

Technical documentation



Support & training



DLP3021-Q1 DLPS183C – MARCH 2020 – REVISED MARCH 2023

# DLP3021-Q1 0.3-Inch WVGA Automotive DMD

# 1 Features

- Qualified for automotive applications:
  - 40°C to 105°C operating DMD array temperature range
- 0.3-inch diagonal micromirror array
  - 7.6-µm micromirror pitch
  - ±12° micromirror tilt angle (relative to flat state)
  - Side illumination for reduced system size
- WVGA (864 × 480) input resolution
- Polarization independent spatial light modulator
- Compatible with LED or laser light sources
- Low-power consumption: 255 mW (maximum)
- Operating temperature range: –40°C to 105°C
- Hermetic package
- JTAG boundary scan to allow in-system validation
- Compatible with the DLPC120-Q1 automotive
  DMD controller
- 80-MHz DDR DMD interface

# 2 Applications

Automotive small light

3 Description

The DLP3021-Q1 Automotive DMD is primarily targeted for automotive exterior light control and display applications, such as ground projection with the ability to display dynamic content. Ground projections can help facilitate vehicle to pedestrian (V2P) communication, such as back up and door open warnings, in addition to orchestrating vehicle communication systems and vehicle personalization options. Due to their small form factor and low power operation, projectors with the DLP3021-Q1 chipset can support many projection applications. They can be placed in many locations in the car including inside the side mirror, door panel, tail light, front grill, and more. This chipset can be coupled with LEDs or lasers to create highly saturated colors with over 125% NTSC color gamut, and can be used with either RGB or white illumination sources. A DLP® Products FPGA configuration can be used to drive the DLP3021-Q1 Automotive DMD to reduce form factor for easy integration in a vehicle. The DLPC120-Q1 Automotive DMD controller can also be used to drive the DLP3021-Q1 Automotive DMD with support for 24-bit RGB video input to increase content flexibility.

#### **Device Information**

PART NUMBER <sup>(1)</sup>	PACKAGE	BODY SIZE (NOM)
DLP3021-Q1	FQR (64)	8.55 mm × 16.80 mm

(1) For all available packages, see the orderable addendum at the end of the data sheet.



### DLP3021-Q1 System Block Diagram

An IMPORTANT NOTICE at the end of this data sheet addresses availability, warranty, changes, use in safety-critical applications, intellectual property matters and other important disclaimers. PRODUCTION DATA.



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# **4 Revision History**

NOTE: Page numbers for previous revisions may differ from page numbers in the current version.

С	hanges from Revision B (March 2022) to Revision C (March 2023) Page
•	Added Automotive Qualification feature bullet1
С	hanges from Revision A (May 2020) to Revision B (March 2022) Page
•	This document is updated per the latest Texas Instruments and industry data sheet standards
С	hanges from Revision * (March 2020) to Revision A (May 2020) Page
•	Changed the device status from Advance Information to Production Data



# **5** Pin Configuration and Functions



Figure 5-1. FQR Package, 64-Pin LGA (Bottom View)



#### Table 5-1. Pin Functions

PIN			DECODIDITION		
NAME	NO.	TYPE	DESCRIPTION		
DATA(0)	A2				
DATA(1)	A4				
DATA(2)	B2				
DATA(3)	B3				
DATA(4)	B5				
DATA(5)	C2	C2 C3			
DATA(6)	C3				
DATA(7)	B4 Data bus. Synchronous to risir		Data bus. Synchronous to rising edge and falling edge of DCLK.		
DATA(8)	C5				
DATA(9)	D2				
DATA(10)	D3				
DATA(11)	D4				
DATA(12)	D5				
DATA(13)	E2				
DATA(14)	F5	IVCMOS input			
DCLK	F4	LVCMOS input	Data clock.		
LOADB	F3		Parallel latch load enable. Synchronous to rising edge and falling edge of DCLK.		
SCTRL	E4		Serial control (sync). Synchronous to rising edge and falling edge of DCLK.		
TRC	F2		Toggle rate control. Synchronous to rising edge and falling edge of DCLK.		
DAD_BUS	B15		Reset control serial bus. Synchronous to rising edge of SAC_CLK.		
RESET_OEZ	C15		Active low. Output enable signal for internal reset driver circuitry.		
RESET_STROBE	B13		Rising edge on RESET_STROBE latches in the control signals.		
SAC_BUS	A15		Stepped address control serial bus. Synchronous to rising edge of SAC_CLK.		
SAC_CLK	A14		Stepped address control clock.		
тск	F15		JTAG clock.		
TDI	E13		JTAG data input. Synchronous to rising edge of TCK. Bond pad connects to internal pull up resistor.		
тро	G15	LVCMOS output	JTAG data output. Synchronous to falling edge of TCK. Tri-state failsafe output buffer.		
TMS	G14	LVCMOS input	JTAG mode select. Synchronous to rising edge of TCK. Bond pad connects to internal pull up resistor.		
TEMP_MINUS	G13	Angle - in set	Calibrated temperature diode used to assist accurate temperature measurements of DMD die.		
TEMP_PLUS	G2	Analog input			
V <sub>BIAS</sub>	D15		Power supply for positive bias level of mirror reset signal.		
V <sub>cc</sub>	A5, B12, C14, D12, F13, G3	Power	Power supply for low voltage CMOS logic. Power supply for normal high voltage at mirror address electrodes. Power supply for offset level of mirror reset signal during power down.		



## Table 5-1. Pin Functions (continued)

PIN		ТҮРЕ	DESCRIPTION		
NAME	NO.	TTPE	DESCRIPTION		
V <sub>OFFSET</sub>	E14		Power supply for high voltage CMOS logic. Power supply for stepped high voltage at mirror address electrodes. Power supply for offset level of mirror reset signal.		
V <sub>REF</sub>	E15		Power supply for low voltage CMOS DDR interface.		
V <sub>RESET</sub>	D14	Power	Power supply for negative reset level of mirror reset signal.		
V <sub>SS</sub>	A3, A13, B14, C4, C12, C13, D13, E3, E5, E12, F12, F14, G4, G12		Common return for all power.		
RESERVED	A1, A12, A16,B1, B16, F1, F16, G1, G5, G16	Reserved	Do not connect.		



