



3.3V 10/100BaseTX/FX MII Physical Layer Transceiver

#### Rev 3.11

### **General Description**

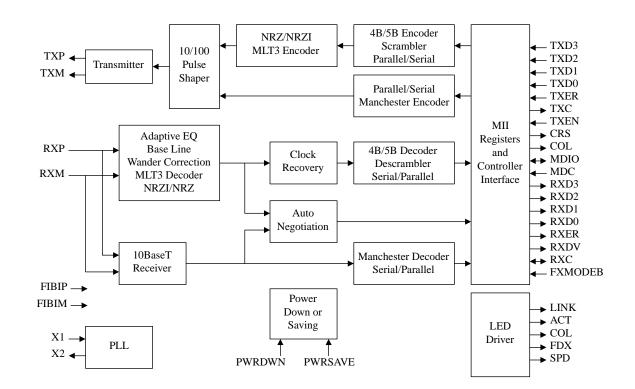
Operating at 3.3 Volts to meet low voltage and low power requirement, the KS8737 is a 10/100BaseTX/FX Physical Layer Transceiver which provides MII interface to transmit and receive data. It contains the 100BaseTX/FX Physical Medium Attachment (PMA), Physical Medium Dependent (PMD), and Physical Coding Sub-layer (PCS) functions. Moreover, the KS8737 has on-chip 10BaseT encoder/decoder and output filtering, which eliminates the need for external filters and makes possible a single set of line magnetics to be used to meet requirement for both 100BaseTX/ FX and 10BaseT.

The KS8737 can automatically configure itself for 100 or 10 Mbps and full or half duplex operation, using on-chip Auto-Negotiation algorithm. It's an ideal choice of physical layer transceiver for 100BaseTX/100BaseFX/10BaseT applications.

Data sheets and support documentation can be found on Micrel's web site at www.micrel.com.

#### Features

- Single chip 100BaseTX/100BaseFX/10BaseT physical layer solution
- 3.3V CMOS design, 70mA operating current (excluding transmit output driver current)
- Fully compliant to IEEE 802.3u standard
- Support Media Independent Interface (MII) mode
- Support 10BaseT, 100BaseTX and 100BaseFX Fiber Channel with Far\_End\_Fault Detection
- Support power down mode and power saving mode
- Configurable through MII serial management ports or via external control pins
- Support auto-negotiation and manual selection for 10Mbps or 100Mbps speed
- Support auto-negotiation and manual selection for fulland half-duplex mode
- Standard CSMA/CD or full-duplex operation at 10Mbps or 100Mbps
- On-chip built-in filtering for both 100BaseTX and 10BaseT



### Functional Diagram

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## **Features (continued)**

- LED outputs for link, activity, full/half duplex, collision and speed
- Supports back to back FX to TX for media converter applications
- Available in 64-pin TQFP surface mount package (10 mm  $\times$  10 mm  $\times$  1.0 mm)

# **Ordering Information**

Part Number	Temperature Range	Package
KS8737	0°C to +70°C	64-Pin TQFP

# **Revision History**

Revision	Date	Summary of Changes
3.0	7/01/02	Update to company logo and format. Add new feature on pin 33(DISTX/LPBK); disable the transmit only during media converter mode and select the loopback mode with TST2 pin. Change RXC type from I/O to O. Change the Register 1fh.9 to reserved. Change the Register 1fh.5 mode from RW to RO. Update on the 10/100BT MII receiving timing. Change on register 1fh.1 to reserved. Add the fiber mode description.
3.1	4/01/03	Change the company logo, legal disclaimer, contact info.
3.11	8/29/03	Convert to new format.

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# Pin Description

Pin Number	Pin Name	Type <sup>(Note 1)</sup>	Pin Function		
1	CRS	0	MII Carrier Sense	Dutput. Ac	tive High. High impedance when PHY is isolated.
2	INTRPT				requires an external 10k pull-up resistor. This pin y testing in the test mode.
3	RXENB	I/O	MII Receive Enable Input. Active Low. If this pin is High, the Receive output of MII (RXD[3:0], RXDV, RXER, RXC) will be in tri-state. This pin becomes an I/O pin for factory testing in the test mode.		
4	PHYAD4	I/O	PHY Address Bit [4 test mode.	l] Input. T	nis pin becomes an I/O pin for factory testing in the
5	PHYAD3	I/O	PHY Address Bit [3 test mode.	3] Input. T	nis pin becomes an I/O pin for factory testing in the
6	PHYAD2	I/O	PHY Address Bit [2 test mode.	2] Input. T	nis pin becomes an I/O pin for factory testing in the
7	PHYAD1	I/O	PHY Address Bit [1 test mode.	] Input. T	his pin becomes an I/O pin for factory testing in the
8	PHYAD0	I/O	PHY Address Bit [0 test mode.	)] Input. T	nis pin becomes an I/O pin for factory testing in the
9, 19, 24, 37, 53	VDD	Р	3.3V power supply		
10, 22, 26, 31, 43, 52, 63	GND	GND	Ground.		
11	X2	0	Crystal Oscillator Output. This pin is connected to the other terminal of the 25MHz crystal. If X1 is driven by an external clock, X2 must be left open.		
12	X1	I	Crystal Oscillator Input. Input for a crystal or an external 25MHz clock.		
13	FDX	I/O	Full-Duplex Input. If this pin is High, it sets full-duplex operation. If this pin is Low, it sets half duplex operation. The input signal of this pin is latched at reset. After reset, this pin becomes test pin for factory test.		
14 15	MODE0 MODE1	I/O	Mode Select Input. These pins carry encoded input signals that are latched at reset and power up to set mode of operation. After reset, they become test pins for factory test. These pins are I/O pins in the test mode.		
16	RSTB*	I	Hardware Reset In Internal 100kΩ pull		e Low signal. It forces the device to a known state.
17,35, 36	NC		No Connect.		
33	DISTX/LPBK	I	converter mode. Fl	oating is fo	pin is high, it disables transmit only during the Media or normal operation. The DISTX/LPBK pin also other with the TST2 pin.
			LPBK/DISTX	TST2	
			High	High	Disable Transmit
			High	Float	Local Loopback
			Low	High	Remote Loopback
			Low	Float	Remote Loopback
18	FXSD	I	Fiber Signal Detect	t. To dete	t fiber signal. Left open when not in use.
20	PWRSAVE/ FXSD_THD	I	Power Saving Mode Initialization Input. (Affecting Register 1f.15). To disable power saving mode, tie this pin low; otherwise, power saving mode is asserted. This pin can also be used to set FX signal detect threshold in fiber mode.		

