# **Programmable Precision References**

The NCP431/NCP432 integrated circuits are three-terminal programmable shunt regulator diodes. These monolithic IC voltage references operate as a low temperature coefficient zener which is programmable from Vref to 36 V using two external resistors. These devices exhibit a wide operating current range of 40  $\mu$ A to 100 mA with a typical dynamic impedance of 0.22  $\Omega$ . The characteristics of these references make them excellent replacements for zener diodes in many applications such as digital voltmeters, power supplies, and op amp circuitry. The 2.5 V reference makes it convenient to obtain a stable reference from 5.0 V logic supplies, and since the NCP431/NCP432 operates as a shunt regulator, it can be used as either a positive or negative voltage reference. Low minimum operating current makes this device an ideal choice for secondary regulators in SMPS adapters with extremely low no–load consumption.

#### Features

- Programmable Output Voltage to 36 V
- Low Minimum Operating Current: 40 μA, Typ @ 25°C
- Voltage Reference Tolerance: ±0.5%, Typ @ 25°C (NCP431B/NCP432B)
- Low Dynamic Output Impedance,  $0.22 \Omega$  Typical
- Sink Current Capability of 40 µA to 100 mA
- Equivalent Full–Range Temperature Coefficient of 50 ppm/°C Typical
- Temperature Compensated for Operation over Full Rated Operating Temperature Range
- SC Prefix for Automotive and Other Applications Requiring Unique Site and Control Change Requirements; AEC–Q100 Qualified and PPAP Capable
- These are Pb–Free Devices

#### **Typical Applications**

- Voltage Adapters
- Switching Power Supply
- Precision Voltage Reference
- Charger
- Instrumentation



## **ON Semiconductor®**





#### **ORDERING AND MARKING INFORMATION**

See detailed ordering and shipping information in the package dimensions section on page 14 of this data sheet.



This device contains 20 active transistors

#### **MAXIMUM RATINGS** (Full operating ambient temperature range applies, unless otherwise noted)

Symbol	Rating	Value	Unit
V <sub>KA</sub>	Cathode to Anode Voltage	37	V
١ <sub>K</sub>	Cathode Current Range, Continuous	-100 to +150	mA
I <sub>ref</sub>	Reference Input Current Range, Continuous	–5 to +10	mA
ТJ	Operating Junction Temperature	150	°C
T <sub>A</sub>	Operating Ambient Temperature Range	-40 to +125	°C
T <sub>stg</sub>	Storage Temperature Range	-65 to +150	°C
P <sub>D</sub>	Total Power Dissipation @ T <sub>A</sub> = 25°C Derate above 25°C Ambient Temperature D, LP Suffix Plastic Package SN1 Suffix Plastic Package	0.70 0.52	W
PD	Total Power Dissipation @ T <sub>C</sub> = 25°C Derate above 25°C Case Temperature D, LP Suffix Plastic Package	1.5	W
HBM CDM	ESD Rating (Note 1) Human Body Model per JEDEC JESD22–A114F Charged Device Model per JEDEC JESD22–C101E	>2000 >1000	V

Stresses exceeding those listed in the Maximum Ratings table may damage the device. If any of these limits are exceeded, device functionality should not be assumed, damage may occur and reliability may be affected.

1. This device contains latch-up protection and exceeds ±100 mA per JEDEC standard JESD78.

#### **RECOMMENDED OPERATING CONDITIONS**

Symbol	Condition	Min	Max	Unit
V <sub>KA</sub>	Cathode to Anode Voltage	V <sub>ref</sub>	36	V
۱ <sub>K</sub>	Cathode Current	0.04	100	mA

Functional operation above the stresses listed in the Recommended Operating Ranges is not implied. Extended exposure to stresses beyond the Recommended Operating Ranges limits may affect device reliability.

## THERMAL CHARACTERISTICS

Symbol	Characteristic	LP Suffix Package (50 mm <sup>2</sup> x 35 μm Cu)	D Suffix Package (50 mm <sup>2</sup> x 35 μm Cu)	SN Suffix Package (10 mm <sup>2</sup> x 35 μm Cu)	Unit
$R_{\ThetaJA}$	Thermal Resistance, Junction-to-Ambient	176	210	255	°C/W
$R_{\Theta JL}$	Thermal Resistance, Junction-to-Lead (Lead 3)	75	68	80	°C/W