# 6 Specifications

# 6.1 Absolute Maximum Ratings

#### See (2)

		MIN	MAX	UNIT
SUPPLY VOLTAGE <sup>(1)</sup>				
V <sub>REF</sub>	LVCMOS logic supply voltage	-0.5	4	V
V <sub>CC</sub>	LVCMOS logic supply voltage	-0.5	4	V
V <sub>OFFSET</sub>	Mirror electrode and HVCMOS voltage	-0.5	8.75	V
V <sub>BIAS</sub>	Mirror electrode voltage	-0.5	17	V
V <sub>BIAS</sub> – V <sub>OFFSET</sub>	Supply voltage delta <sup>(3)</sup>		8.75	V
V <sub>RESET</sub>	Mirror electrode voltage	-11	0.5	V
Input voltage: other inputs		-0.5	V <sub>REF</sub> + 0.3	V
f <sub>DCLK</sub>	Clock frequency	60	82	MHz
I <sub>TEMP_DIODE</sub>	Temperature diode current		500	μA
ENVIRONMENTAL				
T <sub>ARRAY</sub>	Operating DMD array temperature <sup>(4)</sup>	-40	105	°C

(1) All voltage values are with respect to the ground pins (V<sub>SS</sub>).

(2) Operation outside the Absolute Maximum Ratings may cause permanent device damage. Absolute Maximum Ratings do not imply functional operation of the device at these or any other conditions beyond those listed under Recommended Operating Conditions. If outside the Recommended Operating Conditions but within the Absolute Maximum Ratings, the device may not be fully functional, and this may affect device reliability, functionality, performance, and shorten the device lifetime.

(3) To prevent excess current, the supply voltage delta |V<sub>BIAS</sub> - V<sub>OFFSET</sub>| must be less than or equal to 8.75 V.

(4) See <u>Section</u> 7.6 section.

# 6.2 Storage Conditions

Applicable for the DMD as a component or non-operating in a system. The device is not designed to be exposed to corrosive environments.

		MIN	MAX	UNIT
T <sub>stg</sub>	DMD storage temperature	-40	125	°C

### 6.3 ESD Ratings

				VALUE	UNIT
		Human-body model (HBM), per AEC Q100-002 <sup>(1)</sup>	All Pins	±2000	
V <sub>(ESD)</sub>	Electrostatic discharge	Charged-device model (CDM) per AEC Q100-011	All Pins	±750	V
			Corner Pins <sup>(2)</sup>	±750	

(1) AEC Q100-002 indicates that HBM stressing shall be in accordance with the ANSI/ESDA/JEDEC JS-001 specification.

(2) Corner pins are A1, G1, A16, and G16.



# 6.4 Recommended Operating Conditions

Over operating free-air temperature range (unless otherwise noted)

			MIN	NOM	MAX	UNIT
SUPPLY VOLTAGE	RANGE					
V <sub>REF</sub>	LVCMOS interface power supply voltage		1.65	1.8	1.95	V
V <sub>CC</sub>	LVCMOS logic power supply voltage		2.25	2.5	2.75	V
V <sub>OFFSET</sub>	Mirror electrode and HVCMOS voltage		8.25	8.5	8.75	V
V <sub>BIAS</sub>	Mirror electrode voltage		15.5	16	16.5	V
V <sub>BIAS</sub> – V <sub>OFFSET</sub>	Supply voltage delta <sup>(2)</sup>				8.75	V
V <sub>RESET</sub>	Mirror electrode voltage		-9.5	-10	-10.5	V
V <sub>P</sub> VT+	Positive going threshold voltage		0.4 × V <sub>REF</sub>		$0.7 \times V_{REF}$	V
V <sub>N</sub> VT–	Negative going threshold voltage		0.3 × V <sub>REF</sub>		$0.6 \times V_{REF}$	V
V <sub>H</sub> ΔVT	Hysteresis voltage (Vp – Vn)		0.1 × V <sub>REF</sub>		$0.4 \times V_{REF}$	V
I <sub>OH_TDO</sub>	High level output current @ Voh = 2.25 V, TDO, Vcc	High level output current @ Voh = 2.25 V, TDO, Vcc = 2.25 V			-2	mA
I <sub>OL_TDO</sub>	Low level output current @ Vol = 0.4 V, TDO, Vcc = 2.25 V				2	mA
TEMPERATURE D	ODE					
ITEMP_DIODE	Max current source into temperature diode <sup>(4)</sup>				120	μA
ENVIRONMENTAL			1		I	
T <sub>ARRAY</sub> <sup>(5)</sup>	Operating DMD array temperature - steady state <sup>(1)</sup>		-40		105	°C
ILL <sub>UV</sub> <sup>(3)</sup>	Illumination, wavelength < 395 nm				2.0	mW/cm <sup>2</sup>
ILL <sub>OVERFILL</sub>	Illumination overfill maximum heat load in area shown in Figure 6-1 $^{(6)}$	<sub>RAY</sub> ≤ 75°C			26	mW/mm <sup>2</sup>
ILL <sub>OVERFILL</sub>	Illumination overfill maximum heat load in area shown in Figure 6-1 $^{(6)}$	<sub>RAY</sub> > 75°C			20	11100/111102

(1) DMD active array temperature can be calculated as shown in Section 7.6 section and assumes uniform illumination across the array.

(2) To prevent excess current, the supply voltage delta  $|V_{BIAS} - V_{OFFSET}|$  must be less than or equal to 8.75 V.

(3) The maximum operation conditions for DMD array temperature and illumination UV shall not be implemented simultaneously.

(4) Temperature diode is to assist in the calculation of the DMD array temperature during operation.