**Note 1.** P = power supply

G = ground

I = input

O = output

I/O = bi-directional

Pin Number	Pin Name	Type <sup>(Note 1)</sup>	Pin Function	
21 23	TXP TXM	0	Twisted Pair Transmit Outputs. Differential transmit outputs for 100BaseTX or 10BaseT to magnetic.	
25	ISET	0	Transmit Current Set. Connecting an external reference resistor to set transmitter output current. This pin connected to a 22.1k $\Omega$ 1% resistor to ground if a transformer of 1:1 turns ratio is used.	
27 28	FIBIP FIBIM	I	Fiber Receive Inputs. Differential pseudo-ECL receive pairs compatible with standard fiber transceiver for 100BaseFX. Both pins should be tied to ground if not used or if not in the FX mode.	
29 30	RXP RXM	I	Twisted Pair Receive Input. Differential receive input pins for 100BaseTX or 10BaseT from the magnetics.	
32	TST2	I	Test Pin. During normal operation this pin should be left open. When tied high through a 1k resistor the chip will operate in back to back TX to FX mode. In this case, TXC becomes an input pin.	
34	PWRDWN	I	Power Down Select Input. When this pin is tied high, the chip is in power down mode. When this pin is open or tied low, the chip is in normal operation	
38	LEDSPD	I/O	LED Output. During normal operation, this pin lights the SPEED LED to indicate 100Mbps is selected. This pin becomes an I/O pin for factory testing in the test mode. Active Low.	
39	LEDCOL	I/O	LED Output. During normal operation, this pin lights the COL LED to indicate a collision. It will flash at a rate of 50ms high and 50ms low when active. This pin becomes an I/O pin for factory testing in the test mode. Active Low.	
40	LEDLINK	I/O	LED Output. During normal operation, this pin lights the LINK LED to indicate a good link is detected. This pin becomes an I/O pin for factory testing in the test mode. Active Low.	
41	LEDACT	I/O	LED Output. During normal operation, this pin lights the Activity LED when transmitting or receiving. It will flash at a rate of 50ms high and 50ms low when active. This pin becomes an I/O pin for factory testing in the test mode. Active Low.	
42	LEDFDX	I/O	LED Output. During normal operation, this pin lights the FDX LED to indicate a full-duplex mode. This pin becomes an I/O pin for factory testing in the test mode. Active Low.	
44	MDIO	I/O	Serial Management Data Input/Output. This pin requires an external 10k pull-up resistor.	
45	MDC	I	Serial Management Interface Clock Input. This pin is synchronous to the MDIO data interface.	
46	FXMODEB	I	FX Mode Select Input. Active Low. When this pin is low, the KS8737 is in the 100BaseFX mode.	
47 48 49 50	RXD3 RXD2 RXD1 RXD0	0	MII Receive Data Output. Active High, clocked out on the falling edge of RXCLK. RXD0 is the LSB. High impedance when PHY is isolated or if RXEN is deasserted.	
51	RXDV/ CRSDV	0	MII Receive Data Valid Output. Active High, clocked out on the falling edge of RXCLK. This signal indicates that recovered and decoded data nibbles are being presented synchronously to RXCLK. High impedance when PHY is isolated or if RXEN is de-asserted.	
54	RXC	0	MII Receive Clock Output. 25MHz in 100BaseTX mode, 2.5MHz in 10BaseT nibble mode. High impedance when PHY is isolated or if RXEN is de-asserted.	

**Note 1.** P = power supply

- G = ground
- I = input

O = output

I/O = bi-directional

Pin Number	Pin Name	Type <sup>(Note 1)</sup>	Pin Function	
55	RXER	0	MII Receive Error Output. Driven High synchronously on the falling edge of RXCLK when invalid symbol has been detected in 100BaseTX mode. This pin is ignored in 10BaseT operation. High impedance when PHY is isolated	
56	TXER	I	MII Transmit Error Input. A High on this pin causes the 4B/5B encode process to substitute the Transmit Error code-group for the encoded data word. This pin is ignored in a 10BaseT operation. When TXER is not used, this pin should be tied Low through a $10k\Omega$ resistor.	
57	TXC	I/O	MII Transmit Clock Output / Back to Back Mode Clock Input. During normal operation TXC is an output pin. It provides 25MHz in 100BaseTX mode, 2.5MHz in 10BaseT nibble mode. In back to back mode it becomes an input pin. High impedance when PHY is isolated.	
58	TXEN	I/O	MII Transmit Enable Input. A High on this pin causes the transmit data TXD[3:0] to be encoded and scrambled for transmission.	
59 60 61 62	TXD0 TXD1 TXD2 TXD3	I	MII Transmit Data Input. TXD0 is the LSB. High impedance when PHY is isolated.	
64	COL	0	MII Collision Detect Output. Active High. High impedance when PHY is isolated. This signal is de-asserted in full-duplex operation.	

**Note 1.** P = power supply

G = ground

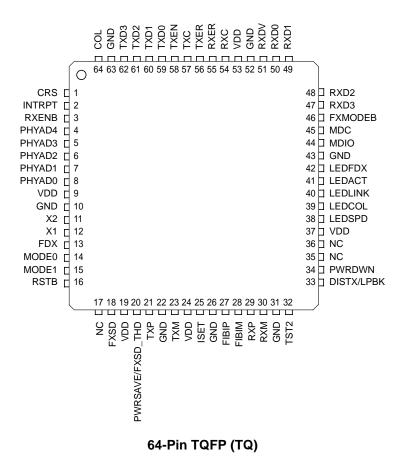
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## **Pin Configuration**



## **Functional Description**

### 100BaseTX Transmit

The 100BaseTX transmit function performs parallel to serial conversion, NRZ to NRZI conversion, MLT-3 encoding and transmission. The circuit starts with a parallel to serial conversion, which converts the 25 MHz, 4-bit nibbles into a 125 MHz serial bit stream. The incoming data is clocked in at the positive edge of the TXC signal. The serialized data is further converted from NRZ to NRZI format, then transmitted in MLT3 current output. The output current is set by an external 1% 22.1k $\Omega$  resistor for the 1: 1 transformer ratio. It has a typical rise/fall times of 4 ns and is complied to the ANSI TP-PMD standard regarding amplitude balance, overshoot and timing jitters. The wave-shaped 10BaseT output driver is also incorporated into the 100BaseTX driver, and the total output capacitance is typical 7pF with short PC board traces assumed.

### 100BaseTX Receive

The 100BaseTX receive function performs adaptive equalization, DC restoration, MLT-3 to NRZI conversion, data and clock recovery, NRZI to NRZ conversion, and serial to parallel conversion. The receiving side starts with the equalization filter to compensate inter-symbol interference (ISI) over the twisted pair cable. Since the amplitude loss and phase distortion is a function of the length of the cable, the equalizer has to adjust its characteristic to optimize the performance. In this design, the variable equalizer will make an initial estimation based on comparisons of incoming signal strength against some known cable characteristics, then tunes itself for optimization. This is an ongoing process and can self adjust against the environmental changes such as temperature variations.

The equalized signal then goes through a DC restoration and data conversion block. The DC restoration circuit is used to compensate effect of base line wander and improve the dynamic range. The differential data conversion circuit converts the MLT3 format back to NRZI. The slicing threshold is also adaptive.

