assume the V <sub>IL</sub> of V <sub>REF1</sub> Side Devices is $0.3 \times V_{REF1}$ )						
$\begin{array}{c c c c c c c c c c c c c c c c c c c $						
0.9V	≤0.15V	≤0.1V	Not Recommended			
1.2V	≤0.2V	≤0.15V	Not Recommended			
1.5V	≤0.3V	≤0.25V	≤0.2V			
1.8V	≤0.4V	≤0.35V	≤0.3V			





## **Pullup Resistors and Minimum Values**

Sizing the pullup resistor on an open-drain bus is specific to the individual application and is dependent on the following driver characteristics:

- The driver sink current
- The V<sub>OL</sub> of driver
- The  $V_{OL}$  of the PI6ULS5V9306
- The V<sub>IL</sub> of the driver
- Frequency of operation

The following tables can be used to estimate the pullup resistor value in different use cases so that the minimum resistance for the pullup resistor can be found.

The tables below show suggested minimum values of pullup resistors for the PI6ULS5V9306 with typical voltage translation levels and drive currents. The calculated values assume that both drive currents are the same.

 $V_{OL} = V_{IL} = 0.1 \times VCC$  and accounts for a 5% VCC tolerance of the supplies, 1 % resistor values. Note that the resistor chosen in the final application should be equal to or larger than the values shown in the table to ensure that the pass voltage is less than 10% of the VCC voltage, and the external driver should be able to sink the total current from both pullup resistors.

A Side			B Side		
A Side	1.5V 1.8V		2.5V	3.3V	5.0V
0.9V	$R_{RPU1} = 859\Omega$ $R_{RPU2} = 859\Omega$	$\begin{aligned} R_{RPU1} &= 970\Omega \\ R_{RPU2} &= 970\Omega \end{aligned}$	$R_{RPU1} = none$ $R_{RPU2} = 896\Omega$ or both 1.23kΩ	$R_{RPU1} = none$ $R_{RPU2} = 1.19k\Omega$ or both 1.53kΩ	$R_{RPU1} = none$ $R_{RPU2} = 1.82k\Omega$ or both 2.16kΩ
1.2V	_	$\begin{aligned} R_{RPU1} &= 1.07 k\Omega \\ R_{RPU2} &= 1.07 k\Omega \end{aligned}$	$R_{RPU1} = none$ $R_{RPU2} = 886\Omega$ or both 1.33kΩ	$R_{RPU1} = none$ $R_{RPU2} = 1.18k\Omega$ or both 1.63kΩ	$R_{RPU1} = none$ $R_{RPU2} = 1.81k\Omega$ or both 2.26kΩ
1.5V	_	_	$R_{RPU1} = none$ $R_{RPU2} = 875\Omega$ or both 1.43kΩ	$R_{RPU1} = none$ $R_{RPU2} = 1.17k\Omega$ or both 1.73kΩ	$R_{RPU1} = none$ $R_{RPU2} = 1.8k\Omega$ or both 2.36kΩ
1.8V	_	_	$\begin{aligned} R_{RPU1} &= 1.53 k\Omega \\ R_{RPU2} &= 1.53 k\Omega \end{aligned}$	$R_{RPU1} = none$ $R_{RPU2} = 1.16k\Omega$ or both 1.82kΩ	$\begin{aligned} R_{RPU1} &= none \\ R_{RPU2} &= 1.79 k\Omega \\ or both 2.46 k\Omega \end{aligned}$
2.5V				$\begin{aligned} R_{RPU1} &= 2.06 k\Omega \\ R_{RPU2} &= 2.06 k\Omega \end{aligned}$	$\begin{split} R_{RPU1} &= none \\ R_{RPU2} &= 1.77 k\Omega \\ or both 2.69 k\Omega \end{split}$
3.3V	_	_	_	_	$\begin{split} R_{RPU1} &= none \\ R_{RPU2} &= 1.74 k\Omega \\ or both 2.96 k\Omega \end{split}$