			NCP431AC		NCP431AI		NCP431AV/ SC431AV				
Symbol	Characteristic	Min	Тур	Max	Min	Тур	Max	Min	Тур	Max	Unit
V <sub>ref</sub>	Reference Input Voltage $V_{KA} = V_{ref}, I_K = 1 \text{ mA}$ $T_A = 25^{\circ}C$ $T_A = T_{low}$ to $T_{high}$ (Figure 3, Note 2)		2.500 2.500	2.525 2.525	2.475 2.465		2.525 2.525	2.475 2.460	2.500 2.500	2.525 2.525	V
$\Delta V_{refT}$	Reference Input Voltage Deviation Over Temperature Range (Figure 3, Notes 3, 4) $V_{KA}$ = $V_{ref}$ , $I_K$ = 1 mA	-	_	_	-	5.0	10	-	10	15	mV
$\frac{\Delta V_{\text{ref}}}{\Delta V_{\text{KA}}}$	Ratio of Change in Reference Input Voltage to Change in Cathode to Anode Voltage $I_K = 1 \text{ mA}$ (Figure 4), $\Delta V_{KA} = 10 \text{ V to } V_{ref}$ $\Delta V_{KA} = 36 \text{ V to } 10 \text{ V}$		-1.85 -0.80	-3.1 -1.8		-1.85 -0.80	-3.1 -1.8		-1.85 -0.80	-3.1 -1.8	mV/ V
I <sub>ref</sub>	Reference Input Current (Figure 4) $I_K = 1 \text{ mA}, R1 = 220 \text{ k}, R2 = \infty$ $T_A = -40^{\circ}\text{C} \text{ to } +125^{\circ}\text{C}$	-	81	190	_	81	190	-	81	190	nA
$\Delta I_{refT}$	Reference Input Current Deviation Over Temperature Range (Figure 4, Note 3) $I_K = 1 \text{ mA}, R1 = 10 \text{ k}, R2 = \infty$	-	22	55	_	22	55	-	22	55	nA
I <sub>min</sub>	Minimum Cathode Current For Regulation $V_{KA} = V_{ref}$ (Figure 3)	_	40	60	-	40	60	-	40	60	μΑ
I <sub>off</sub>	Off-State Cathode Current (Figure 5) $V_{KA} = 36 V, V_{ref} = 0 V$	-	180	1000	-	180	1000	-	180	1000	nA
Z <sub>KA</sub>	Dynamic Impedance (Figure 3, Note 5) $V_{KA} = V_{ref}, \Delta I_{K} = 1.0 \text{ mA to } 100 \text{ mA}$ f $\leq 1.0 \text{ kHz}$	-	0.22	0.5	-	0.22	0.5	-	0.22	0.5	Ω

#### **ELECTRICAL CHARACTERISTICS** (T<sub>A</sub> = 25°C, unless otherwise noted.)

Product parametric performance is indicated in the Electrical Characteristics for the listed test conditions, unless otherwise noted. Product performance may not be indicated by the Electrical Characteristics if operated under different conditions.

2. T<sub>low</sub> = -40°C for NCP431AI, NCP431AV, SC431AV

= 0°C for NCP431AC

T<sub>high</sub> = 70°C for NCP431AC

= 85°C for NCP431AI = 125°C for NCP431AV, SC431AV

3. Guaranteed by design

4. The deviation parameter  $\Delta V_{refT}$  is defined as the difference between the maximum and minimum values obtained over the full operating ambient temperature range that applies.



The average temperature coefficient of the reference input voltage, Vref is defined as:

$$v_{\text{ref}} \frac{\text{ppm}}{^{\circ}\text{C}} = \frac{\left(\frac{\Delta V_{\text{ref}}}{V_{\text{ref}} @25^{\circ}\text{C}}\right) \times 10^{6}}{\Delta T_{\text{A}}} = \frac{\Delta V_{\text{ref}} \times 10^{6}}{\Delta T_{\text{A}} (V_{\text{ref}} @25^{\circ}\text{C})}$$

aVref can be positive or negative depending on whether Vref Min or Vref Max occurs at the lower ambient temperature.

Example:  $\Delta V_{refT}$  = 17 mV and slope is positive  $V_{ref} = 2.5 \text{ V}, \Delta T_A = 165^{\circ}\text{C} \text{ (from } -40^{\circ}\text{C to } +125^{\circ}\text{C} \text{)}$ 

$$\alpha V_{ref} = \frac{0.017 \cdot 10^{6}}{165 \cdot 2.5} = 41.2 \text{ ppm/}^{\circ}\text{C}$$

- The dynamic impedance Z<sub>KA</sub> is defined as: (|Z<sub>KA</sub>| = (ΔV<sub>KA</sub>/ΔI<sub>K</sub>). When the device is programmed with two external resistors, R1 and R2, the total dynamic impedance of the circuit is defined as: |Z<sub>KA</sub>'| ≈ |Z<sub>KA</sub>| (1 + (R1/R2)).
  SC431AVSNT1G T<sub>low</sub> = -40°C, T<sub>high</sub> = 125°C. Guaranteed by design. SC Prefix for Automotive and Other Applications Requiring Unique
- Site and Control Change Requirements; AEC-Q100 Qualified and PPAP Capable.