The clock recovery circuit extracts the 125MHz clock from the edges of the NRZI signal. This recovered clock is then used to convert the NRZI signal into the NRZ format. Finally, the NRZ serial data is converted to 4-bit parallel 4B nibbles. A synchronized 25MHz RXC is generated so that the 4B nibbles is clocked out at the negative edge of RCK25 and is valid for the receiver at the positive edge. When no valid data is present, the clock recovery circuit is locked to the 25MHz reference clock and both TXC and RXC clocks continue to run.

### PLL Clock Synthesizer

The KS8737 generates 125MHz, 25MHz and 20MHz clocks for system timing. An internal crystal oscillator circuit provides the reference clock for the synthesizer.

#### Scrambler/De-scrambler (100BaseTX only)

The purpose of the scrambler is to spread the power spectrum of the signal in order to reduce EMI and baseline wander. The KS8737 provides a scrambler-bypass mode for testing purpose. Bypassing the scrambler causes the PCS-layer encoder to be bypassed such that the MII is operated in the 5B mode.

#### 10BaseT Transmit

When TXEN (transmit enable) goes high, data encoding and transmission will begin. The KS8737 will continue to encode and transmit data as long as TXEN remains high. The data transmission will end when TXEN goes low. The last transition occurs at the boundary of the bit cell if the last bit is zero, or at the center of the bit cell if the last bit is one. The output driver is incorporated into the 100Base driver to allow transmission with the same magnetic. They are internally wave-shaped and preemphasized into outputs with typical 2.5V amplitude. The harmonic contents are at least 27dB below the fundamental when driven by an all-ones Manchester-encoded signal.

#### **10BaseT Receive**

On the receive side, input buffer and level detecting squelch circuit are employed. A differential input receiver circuit and a PLL performs the decoding function. The Manchester-encoded data stream is separated into clock signal and NRZ data. A squelch circuit rejects signals with levels less than 300mV or with short pulse widths in order to prevent noises at the RXP or RXM input from falsely trigger the decoder. When the input exceeds the squelch limit, the PLL locks onto the incoming signal and the KS8737 decodes a data frame. This activates the carrier sense (CRS) ad RXDV signals and makes the receive data (RXD) available. The receive clock is maintained active during idle periods in between data reception. The KS8737 supports extended length cables for 10BaseT by selecting a lower squelch level around 150mV.

### SQE and Jabber Function (10BaseT only)

In 10BaseT operation, a short pulse will be put out on the COL pin after each packet is transmitted. This is required as a test of the 10BaseT transmit/receive path and is called SQE test. The 10BaseT transmitter will be disabled and COL will go High if TXEN is High for more than 46 us (Jabbering) If TXEN then goes Low for more than 368 us, the 10BaseT transmitter will be re-enabled and COL will go Low.

The KS8737 performs auto-negotiation by hardware (mode[1:0]) or software (Register 0.12). It will automatically choose its mode of operation by advertising its abilities and comparing them with those received from its link partner whenever auto-negotiation is enabled. It can also be configured to advertise 100BaseTX or 10BaseT in either full- or half-duplex mode. The auto-negotiation is disabled in the FX mode.

During auto-negotiation, the contents of Register 4, coded in Fast Link Pulse (FLP), will be sent to its link partner under the conditions of power-on, link-loss or re-start. At the same time, the KS8737 will monitor incoming data to determine its mode of operation. Parallel detection circuit will be enabled as soon as either 10BaseT idle or 100BaseTX idle is detected. The operation mode gets configured based on the following priority:

Priority 1: 100BaseTX, Full-duplex

Priority 2: 100BaseTX, Half-duplex

Priority 3: 10BaseT, Full-duplex

Priority 4: 10BaseT, Half-duplex

When the KS8737 receives a burst of FLP from its link partner with 3 identical link code words (ignoring acknowledge bit), it will store these code words in Register 5 and wait for the next 3 identical code words. Once the KS8737 detects the second code words, it then configures itself according to above-mentioned priority. In addition, the KS8737 also checks 100BaseTX idle or 10BaseT NLP symbol. If either is detected, the KS8737 automatically configures to match the detected operating speed.

#### **MII Management Interface**

The KS8737 supports the IEEE 802.3 MII Management Interface, also known as the Management Data Input / Output (MDIO) Interface. This interface allows upper-layer devices to monitor and control the state of the KS8737. The MDIO interface consists of the following:

- A physical connection including a data line (MDIO), a clock line (MDC) and an optional interrupt line (INTRPT)
- A specific protocol which runs across the above-mentioned physical connection and allows one controller to communicate with multiple KS8737 devices. Each KS8737 assigned an MII address between 0 and 31 by the PHYAD inputs.
- An internal addressable set of fourteen 16-bit MDIO registers. Register [0:6] are required and their functions are specified by the IEEE 802.3 specifications. Additional registers are provided for expanded functionality.

The INTPRT pin functions as a management data interrupt in the MII. An active Low or High in this pin indicates a status change on the KS8737 based on 1fh.14 level control. Register 1bh[15:8] are the interrupt enable bits. Register 1bh[7:0] are the interrupt conditions bits. The interrupt is activated when changes made to the following conditions:

- Link Status
- Duplex Status
- Reading Register 1bh clears this interrupt.

#### **MII Data Interface**

The data interface consists of separate channels for transmitting data from a 10/100 802.3 compliant Media Access Controller (MAC) to the KS8737, and for receiving data from the line. Normal data transmission is implemented in 4B Nibble Mode (4bit wide nibbles).

**Transmit Clock (TXC):** The transmit clock is normally generated by the KS8737 from an external 25MHz reference source at the X1 input. The transmit data and control signals must always be synchronized to the TXC by the MAC. The KS8737 normally samples these signals on the rising edge of the TXC.

**Receive Clock (RXC):** For 100BaseTX links, the receive clock is continuously recovered from the line. If the link goes down, and auto-negotiation is disabled, receive clock operates off the master input clock (X1 or TXC). For 10BaseT links, received is recovered from the line while carrier is active, and operates from the master input clock when the line is idled. The KS8737 synchronizes the receive data and control signals on the falling edge of RXC in order to stabilize the signals at the rising edge of the clock with 10ns setup and hold times.

**Transmit Enable:** The MAC must assert TXEN the same time as the first nibble of preamble, and de-assert TXEN after the last bit of the packet.

**Receive Data Valid:** The KS8737 asserts RXDV when it receives a valid packet. Line operating speed and MII mode will determine timing changes in the following way:

- For 100BaseTX link with the MII in 4B mode, RXDV is asserted from the first nibble of preamble to the last nibble of the data packet.
- For 10BaseT links, the entire preamble is truncated. RXDV is asserted with the first nibble of the SFD " 5D" and remains asserted until the end of the packet.

**Error Signals:** Whenever the KS8737 receives an error symbol from the network, it asserts RXER and drives "1110" (4B) on the RXD pins. When the MAC asserts TXER, the KS8737 will drive "H" symbols out on the line.