(5) Operating profile information for device micromirror landed duty-cycle and temperature may be provided if requested.

(6) The active area of the DLP3021-Q1 device is surrounded by an aperture on the inside of the DMD window surface that masks structures of the DMD device assembly from normal view. The aperture is sized to anticipate several optical conditions. Overfill light illuminating the area outside the active array can scatter and create adverse effects to the performance of an end application using the DMD. The illumination optical system should be designed to minimize light flux incident outside the active array. Depending on the particular system's optical architecture and assembly tolerances, the amount of overfill light on the outside of the active array may cause system performance degradation. Overfill illumination in excess of this specification may also impact thermal performance.







# 6.5 Thermal Information

THERMAL METRIC <sup>(1)</sup>		DLP3021-Q1	
		FQR (LGA)	UNIT
		64 PINS	
Thermal resistance	Active area to test point 1 (TP1) <sup>(1)</sup>	7.0	°C/W

(1) The DMD is designed to conduct absorbed and dissipated heat to the back of the package. The cooling system must be capable of maintaining the package within the temperature range specified in the Section 6.4. The total heat load on the DMD is largely driven by the incident light absorbed by the active area, although other contributions include light energy absorbed by the window aperture and electrical power dissipation of the array. Optical systems should be designed to minimize the light energy falling outside the window clear aperture since any additional thermal load in this area can significantly degrade the reliability of the device.

# **6.6 Electrical Characteristics**

Over operating free-air temperature range (unless otherwise noted)<sup>(2)</sup>

	PARAMETER	TEST CONDITIONS <sup>(1)</sup>	MIN	TYP	MAX	UNIT
V <sub>OH</sub>	High level output voltage	VCC = 2.25 V	1.7			V
∙он	Thigh level output voltage	I <sub>OH</sub> = -8 mA	1.7			v
V <sub>OH2</sub>	High level output voltage <sup>(6)</sup>	VREF = 1.8 V	1.44	4		V
♥ OH2	Thigh level output voltage	I <sub>OH</sub> = -2 mA	1.44			v
V	Low level output voltage	VCC = 2.75 V		0.4	0.4	V
V <sub>OL</sub>	Low level output voltage	I <sub>OL</sub> = 8 mA			0.4	v
V	Low level output voltage <sup>(6)</sup>	VREF = 1.8 V		0.26	0.36	V
V <sub>OL2</sub>		I <sub>OL</sub> = 2 mA			0.30	v
		VREF = 1.95 V	-10			
1	Output high impedance current	$V_{OL} = 0 V$	-10			
I <sub>OZ</sub>	Output high impedance current	VREF = 1.95 V			10	μA
		V <sub>OH</sub> = VREF			10	
	Low lovel input ourrept(3)	VREF = 1.95 V	E			
IIL	Low level input current <sup>(3)</sup>	$V_1 = 0 V$	-5	-5		μA
l <sub>IH</sub>	High level input current <sup>(3)</sup>	VREF = 1.95 V			6	
		V <sub>I</sub> = VREF		6		μA



### 6.6 Electrical Characteristics (continued)

Over operating free-air temperature range (unless otherwise noted)<sup>(2)</sup>

	PARAMETER	TEST CONDITIONS <sup>(1)</sup>	MIN	TYP	MAX	UNIT
1	Low lovel input current <sup>(4)</sup>	VREF = 1.95 V	-785			
I <sub>IL2</sub>	Low level input current <sup>(4)</sup>	V <sub>1</sub> = 0 V	-700			μA
1	Lligh lovel input surrent <sup>(4)</sup>	VREF = 1.95 V				
I <sub>IH2</sub>	High level input current <sup>(4)</sup>	V <sub>I</sub> = VREF			6	μA
1	Low level input current <sup>(5)</sup>	VREF = 1.95 V	-5			
I <sub>IL3</sub>	Low level input current(*)	V <sub>1</sub> = 0 V	-5			μA
1	High level input current <sup>(5)</sup>	VREF = 1.95 V			785 µA	
I <sub>IH3</sub>	nigh level linput current."	V <sub>I</sub> = VREF		785		μΑ
CURRENT					·	
I <sub>REF</sub>	Current at V <sub>REF</sub> = 1.95 V	f <sub>DCLK</sub> = 80 MHz			2.80	mA
I <sub>cc</sub>	Current at $V_{CC}$ = 2.75 V	f <sub>DCLK</sub> = 80 MHz			59.90	mA
IOFFSET	Current at V <sub>OFFSET</sub> = 8.75 V				2.93	mA
I <sub>BIAS</sub>	Current at V <sub>BIAS</sub> = 16.5 V				2.30	mA
I <sub>RESET</sub>	Current at V <sub>RESET</sub> = –10.5 V				-2.00	mA
POWER		i i				
P <sub>REF</sub>	Power at V <sub>REF</sub> = 1.95 V	f <sub>DCLK</sub> = 80 MHz			5.46	mW
P <sub>CC</sub>	Power at V <sub>CC</sub> = 2.75 V	f <sub>DCLK</sub> = 80 MHz			164.73	mW
POFFSET	Power at V <sub>OFFSET</sub> = 8.75 V				25.64	mW
P <sub>BIAS</sub>	Power at V <sub>BIAS</sub> = 16.5 V				37.95	mW
P <sub>RESET</sub>	Power at V <sub>RESET</sub> = –10.5 V				21.00	mW
P <sub>TOTAL</sub>	Total power at nominal conditions	f <sub>DCLK</sub> = 80 MHz			254.77	mW
CAPACITA	NCE				I	
C <sub>IN</sub>	Input pin capacitance	f = 1 MHz			20	pF
C <sub>A</sub>	Analog pin capacitance (TEMP_PLUS and TEMP_MINUS pins)	f = 1 MHz			65	pF
Co	Output pin capacitance	f = 1 MHz			20	pF

(1) All voltage values are with respect to the ground pins (V<sub>SS</sub>).