			NCP431BC NCP432BC		NCP431BI NCP432BI		NCP/SC431BV NCP/SC432BV				
Symbol	Characteristic	Min	Тур	Max	Min	Тур	Мах	Min	Тур	Max	Unit
V <sub>ref</sub>	Reference Input Voltage $V_{KA} = V_{ref}, I_K = 1 \text{ mA}$ $T_A = 25^{\circ}C$ $T_A = T_{low}$ to $T_{high}$ (Figure 3, Note 7)				2.4875 2.4775						V
$\Delta V_{refT}$	Reference Input Voltage Deviation Over Temperature Range (Figure 3, Notes 8, 9) $V_{KA}$ = $V_{ref}$ , $I_{K}$ = 1 mA		-	-	-	5.0	10 1–		10	15 15	mV
$\frac{\Delta V_{\text{ref}}}{\Delta V_{\text{KA}}}$	Ratio of Change in Reference Input Voltage to Change in Cathode to Anode Voltage $I_K = 1 \text{ mA}$ (Figure 4), $\Delta V_{KA} = 10 \text{ V to } V_{ref}$ $\Delta V_{KA} = 36 \text{ V to } 10 \text{ V}$		-1.85 -0.80			-1.85 -0.80	-3.1 -1.8		-1.85 -0.80	-3.1 -1.8	mV/ V
I <sub>ref</sub>	Reference Input Current (Figure 4) I <sub>K</sub> = 1 mA, R1 = 220 k, R2 = $\infty$ T <sub>A</sub> = -40°C to +125°C	-	81	190	-	81	190	-	81	190	nA
$\Delta I_{refT}$	Reference Input Current Deviation Over Temperature Range (Figure 4, Note 8) $I_K = 1 \text{ mA}, R1 = 10 \text{ k}, R2 = \infty$	-	22	55	-	22	55	-	22	55	nA
I <sub>min</sub>	Minimum Cathode Current For Regulation $V_{KA} = V_{ref}$ (Figure 3)	-	40	60	_	40	60	-	40	60	μΑ
I <sub>off</sub>	Off–State Cathode Current (Figure 5) $V_{KA} = 36 V$ , $V_{ref} = 0 V$	-	180	1000	-	180	1000	-	180	1000	nA
Z <sub>KA</sub>	Dynamic Impedance (Figure 3, Note 10) $V_{KA} = V_{ref}, \Delta I_K = 1.0 \text{ mA to } 100 \text{ mA}$ f $\leq 1.0 \text{ kHz}$	-	0.22	0.5	-	0.22	0.5	-	0.22	0.5	Ω

#### FLECTRICAL CHARACTERISTICS (T. - 25°C unless otherwise noted)

Product parametric performance is indicated in the Electrical Characteristics for the listed test conditions, unless otherwise noted. Product performance may not be indicated by the Electrical Characteristics if operated under different conditions.

7. T<sub>Iow</sub> = -40°C for NCP431BI, NCP431BV, NCP432BI, NCP432BV, SC431B, SC432B

= 0°C for NCP431BC, NCP432BC

T<sub>high</sub> = 70°C for NCP431BC, NCP432BC

= 85°C for NCP431BI, NCP432BI

= 125°C for NCP431BV, NCP432BV, SC431BV, SC432BV

8. Guaranteed by design

9. The deviation parameter  $\Delta V_{refT}$  is defined as the difference between the maximum and minimum values obtained over the full operating ambient temperature range that applies.



The average temperature coefficient of the reference input voltage, Vref is defined as:

$$v_{\text{ref}} \frac{\text{ppm}}{^{\circ}\text{C}} = \frac{\left(\frac{\Delta V_{\text{ref}}}{V_{\text{ref}} @25^{\circ}\text{C}}\right) \times 10^{6}}{\Delta T_{\text{A}}} = \frac{\Delta V_{\text{ref}} \times 10^{6}}{\Delta T_{\text{A}} (V_{\text{ref}} @25^{\circ}\text{C})}$$

aVref can be positive or negative depending on whether Vref Min or Vref Max occurs at the lower ambient temperature.