**Carrier Sense (CRS):** For 100TX links, a start-of-stream delimiter, or /J/K symbol pair causes assertion of Carrier Sense (CRS). An end-of-stream delimiter, or /T/R symbol pair causes de-assertion of CRS. The PMA layer will also de-assert CRS if IDLE symbols are received without /T/R, yet in this case RXER will be asserted for one clock cycle when CRS is de-asserted. For 10T links, CRS assertion is based on reception of valid preamble, and de-assertion on reception of an end-of-frame (EOF) marker.

**Collision:** Whenever the line state is half-duplex and the transmitter and receiver are active at the same time, the KS8737 asserts its collision signal which is asynchronous to any clock.

#### **Power Management**

The KS8737 offers the following modes for power management:

- Power Down Mode: This mode can be achieved by writing to Register 0.11 or pulling pin PWRDWN High.
- Power Saving Mode: This mode can be enabled by writing to Register 1fh.15. or using an external initialization pin. The KS8737 will then turn off everything except for the Energy Detect and PLL circuits when the cable is not installed. In other words, the KS8737 will shutdown most of the internal circuits to save power if there is no link.

#### Fiber Mode

Fiber mode is activated by setting FXMODEB (pin46) low. Under this mode, the FIBP/M are become the receiving port, and the TXP/M are the transmit port. FXSD (+) and FXSD\_THD (-) are used as differential signal for Fiber signal detect port. If driven single-ended with FXSD and FXSD\_THD should be set by an external voltage divider for the proper reference voltage, there is no internal voltage for this pin.

Under Fiber mode, the link is up only when FXSD>FXSD\_THD and the proper idle pattern is received. If FXSD<FXSD\_THD, then a Far-end-fault (FEF) pattern will be sent out. The link partner that receives FEF will have its link turn off. The link partner, however, will still send the normal transmission while receiving FEF. This feature can be turned off by setting the Pin 17 low.

#### Media Converter Mode

The KS8737 provides a special fiber mode in which allows back-to-back FX to TX media conversion using two KS8737's. This special mode can be activated by pulling pin 32 high through an external 1k resistor. The detailed connection between the two KS8737's is shown in the application circuit on the data sheet. In this case TXC become an input pin. The internal FIFO's will take care of the transition of the receive to transmit clock domain changes. The KS8737 in media converter mode can also handle the jumbo frames. The recovered clock and parallel data (RXC, RXD[0:3]) from the other chip will feed the chip through TXC and TXD[0:3].

Under the Media Converter mode, the KS8737 support disable transmit feature by pull-up on DSITX/LPBK (pin 33) pin. For more details, see "*Pin Description*" section.

# **Register Map**

Register No.	Description
Oh	Basic Control Register
1h	Basic Status Register
2h	PHY Identifier I
3h	PHY Identifier II
4h	Auto-Negotiation Advertisement Register
5h	Auto-Negotiation Link Partner Ability Register
6h	Auto-Negotiation Expansion Register
7h	Auto-Negotiation Next Page Register
15h	RXER Counter Register
1bh	Interrupt Control/Status Register
1ch	Reserved
1dh	Reserved
1eh	Reserved
1fh	100BaseTX PHY Control Register

Address	Name	Description	Mode <sup>(Note 1)</sup>	Default		
Register 0h - Basic Control						
0.15	Reset	1 = software reset. Bit is self-clearing	RW/SC	0		
0.14	Loopback	1 = loopback mode 0 = normal operation	RW	0		
0.13	Speed Select	1 = 100Mbps 0 = 10Mbps Ignored if Auto-Negotiation is enabled (0.12 = 1)	RW	Set by MODE Selection Table		
0.12	Auto-Negotiation Enable	1 = enable auto-negotiation process (override 0.13 and 0.8) 0 = disable auto-negotiation process	RW	Set by MODE Selection Table		
0.11	Power Down	1 = power down mode 0 = normal operation	RW	0		
0.10	Isolate	1 = electric isolation of PHY from MII and TXP/TXM 0 = normal operation	RW	Set by MODE Selection Table		
0.9	Restart Auto-Negotiation	<ul><li>1 = restart auto-negotiation process</li><li>0 = normal operation. Bit is self-clearing</li></ul>	RW/SC	0		
0.8	Duplex Mode	1 = full-duplex 0 = half duplex	RW	Set by FDX (pin 13)		
0.7	Collision Test	1 = enable COL test 0 = disable COL test	RW	0		
0.6:0	Reserved		RO	0		
Register	1h - Basic Status					
1.15	100BaseT4	1 = T4 capable 0 = no T4 capable	RO	0		
1.14	100BaseTX Full-Duplex	1 = capable of 100BaseX full-duplex 0 = not capable of 100BaseX full-duplex	RO	1		
1.13	100BaseTX Half-Duplex	1 = capable of 100BaseX half duplex 0 = not capable of 100BaseX half duplex	RO	1		

Address	Name	Description	Mode <sup>(Note 1)</sup>	Default
1.12	10BaseT Full-Duplex	1 = 10Mbps with full-duplex 0 = no 10Mbps with full-duplex capability	RO	1
1.11	10BaseT Half-Duplex	1 = 10Mbps with half duplex 0 = no 10Mbps with half duplex capability	RO	1
1.10:7	Reserved		RO	0000
1.6	No Preamble	1 = preamble suppression 0 = normal preamble	RO	0
1.5	Auto-Negotiation Complete	<ul><li>1 = auto-negotiation process completed</li><li>0 = auto-negotiation process not completed</li></ul>	RO	0
1.4	Remote Fault	1 = remote fault 0 = no remote fault	RO/LH	0
1.3	Auto-Negotiation Ability	<ul><li>1 = capable to perform auto-negotiation</li><li>0 = unable to perform auto-negotiation</li></ul>	RO	1
1.2	Link Status	1 = link is up 0 = link is down	RO/LL	0
1.1	Jabber Detect	<ul><li>1 = jabber detected</li><li>0 = jabber not detected. Default is Low</li></ul>	RO/LH	0
1.0	Extended Capability	1 = supports extended capabilities registers	RO	1
Register 2	h - PHY Identifier 1			
2.15:0	PHY ID Number	Assigned to the 3rd through 18th bits of the Organizationally Unique Identifier (OUI). Micrel Semiconductor's OUI is 0010A1 (hex)	RO	0022h
Register 3	Sh - PHY Identifier 2			
3.15:10	PHY ID Number	Assigned to the 19th through 24th bits of the Organizationally Unique Identifier (OUI). Micrel Semiconductor's OUI is 0010A1 (hex)	RO	000101
3.9:4	Model Number	Six bit manufacturer's model number	RO	110010
3.3:0	Revision Number	Four bit manufacturer's model number	RO	0000
Register 4	h - Auto-Negotiation Adve	rtisement	I	1
4.15	Next Page	1 = next page capable 0 = no next page capability. KS8737 supports next page capability	RW	0
4.14	Reserved		RO	00
4.13	Remote Fault	1 = remote fault supported 0 = no remote fault	RW	0
4.12:11	Reserved		RO	0
4.10	Pause	<ul><li>1 = pause function supported</li><li>0 = no pause function</li></ul>	RW	0
4.9	100BaseT4	1 = T4 capable 0 = no T4 capability	RO	0
4.8	100BaseTX Full-Duplex	1 = 100 full-duplex capable 0 = no 100 full-duplex capability	RW	Set by MODE Selection Table
4.7	100BaseTX Half-Duplex	1 = 100Mbps half-duplex capable 0 = no 100Mbps half-duplex capability	RW	Set by MODE Selection Table
4.6	10BaseT Full-Duplex	1 = 10Mbps full-duplex 0 = no 10Mbps full-duplex capability	RW	Set by MODE Selection Table
4.5	10BaseT Half-Duplex	1 = 10Mbps half-duplex capable 0 = no 10Mbps half-duplex capability	RW	Set by MODE Selection Table
4.4:0	Selector Field	[00001] = IEEE 802.3	RW	00001