Pullup Resistor Minimum Values, 3mA Driver Sink Current for PI6ULS5V9306





A Side			B Side		
A Slue	1.5V	1.8V	2.5V	3.3V	5.0V
0.9V	$\begin{aligned} R_{RPU1} &= 258\Omega \\ R_{RPU2} &= 258\Omega \end{aligned}$	$R_{RPU1} = 291\Omega$ $R_{RPU2} = 291\Omega$	$\begin{array}{c} R_{RPU1} = none \\ R_{RPU2} = 269\Omega \\ or both 369\Omega \end{array}$	$R_{RPU1} = none$ $R_{RPU2} = 358\Omega$ or both 458\Omega	$R_{RPU1} = none$ $R_{RPU2} = 546\Omega$ or both 646\Omega
1.2V	_	$R_{RPU1} = 321\Omega$ $R_{RPU2} = 321\Omega$	$R_{RPU1} = none$ $R_{RPU2} = 266\Omega$ or both 399 $\Omega$	$R_{RPU1} = none$ $R_{RPU2} = 355\Omega$ or both 488\Omega	$R_{RPU1} = none$ $R_{RPU2} = 543\Omega$ or both 677\Omega
1.5V	_	_	$\begin{array}{c} R_{RPU1} = none \\ R_{RPU2} = 263\Omega \\ or both 429\Omega \end{array}$	$R_{RPU1} = none$ $R_{RPU2} = 352\Omega$ or both 518\Omega	$R_{RPU1} = none$ $R_{RPU2} = 540\Omega$ or both 707\Omega
1.8V	_	_	$\begin{array}{c} R_{RPU1} = 460\Omega \\ R_{RPU2} = 460\Omega \end{array}$	$R_{RPU1} = none$ $R_{RPU2} = 348\Omega$ or both 548\Omega	$R_{RPU1} = none$ $R_{RPU2} = 537\Omega$ or both 737\Omega
2.5V	_	_	_	$\begin{aligned} R_{RPU1} &= 619\Omega \\ R_{RPU2} &= 619\Omega \end{aligned}$	$R_{RPU1} = none$ $R_{RPU2} = 521\Omega$ or both 808Ω
3.3V	_	_	_	_	$R_{RPU1} = none$ $R_{RPU2} = 522\Omega$ or both 889\Omega

## Pullup Resistor Minimum Values, 10mA Driver Sink Current for PI6ULS5V9306

Pullup Resistor Minimum Values, 15mA Driver Sink Current for PI6ULS5V9306

A Side			B Side		
A Slue	1.5V	1.8V	2.5V	3.3V	5.0V
0.9V	$\begin{aligned} R_{RPU1} &= 172\Omega \\ R_{RPU2} &= 172\Omega \end{aligned}$	$R_{RPU1} = 194\Omega$ $R_{RPU2} = 194\Omega$	$R_{RPU1} = none$ $R_{RPU2} = 179\Omega$ or both 246Ω	$R_{RPU1} = none$ $R_{RPU2} = 238\Omega$ or both 305\Omega	$R_{RPU1} = none$ $R_{RPU2} = 364\Omega$ or both 431Ω
1.2V	_	$R_{RPU1} = 214\Omega$ $R_{RPU2} = 214\Omega$	$R_{RPU1} = none$ $R_{RPU2} = 177\Omega$ or both 266Ω	$R_{RPU1} = none$ $R_{RPU2} = 236\Omega$ or both 325\Omega	$R_{RPU1} = none$ $R_{RPU2} = 362\Omega$ or both 451Ω
1.5V	_	_	$R_{RPU1} = none$ $R_{RPU2} = 175\Omega$ or both 286\Omega	$R_{RPU1} = none$ $R_{RPU2} = 234\Omega$ or both 345\Omega	$R_{RPU1} = none$ $R_{RPU2} = 360\Omega$ or both 471\Omega
1.8V	_	_	$\begin{array}{l} R_{RPU1} = 306\Omega \\ R_{RPU2} = 306\Omega \end{array}$	$R_{RPU1} = none$ $R_{RPU2} = 232\Omega$ or both 366\Omega	$R_{RPU1} = none$ $R_{RPU2} = 358\Omega$ or both 492 $\Omega$
2.5V	_	_	_	$\begin{array}{l} R_{RPU1} = 413\Omega \\ R_{RPU2} = 413\Omega \end{array}$	$R_{RPU1} = none$ $R_{RPU2} = 354\Omega$ or both 539\Omega
3.3V	_	_	_	—	$R_{RPU1} = \text{none}$ $R_{RPU2} = 348\Omega$ or both 593\Omega