(2) Device electrical characteristics are over <u>Section 6.4</u> unless otherwise noted.

(3) Specification is for LVCMOS input pins, which do not have pull up or pull down resistors. See Section 5 section.

(4) Specification is for LVCMOS input pins which do have pull up resistors (JTAG: TDI, TMS). See Section 5 section.

(5) Specification is for LVCMOS input pins which do have pull down resistors. See Section 5 section.

(6) Specification is for LVCMOS JTAG output pin TDO.

# 6.7 Timing Requirements

Over Section 6.4 unless otherwise noted.

		MIN	NOM MAX	UNIT
DMD N	MIRROR AND SRAM CONTROL LOGIC SIGNALS			
t <sub>SU</sub>	Setup time SAC_BUS low before SAC_CLK↑	1.0		ns
t <sub>H</sub>	Hold time SAC_BUS low after SAC_CLK↑	1.0		ns
t <sub>SU</sub>	Setup time DAD_BUS high before SAC_CLK↑	1.0		ns
t <sub>H</sub>	Hold time DAD_BUS after SAC_CLK↑	1.0		ns
t <sub>C</sub>	Cycle time SAC_CLK	12.5	16.67	ns
t <sub>W</sub>	Pulse width 50% to 50% reference points: SAC_CLK high or low	5.0		ns
t <sub>R</sub>	Rise time 20% to 80% reference points: SAC_CLK		2.5	ns
t <sub>F</sub>	Fall time 80% to 20% reference points: SAC_CLK		2.5	ns
DMD D	DATA PATH AND LOGIC CONTROL SIGNALS			



Over Section 6.4 unless otherwise noted.

		MIN	NOM	MAX	UNIT
t <sub>SU</sub>	Setup time DATA(14:0) before DCLK $\uparrow$ or DCLK $\downarrow$	1.0			ns
t <sub>H</sub>	Hold time DATA(14:0) after DCLK↑ or DCLK↓	1.0	·		ns
t <sub>SU</sub>	Setup time SCTRL before DCLK $\uparrow$ or DCLK $\downarrow$	1.0			ns
t <sub>H</sub>	Hold time SCTRL after DCLK↑ or DCLK↓	1.0			ns
t <sub>SU</sub>	Setup time TRC before DCLK $\uparrow$ or DCLK $\downarrow$	1.0	·		ns
t <sub>H</sub>	Hold time TRC after DCLK↑ or DCLK↓	1.0	·		ns
t <sub>SU</sub>	Setup time LOADB low before DCLK↑	1.0	·		ns
t <sub>H</sub>	Hold time LOADB low after DCLK↓	1.0			ns
t <sub>SU</sub>	Setup time RESET_STROBE high before DCLK↑	1.0			ns
t <sub>H</sub>	Hold time RESET_STROBE after DCLK↑	3.5			ns
t <sub>C</sub>	Cycle time DCLK	12.5		16.67	ns
t <sub>W</sub>	Pulse width 50% to 50% reference points: DCLK high or low	5.0	·		ns
t <sub>W</sub> (L)	Pulse width 50% to 50% reference points: LOADB low	7.0			ns
t <sub>W</sub> (H)	Pulse width 50% to 50% reference points: RESET_STROBE high	7.0			ns
t <sub>R</sub>	Rise time 20% to 80% reference points: DCLK, DATA, SCTRL, TRC, LOADB			2.5	ns
t <sub>F</sub>	Fall time 80% to 20% reference points: DCLK, DATA, SCTRL, TRC, LOADB			2.5	ns
JTAG B	OUNDARY SCAN CONTROL LOGIC SIGNALS				
f <sub>TCK</sub>	Clock frequency TCK			10	MHz
t <sub>C</sub>	Cycle time TCK	100			ns
t <sub>W</sub>	Pulse width 50% to 50% reference points: TCK high or low	10			ns
t <sub>SU</sub>	Setup time TDI valid before TCK↑	5			ns
t <sub>H</sub>	Hold time TDI valid after TCK↑	25			ns
t <sub>SU</sub>	Setup time TMS valid before TCK↑	5			ns
t <sub>H</sub>	Hold time TMS valid after TCK↑	25			ns
t <sub>R</sub>	Rise time 20% to 80% reference points: TCK, TDI, TMS			2.5	ns
t <sub>R</sub>	Fall time 80% to 20% reference points: TCK, TDI, TMS			2.5	ns







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Figure 6-2. DMD Mirror and SRAM Control Logic Timing Requirements

 $V_{\text{SS}}$ 





Figure 6-3. DMD Data Path and Control Logic Timing Requirements





Figure 6-4. JTAG Boundary Scan Control Logic Timing Requirements

t<sub>F</sub>→↔

t<sub>R</sub>→→



### 6.8 Switching Characteristics

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT			
t <sub>PD</sub>	Output propagation, clock to Q (see Figure 6-4)	C <sub>L</sub> = 11 pF, from (Input) falling edge of TCK to (Output) TDO, see Figure 6-4	3		25	ns			
	Data Sheet Timing Reference Point								
	Ŭ ↓								
	Device Pir Output Under Tes	> Lester Channe	əl						

Over operating free-air temperature range (unless otherwise noted).

See Section 7.3.1 section for more information.

### Figure 6-5. Test Load Circuit for Output Propagation Measurement

# 6.9 System Mounting Interface Loads

PARAMETER	MIN	NOM	MAX	UNIT
Uniformly distributed within the Thermal Interface Area shown in Figure 6-6			70	Ν
Uniformly distributed within the Electrical Interface Area shown in Figure 6-6			100	Ν



Thermal Interface Area

Figure 6-6. System Interface Loads



# 6.10 Physical Characteristics of the Micromirror Array

PARAMETI	ER	VALUE	UNIT
N Number of active columns	See Figure 6-7	684	micromirrors
M Number of active rows	See Figure 6-7	608	micromirrors
ε Micromirror (pixel) pitch – diagonal	See Figure 6-8	7.6	μm
P Micromirror (pixel) pitch – horizontal and vertical	See Figure 6-8	10.8	μm
Micromirror active array width	P × M + P / 2; see Figure 6-7	6.5718	mm
Micromirror active array height	(P × N) / 2 + P / 2; see Figure 6-7	3.699	mm
Micromirror active border	Pond of micromirror (POM) <sup>(1)</sup>	10	micromirrors/side

(1) The structure and qualities of the border around the active array includes a band of partially functional micromirrors called the POM. These micromirrors are structurally and/or electrically prevented from tilting toward the bright or ON state, but still require an electrical bias to tilt toward OFF.