Address	Name	Description	Mode <sup>(Note 1)</sup>	Default
Register :	5h - Auto-Negotiation Link	Partner Ability		
5.15	Next Page	1 = next page capable 0 = no next page capability KS8737 supports next page capability.	RO	0
5.14	Acknowledge	<ul><li>1 = link code word received from partner</li><li>0 = link code word not yet received</li></ul>	RO	0
5.13	Remote Fault	1 = remote fault detected 0 = no remote fault	RO	0
5.12:11	Reserved		RO	00
5.10	Pause	1 = pause function supported 0 = no pause function	RO	0
5.9	100 BaseT4	1 = T4 capable 0 = no T4 capability	RO	0
5.8	100BaseTX Full-Duplex	1 = 100Mbps full-duplex capable 0 = no 100Mbps full-duplex capability	RO	0
5.7	100BaseTX Half-Duplex	1 = 100Mbps half-duplex capable 0 = no 100Mbps half-duplex capability	RO	0
5.6	10BaseT Full-Duplex	1 = 10Mbps with full-duplex capable 0 = no 10Mbps full-duplex capability	RO	0
5.5	10BaseT Half-Duplex	1 = 10Mbps half-duplex capable 0 = no 10Mbps half-duplex capability	RO	0
5.4:0	Selector Field	[00001] = IEEE 802.3	RO	00000
Register (	6h - Auto-Negotiation Exp	ansion		
6.15:5	Reserved		RO	0×00
6.4	Parallel Detection Fault	<ul><li>1 = fault detected by parallel detection</li><li>0 = no fault detected by parallel detection.</li></ul>	RO/LH	0
6.3	Link Partner Next Page Able	<ul><li>1 = link partner has next page capability</li><li>0 = link partner does not have next page capability</li></ul>	RO	0
6.2	Next Page Able	<ul><li>1 = local device has next page capability</li><li>0 = local device does not have next page capability</li></ul>	RO	1
6.1	Page Received	1 = new page received 0 = new page not yet received	RO/LH	0
6.0	Link Partner Auto-Negotiation Able	<ul><li>1 = link partner has auto-negotiation capability</li><li>0 = link partner does not have auto-negotiation capability</li></ul>	RO	0
Register 7	7h - Auto-Negotiation Nex	t Page		
7.15	Next Page	1 = additional next page(s) will follow 0 = last page	RW	0
7.14	Reserved		RO	0
7.13	Message Page	1 = message page 0 = unformatted page	RW	1
7.12	Acknowledge2	1 = will comply with message 0 = cannot comply with message	RW	0
7.11	Toggle	<ul> <li>1 = previous value of the transmitted link code word</li> <li>equaled logic One</li> <li>0 = logic Zero</li> </ul>	RO	0
7.10:0	Message Field	11-bit wide field to encode 2048 messages	RW	0×01
Register 2	15h - RXER Counter			
15h.15:0	RXER Counter	RX Error counter for the RX_ER in each package	RO	0×00

Address	Name	Description	Mode <sup>(Note 1)</sup>	Default
Register	1bh - Interrupt Control/Stat	us Register		
1bh.15	Jabber Interrupt Enable	1 = Enable jabber interrupt 0 = Disable jabber interrupt	RW	0
1bh.14	Receive Error Interrupt Enable	<ul><li>1 = Enable receive error interrupt</li><li>0 = Disable receive error interrupt</li></ul>	RW	0
1bh.13	Page Received Interrupt Enable	<ul><li>1 = Enable page received interrupt</li><li>0 = Disable page received interrupt</li></ul>	RW	0
1bh.12	Parallel Detect Fault Interrupt Enable	<ul><li>1 = Enable parallel detect fault interrupt</li><li>0 = Disable parallel detect fault interrupt</li></ul>	RW	0
1bh.11	Link Partner Acknowledge Interrupt Enable	<ul><li>1 = Enable link partner acknowledge interrupt</li><li>0 = Disable link partner acknowledge interrupt</li></ul>	RW	0
1bh.10	Link Down Interrupt Enable	1 = Enable link down interrupt 0 = Disable link down interrupt	RW	0
1bh.9	Remote Fault Interrupt Enable	1 = Enable remote fault interrupt 0 = Disable remote fault interrupt	RW	0
1bh.8	Link Up Interrupt Enable	1 = Enable link up interrupt 0 = Disable link up interrupt	RW	0
1bh.7	Jabber Interrupt	1 = Jabber interrupt 0 = No jabber interrupt	RO	0
1bh.6	Receive Error Interrupt	1 = Receive error interrupt 0 = No receive error interrupt	RO	0
1bh.5	Page Receive Interrupt	1 = Page receive interrupt 0 = No page receive interrupt	RO	0
1bh.4	Parallel Detect Fault Interrupt	1 = Parallel detect fault interrupt 0 = No parallel detect fault interrupt	RO	0
1bh.3	Link Partner Acknowledge Interrupt	1 = Link partner acknowledge interrupt 0 = No link partner acknowledge interrupt	RO	0
1bh.2	Link Down Interrupt	1 = Link down interrupt 0 = No link down interrupt	RO	0
1bh.1	Remote Fault Interrupt	1 = Remote fault interrupt 0 = No remote fault interrupt	RO	0
1bh.0	Link Up Interrupt	1 = Link up interrupt 0 = No link up interrupt	RO	0
Register	Ifh - 100BaseTX PHY Cont	roller		
1fh.15	Power Saving	1 = enable power saving 0 = disable	RW	1
1fh.14	Interrupt Level	1 = interrupt pin active high 0 = active low	RW	0
1fh.13	Reserved			
1fh.12	Auto-Negotiation Complete	1 = auto-negotiation complete 0 = not complete	RO	0
1fh.11	Enable Link Fail Counter in 100BaseTX	1 = enable link fail counter 0 = disable	RW	1
1fh.10	Enable Jabber	1 = enable jabber counter 0 = disable	RW	1
1fh.9	Reserved	Reserved		
1fh.8	Enable Pause (Flow-Control Result)	1 = flow control capable 0 = no flow control	RO	0