## **Maximum Frequency Application**

The maximum frequency is limited by the minimum pulse width LOW and HIGH as well as rise time and fall time.



The rise and fall times are dependent upon translation voltages, the drive strength, the total node capacitance (CL), and the pullup resistors (RPU) that are present on the bus. The node capacitance is the addition of the PCB trace capacitance and the device capacitance that exists on the bus.

Because of the dependency of the external components, PCB layout and the different device operating states the calculation of rise and fall times is complex and has several inflection points along the curve.

The main component of the rise and fall times is the RC time constant of the bus line when the device is in its two primary operating states: when device is in the ON state and it is low-impedance and when the device is OFF isolating the A-side from the B-side.

There are some basic guidelines to follow that will help maximize the performance of the device:

- Keep trace length to a minimum by placing the PI6ULS5V9306 close to the processor.
- The signal round trip time on trace should be shorter than the rise or fall time of signal to reduce reflections.
- The faster the edge of the signal, the higher the chance for ringing.
- The higher drive strength controlled by the pullup resistor (up to 15mA), the higher the frequency the device can use.

The system designer must design the pullup resistor value based on external current drive strength and limit the node capacitance (minimize the wire, stub, connector, and trace length) to get the desired operation frequency result.





## **Part Marking**

W Package Cu



Z: Die Rev Y: Date Code (Year) W: Date Code (Workweek) 1st X: Assembly Site Code 2nd X: Wafer Fab Site Code Bar above fab code means Cu wire

Note: Bar above "I" means Fab3 og MGN

ZE Package



J: Assembly Site Code G: Fab Site Code XX: Date Code (Year & Workweek) Bar above "A" means Fab3 of MGN W Package Au



Z: Die Rev AB: Date Code (Year & Workweek) K: Assembly Site Code G: Wafer Fab Site Code U Package



Z: Die Rev Y: Date Code (Year) W: Date Code (Workweek) 1st X: Assembly Site Code 2nd X: Wafer Fab Site Code Bar above fab code means Cu wire





# **Packaging Mechanical**

## TDFN-8 (ZE)



13-0155





## Recommended Land Pattern for TDFN2x3-8L







MSOP-8 (U)



16-0242





SOIC-8 (W)



#### For latest package information:

See http://www.diodes.com/design/support/packaging/pericom-packaging/packaging-mechanicals-and-thermal-characteristics/.

### **Ordering Information**

Part Number	Package Code	Package Description
PI6ULS5V9306ZEEX	ZE	8-Pin, 2X3 (TDFN)
PI6ULS5V9306UEX	U	8-Pin, Mini Small Outline Package (MSOP)
PI6ULS5V9306WEX	W	8-Pin,150 mil Wide (SOIC)

Notes:

No purposely added lead. Fully EU Directive 2002/95/EC (RoHS), 2011/65/EU (RoHS 2) & 2015/863/EU (RoHS 3) compliant. 1.

See https://www.diodes.com/quality/lead-free/ for more information about Diodes Incorporated's definitions of Halogen- and Antimony-free, "Green" and 2 Lead-free.

Halogen- and Antimony-free "Green" products are defined as those which contain <900ppm bromine, <900ppm chlorine (<1500ppm total Br + Cl) and 3 <1000ppm antimony compounds.

- E = Pb-free and Green
- X suffix = Tape/Reel 5





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