Figure 6-7. Micromirror Array Physical Characteristics





# Figure 6-8. Mirror (Pixel) Pitch

## 6.11 Micromirror Array Optical Characteristics

#### **Table 6-1. Optical Parameters**

PARAMETER	MIN	NOM	MAX	UNIT
Micromirror tilt angle, landed (on-state or off-state) <sup>(1)</sup> ,		12		0
Micromirror tilt angle tolerance <sup>(1)</sup> ,	-1		1	0
DMD efficiency, 420 nm – 680 nm <sup>(2)</sup>		66%		

(1) For some applications, it is critical to account for the micromirror tilt angle variation in the overall optical system design. With some optical system designs, the micromirror tilt angle variation within a device may result in perceivable non-uniformities in the light field reflected from the micromirror array. With some optical system designs, the micromirror tilt angle variations, the micromirror tilt angle variations, or system contrast variations.

(2) DMD efficiency is measured photopically under the following conditions: 24° illumination angle, F/2.4 illumination and collection apertures, uniform source spectrum (halogen), uniform pupil illumination, the optical system is telecentric at the DMD, and the efficiency numbers are measured with 100% electronic mirror duty cycle and do not include system optical efficiency or overfill loss. Note that this number is measured under conditions described above and deviations from these specified conditions could result in a different efficiency value in a different optical system. The factors that can influence the DMD efficiency related to system application include: light source spectral distribution and diffraction efficiency at those wavelengths (especially with discrete light sources such as LEDs or lasers), and illumination and collection apertures (F/#) and diffraction efficiency. The interaction of these system factors as well as the DMD efficiency factors that are not system dependent are described in detail in DMD Optical Efficiency for Visible Wavelengths.



### **6.12 Window Characteristics**

PAR	AMETER	MIN	NOM	MAX	UNIT
Window material designation		Cor	ning Eagle XG		
Window refractive index	at wavelength 546.1 nm		1.5119		
Window aperture <sup>(1)</sup>				See <sup>(1)</sup>	

(1) See the package mechanical ICD for details regarding the size and location of the window aperture.

### 6.13 Chipset Component Usage Specification

The DLP3021-Q1 DMD is a component of a DLP® chipset including a DLP products controller. Reliable function and operation of the DMD requires that it be used in conjunction with a DLP products controller.

Note

TI assumes no responsibility for image quality artifacts or DMD failures caused by optical system operating conditions exceeding limits described previously



# 7 Detailed Description

# 7.1 Overview

The DLP3021-Q1 DMD has a resolution of 608 × 684 mirrors configured in a diamond format that results in an aspect ratio of 16:9 which creates an effective resolution of 864 × 480 square pixels. By configuring the pixels in a diamond format, the illumination input to the DMD enters from the side allowing for smaller mechanical packaging of the optical system.

# 7.2 Functional Block Diagram





# 7.3 Feature Description

To ensure reliable operation, the DLP3021-Q1 DMD must be used with a DLP products controller.

### 7.3.1 Micromirror Array

The DLP3021-Q1 DMD consists of a two-dimensional array of 1-bit CMOS memory cells that determine the state of the each of the 608 × 684 micromirrors in the array. Refer to Section 6.10 for a calculation of how the 608 × 684 micromirror array represents a 16:9 dimensional aspect ratio to the user. Each micromirror is either "ON" (tilted +12°) or "OFF" (tilted  $-12^\circ$ ). Combined with appropriate projection optical system the DMD can be used to create sharp, colorful, and vivid digital images.

# 7.3.2 Double Data Rate (DDR) Interface

Each DMD micromirror and its associated SRAM memory cell is loaded with data from the DLP controller via the DDR interface (DATA(14:0), DCLK, LOADB, SCRTL, and TRC). These signals are low voltage CMOS nominally operating at 1.8-V level to reduce power and switching noise. This high speed data input to the DMD allows for a maximum update rate of the entire micromirror array to be nearly 5 kHz, enabling the creation of seamless digital images using Pulse Width Modulation (PWM).

#### 7.3.3 Micromirror Switching Control

Once data is loaded onto the DMD, the mirrors switch position  $(+12^{\circ} \text{ or } -12^{\circ})$  based on the timing signal sent to the DMD Mirror and SRAM control logic. The DMD mirrors will be switched from OFF to ON or ON to OFF, or stay in the same position based on control signals DAD\_BUS, RESET\_STROBE, SAC\_BUS, and SAC\_CLK, which are coordinated with the data loading by the DLP controller. In general, the DLP controller loads the DMD SRAM memory cells over the DDR interface, and then commands to the micromirrors to switch position.

At power down, the DMD Mirrors are commanded by the DLP controller to move to a near flat (0°) position as shown in Section 9. The flat state position of the DMD mirrors are referred to as the "Parked" state. To maintain long-term DMD reliability, the DMD must be properly "Parked" prior to every power down of the DMD power supplies.

### 7.3.4 DMD Voltage Supplies

The micromirrors switching requires unique voltage levels to control the mechanical switching. These voltages levels are nominally 16 V, 8.5 V, and –10 V ( $V_{BIAS}$ ,  $V_{OFFSET}$ , and  $V_{RESET}$ ). The specification values for  $V_{BIAS}$ ,  $V_{OFFSET}$ , and  $V_{RESET}$  are shown in Section 6.4.