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Address	Name	Description	Mode <sup>(Note 1)</sup>	Default
1fh.7	Enable SQE Test	1 = enable SQE test 0 = disable	RW	1
1fh.6	Enable Symbol Mode	1 = enable symbol mode 0 = disable	RO	0
1fh.5	Transmit Isolate	1 = isolate transmit 0 = not isolate	RO	0
1fh.4:2	Operation Mode Indication	[000] = still in auto-negotiation [001] = 10BaseT half duplex [010] = 100BaseTX half duplex [011] = default [101] = 10BaseT full-duplex [110] = 100BaseTX full-duplex [111] = PHY/MII isolate	RO	Set by MODE
1fh.1	Reserved	Reserved	RW	0
1fh.0	Disable Data Scrambling	1 = disable scrambler 0 = enable	RW	0

# Mode Selection for Register 1fh 4:2

KS8737 can be forced into a specific mode on reset by configuring MODE pins specified in the following table. The strapping option of MODE pins are latched on the rising edge of reset to set the default value of various registers. The values can be modified by writing into the registers.

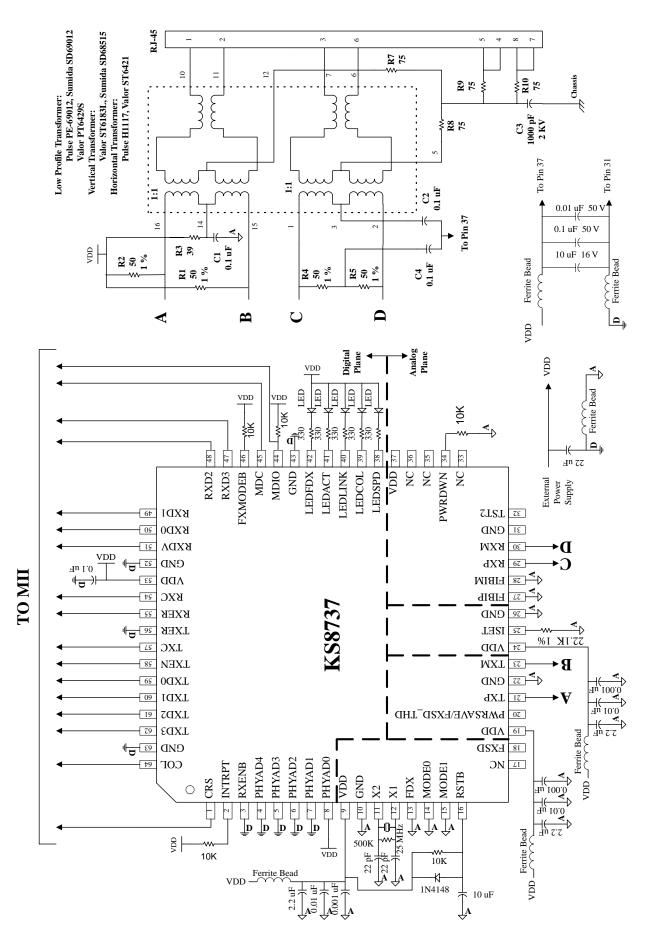
				Default Register Bit Values					
Mod (Note			Mode Definition	Register 0h Bit [8,10,12,16]		Register 1fh Bit 5			
1 0	)	0	Force 10BaseT Half-Duplex. No Auto-Negotiation.	0 0 0 0	1 0 0 0	0			
1 (		1	Force 10BaseT Full-Duplex. No Auto-Negotiation.	1 0 0 0	1 1 0 0	0			
0 1	1	0	Force 100BaseTX Half-Duplex. No Auto-Negotiation.	0 0 0 1	0 0 1 0	0			
0 1	1	1	Force 100BaseTX Full-Duplex. No Auto-Negotiation.	1 0 0 1	0 0 1 1	0			
Z 1	1	*	100 BaseTX Half & Full-Duplex Advertised via Auto-Negotiation.	* 0 1 1	0 0 1 1	0			
ZO	)	*	10 BaseT Half & Full-Duplex Advertised via Auto-Negotiation.	* 0 1 0	1 1 0 0	0			
1 1	1	*	Isolated MII and TXP/TXM	* 1 0 1	0 0 0 0	1			
0 0	)	*	All Capable Auto-Negotiation Enabled	* 0 1 0	1 1 1 1	0			

#### Table 1. Mode Selection

**Note 1.** Z indicates that input is floating.

Note 2. \* indicates that values are controlled by FDX.

## **Typical Application Circuit**



Supply Voltage (V <sub>DD</sub> )	+4.0V
Supply Reference to GND	–0.5V to +7.0V
Input Voltage (All Inputs)	–0.5V to +4.0V
Output Voltage (All Outputs)	–0.5V to +4.0V
Lead Temperature (soldering, 10 sec.)	270°C
Storage Temperature (T <sub>S</sub> )5	55°C to +150°C

# **Operating Ratings (Note 2)**

Supply Voltage (V <sub>DD</sub> )	+3.135V to +3.465V
Ambient Temperature (T <sub>A</sub> )	0°C to +70°C
Package Thermal Resistance, (Note	e 3)
TQFP (θ <sub>.IA</sub> ) No Air Flow	53.7°C/W

## **Electrical Characteristics (Note 4)**

 $V_{DD}$  = 3.3V ±5%;  $T_A$  = 0°C to +70°C; unless noted.

Symbol	Parameter	Condition	Min	Тур	Max	Units
Total Supp	ly Current (including TX output drive	er current)	•	-		
I <sub>DD1</sub>	Normal 100BaseTX			100	120	mA
I <sub>DD2</sub>	Normal 10BaseT (50% utilization)			135	150	mA
I <sub>DD3</sub>	Power Saving Mode 100BaseTX			80	90	mA
I <sub>DD4</sub>	Power Save Mode 10BaseT			40	45	mA
I <sub>DD5</sub>	Power Down Mode			9	10	mA
TTL Inputs	•	-	ł	•		4
VIH	Input High Voltage		2.0			V
V <sub>IL</sub>	Input Low Voltage				0.8	V
I <sub>IN</sub>	Input Current	V <sub>IN</sub> = GND ~ V <sub>DD</sub>	-10		10	μA
TTL Outpu	ts	1	•			<u> </u>
V <sub>OH</sub>	Output High Voltage	$I_{OH} = -4mA$	2.4			V
V <sub>OL</sub>	Output Low Voltage	$I_{OL} = 4mA$			0.4	V
I <sub>oz</sub>	Output Tr-State Leakage				10	μA
100BaseT)	( Receive	•	1			4
V <sub>B</sub>	RXP/RXM Input Bias Voltage			2.7		V
R <sub>IN</sub>	RXP/RXM Differential Input Resistance			8		kΩ
	Propagation Delay	from magnetics to RDTX		50	110	ns
100BaseT)	<pre></pre> C Transmit (measured differentially at a second se	fter 1:1 transformer)	•			<u> </u>
Vo	Peak Differential Output Voltage	50Ω from each output to $V_{DD}$	0.95		1.05	V
V <sub>IMB</sub>	Output Voltage Imbalance	50Ω from each output to $V_{DD}$			2	%
t <sub>r</sub> , t <sub>t</sub>	Rise/Fall Time Rise/Fall Time Imbalance		3 0		5 0.5	ns ns
100BaseT>	C Transmit (measured differentially at	fter 1:1 transformer)	ł	•		
	Duty Cycle Distortion				±0.5	ns
	Overshoot				5	%
V <sub>SET</sub>	Reference Voltage of ISET			1.25		V
-	Propagation Delay	from TDTX to magentics		45	60	ns
	Jitters			0.7	1.4	ns <sub>(pk-pk</sub>

Note 2. The device is not guaranteed to function outside its operating rating. Unused inputs must always be tied to an appropriate logic voltage level (Ground to V<sub>DD</sub>).

Note 3. No HS (heat spreader) in package.

**Note 4.** Specification for packaged product only.

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Symbol	Parameter	Condition	Min	Тур	Max	Units
10BaseTX	Receive	•	•		<u>.</u>	
V <sub>B</sub>	RXP/RXM Input Bias Voltage			1.4		V
R <sub>IN</sub>	RXP/RXM Differential Input Resistance			8		kΩ
V <sub>SQ</sub>	Squelch Threshold	5MHz square wave		400		mV
10BaseTX	Transmit (measured differentially aft	er 1:1 transformer)	•			
V <sub>P</sub>	Peak Differential Output Voltage	$50\Omega$ from each output to V <sub>DD</sub>	2.2		2.8	V
	Jitters Added	$50\Omega$ from each output to V <sub>DD</sub>			±3.5	ns
t <sub>r</sub> , t <sub>t</sub>	Rise/Fall Time			25		ns
Clock Outp	outs	-	•			
X1, X2	Crystal Oscillator			25		MHZ
RXC <sub>100</sub>	Receive Clock, 100TX			25		MHZ
RXC <sub>10</sub>	Receive Clock, 10T			2.5		MHZ
	Receive Clock Jitters			3.0		ns <sub>(pk-pk)</sub>
TXC <sub>100</sub>	Transmit Clock, 100TX			25		MHZ
TXC <sub>10</sub>	Transmit Clock, 10T			2.5		MHZ
	Transmit Clock Jitters			1.8		ns <sub>(pk-pk)</sub>

## **Timing Diagrams**

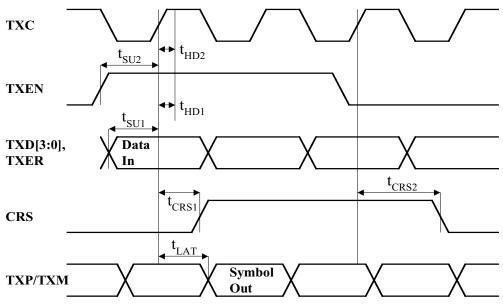


Figure 1. 10BaseT MII Transmit Timing

Symbol	Parameter	Min	Тур	Max	Units
t <sub>SU1</sub>	TXD [3:0] Set-Up to TXC High	10			ns
t <sub>SU2</sub>	TXEN Set-Up to TXC High	10			ns
t <sub>HD1</sub>	TXD [3:0] Hold After TXC High	0			ns
t <sub>HD2</sub>	TXEN Hold After TXC High	0			ns
t <sub>CRS1</sub>	TXEN High to CRS Asserted Latency			2	BT
t <sub>CRS2</sub>	TXEN Low to CRS De-Asserted Latency			5	BT
t <sub>LAT</sub>	TXEN High to TXP/TXM Output (TX Latency)			3	BT
t <sub>SQE</sub>	COL (SQE) Delay Aftter TXEN Ae-Asserted		1.5		μs
t <sub>SQEP</sub>	COL (SQE) Pulse Duration		1.0		μs

Table 2. 10BaseT MII Transmit Timing Parameters

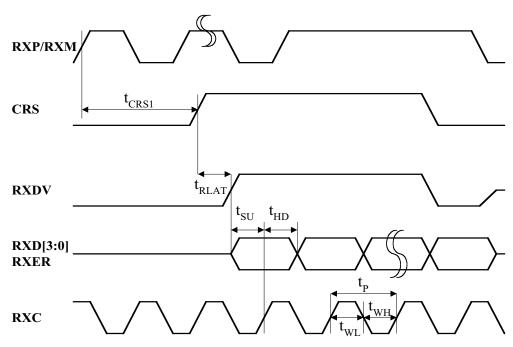


Figure 2. 10BaseT MII Receive Timing

Symbol	Parameter	Min	Тур	Max	Units
t <sub>P</sub>	RXC Period		400		ns
t <sub>WL</sub>	RXC Pulse Width		200		ns
t <sub>WH</sub>	RXC Pulse Width		200		ns
t <sub>SU</sub>	RXD [3:0], RXER, RXDV Set-Up to Rising Edge of RXC		200		ns
t <sub>HD</sub>	RXD [3:0], RXER, RXDV Hold from Rising Edge of RXC		200		ns
t <sub>RLAT</sub>	RXDV Data Vaild from CSR			6.5	μs
t <sub>CRS1</sub>	RXP/RXM Preamble to CRS Asserted		5		BT

#### Table 3. 10BaseT MII Receive Timing Parameters

Note 1. CRS is asserted but RXD/RXDV are driven from SFD as the first byte of packet.

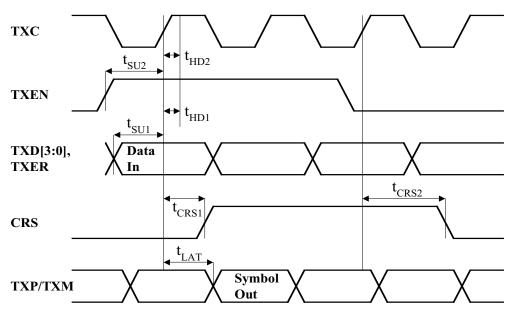


Figure 3. 100BaseT MII Transmit Timing

Symbol	Parameter	Min	Тур	Max	Units
t <sub>SU1</sub>	TXD [3:0] TXER Set-Up to TXC High	10			ns
t <sub>SU2</sub>	TXEN Set-Up to TXC High	10			ns
t <sub>HD1</sub>	TXD [3:0] TXER Hold After TXC High	0			ns
t <sub>HD2</sub>	TXEN Hold After TXC High	0			ns
t <sub>CRS1</sub>	TXEN High to CRS Asserted Latency		1		BT
t <sub>CRS2</sub>	TXEN Low to CRS De-Asserted Latency		1		BT
t <sub>LAT</sub>	TXEN High to TXP/TXM Output (TX Latency)		7		BT