### 7.3.5 Logic Reset

Reset of the DMD is required and controlled by the DLP products controller.

### 7.3.6 Temperature Sensing Diode

The DMD includes a temperature sensing diode designed to be used with the TMP411-Q1 temperature monitoring device. The DLP products controller may monitor the DMD array temperature via the TMP411-Q1 and temperature sense diode.

Figure 7-1 shows the typical connection between the DLP products controller, TMP411-Q1, and the DLP3021-Q1 DMD. The signals to the temperature sense diode are sensitive to system noise, and care should be taken in the routing and implementation of this circuit. See the *TMP411-Q1 data sheet* for detailed PCB layout recommendations.





Figure 7-1. Temperature Sense Diode Typical Circuit Configuration

It is recommended that the host controller manage parking of the DMD based on the allowable temperature specifications and temperature measurements.

### 7.3.6.1 Temperature Sense Diode Theory

A temperature sensing diode is based on the fundamental current and temperature characteristics of a transistor. The diode is formed by connecting the transistor base to the collector. Two different known currents flow through the diode and the resulting diode voltage is measured in each case. The difference in the base-emitter voltages is proportional to the absolute temperature of the transistor.

Refer to the *TMP411-Q1 data sheet* for detailed information about temperature diode theory and measurement. Figure 7-2 and Figure 7-3 illustrate the relationship between the current and voltage through the diode.



Figure 7-2. Temperature Measurement Theory







#### 7.3.7 DMD JTAG Interface

The DMD uses 4 standard JTAG signals for sending and receiving boundary scan test data. TCK is the test clock used to drive an IEEE 1149.1 TAP state machine and logic. TMS directs the next state of the TAP state machine. TDI is the scan data input and TDO is the scan data output.

The DMD does not support IEEE 1149.1 signals TRST (Test Logic Reset) and RTCK (Returned Test Clock). Boundary scan cells on the DMD are Observe-Only. To initiate the JTAG boundary scan operation on the DMD, a minimum of 6 TCK clock cycles are required after TMS is set to logic high.

Refer to Figure 7-4 for a JTAG system board routing example.





The DMD Device ID can be read via the JTAG interface. The ID and 32-bit shift order is shown in Figure 7-5.



Figure 7-5. DMD Device ID and 32-bit Shift Order

Refer to Figure 7-6 for a JTAG boundary scan block diagram for the DMD. These show the pins and the scan order that are observed during the JTAG boundary scan.



### Figure 7-6. JTAG Boundary Scan Path

Refer to Figure 7-7 for a functional block diagram of the JTAG control logic.









# 7.4 System Optical Considerations

Optimizing system optical performance and image performance strongly relates to optical system design parameter trades. Although it is not possible to anticipate every conceivable application, projector image quality and optical performance is contingent on compliance to the optical system operating conditions described in the following sections.

#### 7.4.1 Numerical Aperture and Stray Light Control

The angle defined by the numerical aperture of the illumination and projection optics at the DMD optical area should be the same. This angle should not exceed the nominal device mirror tilt angle unless appropriate apertures are added in the illumination and/or projection pupils to block flat-state and stray light from passing through the projection lens. The mirror tilt angle defines DMD capability to separate the "On" optical path from any other light path, including undesirable flat-state specular reflections from the DMD window, DMD border structures, or other system surfaces near the DMD such as prism or lens surfaces. If the numerical aperture exceeds the mirror tilt angle, or if the projection numerical aperture angle is more than two degrees larger than the illumination numerical aperture angle, contrast ratio can be reduced and objectionable artifacts in the image border and/or active area could occur.

#### 7.4.2 Pupil Match

TI's optical and image quality specifications assume that the exit pupil of the illumination optics is nominally centered within two degrees of the entrance pupil of the projection optics. Misalignment of pupils can create objectionable artifacts in the image border and/or active area, which may require additional system apertures to control, especially if the numerical aperture of the system exceeds the pixel tilt angle.

#### 7.4.3 Illumination Overfill and Alignment

Overfill light illuminating the area outside the active array can create artifacts from the mechanical features and other surfaces that surround the active array. These artifacts may be visible in the projected image. The illumination optical system should be designed to minimize light flux incident outside the active array and on the window aperture. Depending on the particular system's optical architecture and assembly tolerances, this amount of overfill light on the area outside of the active array may still cause artifacts to be visible. Illumination light and overfill can also induce undesirable thermal conditions on the DMD, especially if illumination light impinges directly on the DMD window aperture or near the edge of the DMD window. Refer to Section 6.4 for a specification on this maximum allowable heat load due to illumination overfill.

### 7.5 DMD Image Performance Specification

PARAMETER	MIN	NOM	MAX	UNIT	
Number of non-operational micromirrors <sup>(1)</sup>	Adjacent micromirrors			0	mioromirroro
	Non-adjacent micromirrors			10	micromirrors
Optical performance		See	Section 7.4		

(1) A non-operational micromirror is defined as a micromirror that is unable to transition between the on-state and off-state positions.

### 7.6 Micromirror Array Temperature Calculation

Active array temperature can be computed analytically from measurement points on the outside of the package, the package thermal resistance, the electrical power, and the illumination heat load.

Relationship between array temperature and the reference ceramic temperature (thermocouple location TP1 in Figure 7-8) is provided by the following equations.

T <sub>ARRAY</sub> = T <sub>CERAMIC</sub> + (Q <sub>ARRAY</sub> × R <sub>ARRAY-TO-CERAMIC</sub> )	(1)	)
---	-----	---

 $Q_{ARRAY} = Q_{ELECTRICAL} + Q_{ILLUMINATION}$ 

#### where

T<sub>ARRAY</sub> = computed DMD array temperature (°C)

(2)



- T<sub>CERAMIC</sub> = measured ceramic temperature (TP1 location in Figure 7-8) (°C)
- R<sub>ARRAY-TO-CERAMIC</sub> = DMD package thermal resistance from array to TP1 (°C/watt) (see Section 6.5)
- Q<sub>ARRAY</sub> = total power, electrical plus absorbed, on the DMD array (watts)
- Q<sub>ELECTRICAL</sub> = nominal electrical power dissipation by the DMD (watts)
- $Q_{ILLUMINATION} = (C_{L2W} \times S_L)$
- $C_{L2W}$  = conversion constant for screen lumens to power on the DMD (watts/lumen)
- S<sub>L</sub> = measured screen lumens (lm)

Electrical power dissipation of the DMD is variable and depends on the voltages, data rates, and operating frequencies.