Table 4. 100BaseT MII Transmit Timing Parameters

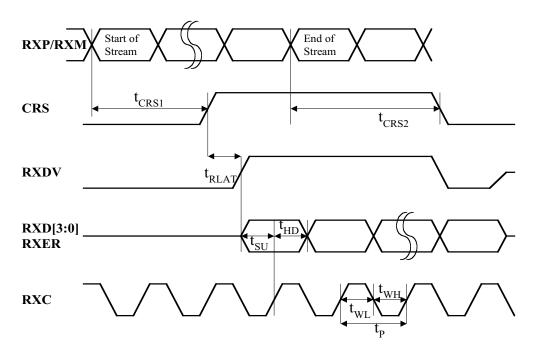


Figure 4. 100BaseT MII Receive Timing

Symbol	Parameter	Min	Тур	Max	Units
t <sub>P</sub>	RXC Period		40		ns
t <sub>WL</sub>	RXC Pulse Width	20			ns
t <sub>WH</sub>	RXC Pulse Width	20			ns
t <sub>SU</sub>	RXD [3:0], RXER, RXDV Set-Up to Rising Edge of RXC		20		ns
t <sub>HD</sub>	RXD [3:0], RXER, RXDV Hold from Rising Edge of RXC		20		ns
t <sub>RLAT</sub>	CRS to RXD Latency, 4B or 5B Aligned		4		BT
t <sub>CRS1</sub>	"Start of Stream" to CSR Asserted		140		ns
t <sub>CRS2</sub>	"End of Stream" to CSR De-Asserted		170		ns

Table 5. 100BaseT MII Receive Timing Parameters

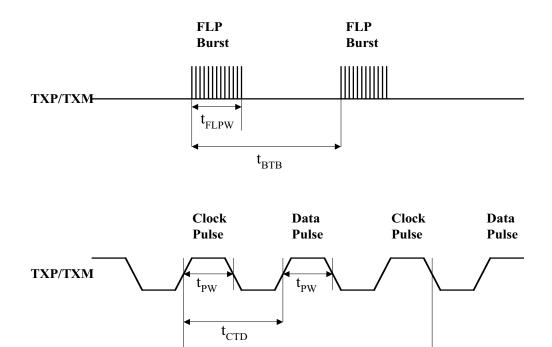
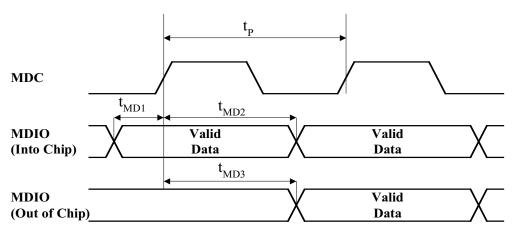


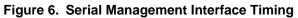
Figure 5. Auto Negotiation/Fast Link Pulse Timing

t<sub>CTC</sub>

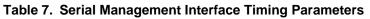
Symbol	Parameter	Min	Тур	Max	Units
t <sub>BTB</sub>	FLP Burst to FLP Burst	8	16	24	ms
t <sub>FLPW</sub>	FLP Burst Width		2		ms
t <sub>PW</sub>	Clock/Data Pulse Width		100		ns
t <sub>CTD</sub>	Clock Pulse to Data Pulse		69		μs
<sup>t</sup> стс	Clock Pulse to Clock Pulse Number of Clock/Data Pulses per Burst	17	136	33	μs μs

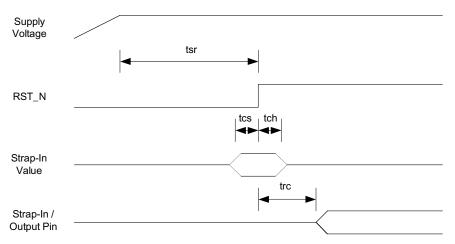
Table 6. Auto Negotiation/Fast Link Pulse Timing Parameters





Symbol	Parameter	Min	Тур	Max	Units
t <sub>P</sub>	MDC Period		400		ns
t <sub>MDI</sub>	MDIO Set-Up to MDC (MDIO as Input)	10			ns
t <sub>MD2</sub>	MDIO Hold After MDC (MDIO as Input)	10			ns
t <sub>MD3</sub>	MDC to MDIO Valid (MDIO as Output)		250		ns







Symbol	Parameter	Min	Тур	Max	Units
t <sub>sr</sub>	Stable Supply Voltages to Reset High	10			ms
t <sub>cs</sub>	Configuration Set-Up Time	50			ns
t <sub>ch</sub>	Configuration Hold Time	50			ns
t <sub>rc</sub>	Reset to Strap-In Pin Output	50			μs

Table 8.	Reset	Timing	Parameters
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One simple 1:1 isolation transformer is needed at the line interface. An isolation transformer with integrated common-mode choke is recommended for exceeding FCC requirements. The following table gives recommended transformer characteristics.

Characteristics Name	Value	Test Condition	
Turns Ratio	1 CT : 1 CT		
Open-Circuit Inductance (min.)	350μΗ	100mV, 100kHz, 8mA	
Leakage Inductance (max.)	0.4µH	1MHz (min.)	
Inter-Winding Capacitance (max.)	12pF		
D.C. Resistance (max.)	0.9Ω		
Insertion Loss (max.)	1.0dB	0MHz – 65MHz	
HIPOT (min.)	1500Vrms		

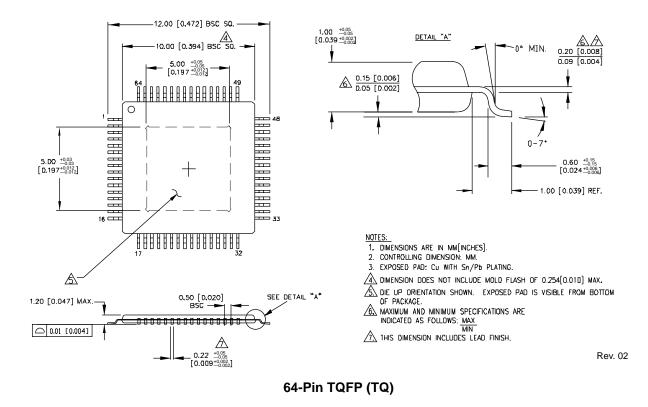
Note 1. The IEEE 802.3u standard for 100BaseTX assumes a transformer loss of 0.5 dB. For the transmit line transformer, insertion loss of up to 1.3dB can be compensated by increasing the line drive current by means of reducing the ISET resistor value.

## Selection of Reference Crystal

An oscillator or crystal with the following typical characteristics is recommended.

Characteristics Name	Value	Units
Frequency	25.00000	MHz
Frequency Tolerance (max.)	±100	ppm
Load Capacitance (max.)	20	pF
Series Resistance (max.)	25	Ω

### **Package Information**



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