Absorbed power from the illumination source is variable and depends on the operating state of the mirrors and the intensity of the light source.

Equations shown previous are valid for a 1-Chip DMD system with total projection efficiency from DMD to the screen of 87%.

The constant  $C_{L2W}$  is based on the DMD array characteristics. It assumes a spectral efficiency of 300 lumens/ watt for the projected light and illumination distribution of 83.7% on the active array, and 16.3% on the array border.

Sample calculation:

- S<sub>L</sub> = 50 lm
- C<sub>L2W</sub> = 0.00293 W/lm
- $Q_{ELECTRICAL} = 0.162 W$
- R<sub>ARRAY-TO-CERAMIC</sub> = 7.0°C/W
- T<sub>CERAMIC</sub> = 55°C

```
Q_{ARRAY} = 0.162 \text{ W} + (0.00293 \times 50 \text{ Im}) = 0.309 \text{ W} (3)
```

T<sub>ARRAY</sub> = 55°C + (0.309 W × 7.0°C/W) = 57.2°C

(4)





Figure 7-8. Thermocouple Location

# 7.7 Micromirror Landed-On/Landed-Off Duty Cycle

The micromirror landed-on/landed-off duty cycle (landed duty cycle) denotes the amount of time (as a percentage) that an individual micromirror is landed in the ON state versus the amount of time the same micromirror is landed in the OFF state.

As an example, assuming a fully-saturated white pixel, a landed duty cycle of 90/10 indicates that the referenced pixel is in the ON state 90% of the time (and in the OFF state 10% of the time), whereas 10/90 would indicate that the pixel is in the OFF state 90% of the time. Likewise, 50/50 indicates that the pixel is ON 50% of the time and OFF 50% of the time.

Note that when assessing landed duty cycle, the time spent switching from one state (ON or OFF) to the other state (OFF or ON) is considered negligible and is thus ignored.

Since a micromirror can only be landed in one state or the other (ON or OFF), the two numbers (percentages) always add to 100.



# 8 Application and Implementation

Note

Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes, as well as validating and testing their design implementation to confirm system functionality.

# 8.1 Application Information

The DLP3021-Q1 DMD was designed to be used in automotive applications such as dynamic ground projection. The information shown in this section describes the dynamic ground projection application.

## **8.2 Typical Application**

The DLP3021-Q1 DMD combined with a DLP products controller are the primary devices that make up the reference design for a dynamic ground projection system as shown in the block diagram Figure 8-1.



Figure 8-1. Dynamic Ground Projection Reference Design Block Diagram

In this architecture, video content is compressed and stored in external flash memory. Low speed SPI commands are sent from a microcontroller or other processor to the DLP products controller to indicate what video content to read from external memory. Storing the video content in memory removes the need for a high speed video interface to the module which improves compatibility with typical vehicle infrastructures. It also decreases overall system size and cost by removing graphics generation and interfaces. The controller decompresses each bit plane of the video data (608 × 684 resolution) and displays them on the DMD in rapid succession to create the full video image. Due to the diamond format of the DMD pixels, the output image has an effective resolution of 864 × 480. The controller synchronizes the DMD bit plane data with the RGB enable timing for the LED color controller and driver circuit.

The controller may connect to a TMP411-Q1 to measure the DLP3021-Q1 temperature using the built-in temperature sensing diode.



The controller combined with the DLP3021-Q1 may be used in RGB LED or laser illumination systems, or in single-color systems as shown in Figure 8-2.



### Figure 8-2. Dynamic Ground Projection Reference Design Block Diagram - Single Color

### 8.3 Application Mission Profile Consideration

Each application is anticipated to have different mission profiles, or number of operating hours at different temperatures. To assist in evaluation, the automotive DMD reliability lifetime estimates Application Report may be provided. See the TI Application team for more information.



# 9 Power Supply Recommendations

# 9.1 Power Supply Sequencing Requirements

• V<sub>BIAS</sub>, V<sub>CC</sub>, V<sub>OFFSET</sub>, V<sub>REF</sub>, V<sub>RESET</sub>, V<sub>SS</sub> are required to operate the DMD.

# CAUTION

- For reliable operation of the DMD, the following power supply sequencing requirements must be followed. Failure to adhere to the prescribed power up and power down procedures may affect device reliability.
- The V<sub>CC</sub>, V<sub>REF</sub>, V<sub>OFFSET</sub>, V<sub>BIAS</sub>, and V<sub>RESET</sub> power supplies have to be coordinated during power up and power down operations. Failure to meet any of the following requirements will result in a significant reduction in the DMD's reliability and lifetime. Refer to Figure 9-1. V<sub>SS</sub> must also be connected.

### DMD Power Supply Power Up Procedure:

- During power up, V<sub>CC</sub> and V<sub>REF</sub> must always start and settle before V<sub>OFFSET</sub>, V<sub>BIAS</sub> and V<sub>RESET</sub> voltages are applied to the DMD.
- During power up, V<sub>BIAS</sub> does not have to start after V<sub>OFFSET</sub>. However, it is a strict requirement that the delta between V<sub>BIAS</sub> and V<sub>OFFSET</sub> must be within ±8.75 V (refer to Note 1 for Figure 9-1).
- During power up, the DMD's LVCMOS input pins shall not be driven high until after V<sub>CC</sub> and V<sub>REF</sub> have settled at operating voltage.
- During power up, there is no requirement for the relative timing of V<sub>RESET</sub> with respect to V<sub>OFFSET</sub> and V<sub>BIAS</sub>.
- Power supply slew rates during power up are flexible, provided that the transient voltage levels follow the requirements listed previously in Section 6.4 and in Figure 9-1.

#### **DMD Power Supply Power Down Procedure**

- V<sub>CC</sub> and V<sub>REF</sub> must be supplied until after V<sub>BIAS</sub>, V<sub>RESET</sub>, and V<sub>OFFSET</sub> are discharged to within 4 V of ground.
- During power down it is not mandatory to stop driving V<sub>BIAS</sub> prior to V<sub>OFFSET</sub>, but it is a strict requirement that the delta between V<sub>BIAS</sub> and V<sub>OFFSET</sub> must be within ±8.75 V (refer to Note 1 for Figure 9-1).
- During power down, the DMD's LVCMOS input pins must be less than V<sub>REF</sub> + 0.3 V.
- During power down, there is no requirement for the relative timing of V<sub>RESET</sub> with respect to V<sub>OFFSET</sub> and V<sub>BIAS</sub>.
- Power supply slew rates during power down are flexible, provided that the transient voltage levels follow the requirements listed previously in Section 6.4 and in Figure 9-1.



#### 9.1.1 Power Up and Power Down



A.  $\pm 8.75$ -V delta,  $\Delta V$ , shall be considered the max operating delta between V<sub>BIAS</sub> and V<sub>OFFSET</sub>. Customers may find that the most reliable way to ensure this is to power V<sub>OFFSET</sub> prior to V<sub>BIAS</sub> during power up and to remove V<sub>BIAS</sub> prior to V<sub>OFFSET</sub> during power down.

### Figure 9-1. Power Supply Sequencing Requirements (Power Up and Power Down)



# 10 Layout

# **10.1 Layout Guidelines**

Refer to the DMD controller and system management controller datasheets for specific PCB layout and routing guidelines. For specific DMD PCB guidelines, use the following:

- $V_{CC}$  should have at least 1 × 2.2-µF and 4 × 0.1-µF capacitors evenly distributed among the 13  $V_{CC}$  pins.
- A 0.1-µF, X7R rated capacitor should be placed near every pin for the V<sub>REF</sub>, V<sub>BIAS</sub>, V<sub>RSET</sub>, and V<sub>OFF</sub>.

### **10.2 Temperature Diode Pins**

The DMD has an internal diode (PN junction) that is intended to be used with an external TI TMP411-Q1 temperature sensing IC. PCB traces from the DMD's temperature diode pins to the TMP411-Q1 are sensitive to noise. See the *TMP411-Q1 data sheet* for specific routing recommendations.

Avoid routing the temperature diodes signals near other traces to reduce coupling of noise onto these signals.



# **11 Device and Documentation Support**

### **11.1 Device Support**

11.1.1 Device Nomenclature



Figure 11-1. Part Number Description

#### 11.1.2 Device Markings

The device marking is shown in Figure 11-2. The marking will include both human-readable information and a 2-dimensional matrix code.

The human-readable information is described in Figure 11-2. The 2-dimensional matrix code is an alpha-numeric character string that contains the DMD part number and lot trace code.



Figure 11-2. DMD Marking



# **11.2 Documentation Support**

# 11.2.1 Related Documentation

For related documentation see the following:

- Texas Instruments, *TMP411-Q1* ±1°C *Remote and Local Temperature Sensor With N-Factor and Series Resistance Correction* data sheet
- Texas Instruments, DMD Optical Efficiency for Visible Wavelengths application report

### **11.3 Receiving Notification of Documentation Updates**

To receive notification of documentation updates, navigate to the device product folder on ti.com. Click on *Subscribe to updates* to register and receive a weekly digest of any product information that has changed. For change details, review the revision history included in any revised document.

## **11.4 Support Resources**

TI E2E<sup>™</sup> support forums are an engineer's go-to source for fast, verified answers and design help — straight from the experts. Search existing answers or ask your own question to get the quick design help you need.

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#### **11.6 Electrostatic Discharge Caution**



This integrated circuit can be damaged by ESD. Texas Instruments recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

# **11.7 Device Handling**

The DMD is an optical device so precautions should be taken to avoid damaging the glass window. Please see the DMD Handling application note for instructions on how to properly handle the DMD.

#### 11.8 Glossary

TI Glossary This glossary lists and explains terms, acronyms, and definitions.

# 12 Mechanical, Packaging, and Orderable Information

The following pages include mechanical, packaging, and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.



# PACKAGING INFORMATION

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead finish/ Ball material	MSL Peak Temp (3)	Op Temp (°C)	Device Marking (4/5)	Samples
DLP3021FFQRQ1	ACTIVE	CLGA	FQR	54	126	RoHS & Green	Call TI	N / A for Pkg Type	-40 to 105		Samples

<sup>(1)</sup> The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

**LIFEBUY:** TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

**PREVIEW:** Device has been announced but is not in production. Samples may or may not be available.

**OBSOLETE:** TI has discontinued the production of the device.

<sup>(2)</sup> RoHS: TI defines "RoHS" to mean semiconductor products that are compliant with the current EU RoHS requirements for all 10 RoHS substances, including the requirement that RoHS substance do not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, "RoHS" products are suitable for use in specified lead-free processes. TI may reference these types of products as "Pb-Free".

**RoHS Exempt:** TI defines "RoHS Exempt" to mean products that contain lead but are compliant with EU RoHS pursuant to a specific EU RoHS exemption.

Green: TI defines "Green" to mean the content of Chlorine (CI) and Bromine (Br) based flame retardants meet JS709B low halogen requirements of <=1000ppm threshold. Antimony trioxide based flame retardants must also meet the <=1000ppm threshold requirement.

<sup>(3)</sup> MSL, Peak Temp. - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

<sup>(4)</sup> There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.

<sup>(5)</sup> Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.

<sup>(6)</sup> Lead finish/Ball material - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

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