Example:  $\Delta V_{refT}$  = 17 mV and slope is positive  $V_{ref} = 2.5 \text{ V}, \Delta T_A = 165^{\circ}\text{C} \text{ (from } -40^{\circ}\text{C to } +125^{\circ}\text{C} \text{)}$ 

$$\alpha V_{ref} = \frac{0.017 \cdot 10^{6}}{165 \cdot 2.5} = 41.2 \text{ ppm/}^{\circ} \text{C}$$

- 10. The dynamic impedance  $Z_{KA}$  is defined as: ( $|Z_{KA}| = (\Delta V_{KA}/\Delta I_K)$ . When the device is programmed with two external resistors, R1 and R2,
- the total dynamic impedance of the circuit is defined as: |Z<sub>KA</sub>| ≈ |Z<sub>KA</sub>| ≈ |Z<sub>KA</sub>| (1 + (R1/R2))
  SC431BVSNT1G, SC432BVSNT1G T<sub>low</sub> = -40°C, T<sub>high</sub> = 125°C. Guaranteed by design. SC Prefix for Automotive and Other Applications Requiring Unique Site and Control Change Requirements; AEC–Q100 Qualified and PPAP Capable.



Figure 3. Test Circuit for  $V_{KA} = V_{ref}$ 

Figure 4. Test Circuit for V<sub>KA</sub> > V<sub>ref</sub> Figure

Figure 5. Test Circuit for Ioff













Figure 20. Test Circuit For Curve A of Stability Boundary Conditions



Figure 21. Test Circuit For Curve B And C of Stability Boundary Conditions

## **TYPICAL APPLICATIONS**



Figure 22. Shunt Regulator







Figure 26. Constant Current Source



Figure 23. High Current Shunt Regulator



Figure 25. Series Pass Regulator



Figure 27. Constant Current Sink



Figure 28. Triac Crowbar



Figure 29. SRC Crowbar

V<sub>OUT</sub>

V+

≈2.0 V







**Temperature–Compensated Threshold** 



Figure 32. Step–Down Switching Converter

#### **APPLICATIONS INFORMATION**

The NCP431/NCP432 is a programmable precision reference which is used in a variety of ways. It serves as a reference voltage in circuits where a non-standard reference voltage is needed. Other uses include feedback control for driving an optocoupler in power supplies, voltage monitor, constant current source, constant current sink and series pass regulator. In each of these applications, it is critical to maintain stability of the device at various operating currents and load capacitances. In some cases the circuit designer can estimate the stabilization capacitance from the stability boundary conditions curve provided in Figure 18. However, these typical curves only provide stability information at specific cathode voltages and at a specific load condition. Additional information is needed to determine the capacitance needed to optimize phase margin or allow for process variation.

A simplified model of the NCP431/NCP432 is shown in Figure 33. When tested for stability boundaries, the load resistance is 150  $\Omega$ . The model reference input consists of an input transistor and a dc emitter resistance connected to the device anode. A dependent current source, Gm, develops a current whose amplitude is determined by the difference between the 1.78 V internal reference voltage source and the input transistor emitter voltage. A portion of Gm flows through compensation capacitance, CP2. The voltage across CP2 drives the output dependent current source, Go, which is connected across the device cathode and anode.

Model component values are:

Vref = 1.78 V

 $Gm = 0.3 + 2.7 \exp(-IC/26 mA)$ 

where IC is the device cathode current and Gm is in mhos Go = 1.25 (Vcp2) µmhos.

Resistor and capacitor typical values are shown on the model. Process tolerances are  $\pm 20\%$  for resistors,  $\pm 10\%$  for capacitors, and  $\pm 40\%$  for transconductances.

An examination of the device model reveals the location of circuit poles and zeroes:

P1 = 
$$\frac{1}{2\pi R_{GM}C_{P1}} = \frac{1}{2\pi \cdot 1.0M \cdot 20 \text{ pF}} = 7.96 \text{ kHz}$$

P2 = 
$$\frac{1}{2\pi R_{P2}C_{P2}} = \frac{1}{2\pi \cdot 10M \cdot 0.265 \text{ pF}} = 60 \text{ kHz}$$

$$Z1 = \frac{1}{2\pi R_{Z1}C_{P1}} = \frac{1}{2\pi \cdot 15.9k \cdot 20 \text{ pF}} = 500 \text{ kHz}$$

In addition, there is an external circuit pole defined by the load:

$$\mathsf{P}_{\mathsf{L}} = \frac{1}{2\pi\mathsf{R}_{\mathsf{L}}\mathsf{C}_{\mathsf{L}}}$$

Also, the transfer dc voltage gain of the NCP431 is:

$$G = G_M R_{GM} Go R_L$$

Example 1:

I<sub>C</sub>=10 mA, R<sub>L</sub>= 230  $\Omega$ ,C<sub>L</sub>= 0. Define the transfer gain. The DC gain is:

$$G = G_M R_{GM} GoR_L = (2.138)(1.0M)(1.25\mu)(230)$$
  
= 615 = 56 dB

Loop gain = 
$$G \frac{8.25k}{8.25k + 15k} = 218 = 47 \text{ dB}$$

The resulting transfer function Bode plot is shown in Figure 34. The asymptotic plot may be expressed as the following equation:

$$Av = 615 \frac{\left(1 + \frac{jf}{500 \text{ kHz}}\right)}{\left(1 + \frac{jf}{8.0 \text{ kHz}}\right)\left(1 + \frac{jf}{60 \text{ kHz}}\right)}$$

The Bode plot shows a unity gain crossover frequency of approximately 600 kHz. The phase margin, calculated from the equation, would be  $55.9^{\circ}$ . This model matches the Open–Loop Bode Plot of Figure 15. The total loop would have a unity gain frequency of about 300 kHz with a phase margin of about 44°.



Figure 33. Simplified NCP431/NCP432 Device Model

#### NCP431/NCP432 OPEN-LOOP VOLTAGE GAIN VERSUS FREQUENCY



Figure 34. Example 1 Circuit Open Loop Gain Plot

Example 2.

 $I_C = 7.5$  mA,  $R_L = 2.2$  k $\Omega$ ,  $C_L = 0.01$   $\mu$ F. Cathode tied to reference input pin. An examination of the data sheet stability boundary curve (Figure 18) shows that this value of load capacitance and cathode current is on the boundary.

Define the transfer gain.

The DC gain is:

$$\begin{split} G &= ~G_M R_{GM} GoR_L = (2.138)(1.0M)(1.25\mu)(230) \\ &= ~6389 = ~76~dB \end{split}$$

The resulting open loop Bode plot is shown in Figure 35. The asymptotic plot may be expressed as the following equation:

$$Av = 615 \frac{\left(1 + \frac{jf}{500 \text{ kHz}}\right)}{\left(1 + \frac{jf}{8.0 \text{ kHz}}\right)\left(1 + \frac{jf}{60 \text{ kHz}}\right)\left(1 + \frac{jf}{7.2 \text{ kHz}}\right)}$$

Note that the transfer function now has an extra pole formed by the load capacitance and load resistance.

Note that the crossover frequency in this case is about 250 kHz, having a phase margin of about  $-46^{\circ}$ . Therefore, instability of this circuit is likely.

# NCP431/NCP432 OPEN-LOOP BODE PLOT WITH LOAD CAP



Figure 35. Example 2 Circuit Open Loop Gain Plot

With three poles, this system is unstable. The only hope for stabilizing this circuit is to add a zero. However, that can only be done by adding a series resistance to the output capacitance, which will reduce its effectiveness as a noise filter. Therefore, practically, in reference voltage applications, the best solution appears to be to use a smaller value of capacitance in low noise applications or a very large value to provide noise filtering and a dominant pole rolloff of the system.

The NCP431/NCP432 is often used as a regulator in secondary side of a switch mode power supply (SMPS).

The benefit of this reference is high and stable gain under low bias currents. Figure 36 shows dependence of the gain (dynamic impedance) on the bias current. Value of minimum cathode current that is needed to assure stable gain is 80  $\mu$ A maximum.



Figure 36. Knee of Reference

Regulator with TL431 or other references in secondary side of a SMPS needs bias resistor to increase cathode current to reach high and stable gain (refer to Figure 37). This bias resistor does not have to be used in regulator with NCP431/NCP432 thanks to its low minimum cathode current.



Figure 37. SMPS Secondary Side and Feedback Connection on Primary Side

The NCP431/NCP432 operates with very low leakage and reference input current. Sum of these currents is lower than 100 nA. Regulator with the NCP431/NCP432 minimizes parasitic power consumption.

The best way to achieve extremely low no-load consumption in SMPS applications is to use NCP431/NCP432 as regulator on the secondary side. The consumption is reduced by minimum parasitic consumption and very low bias current of NCP431/NCP432.



## **MARKING DIAGRAMS**



xxx M=

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(Note: Microdot may be in either location)

#### **ORDERING INFORMATION**

Device	Marking	Tolerance	Operating Temperature Range	Package	Shipping <sup>†</sup>
NCP431ACDR2G	AC	1%	_	SOIC-8 (Pb-Free)	2500 / Tape & Reel
NCP431ACSNT1G	VRF	1%		SOT-23-3 (Pb-Free)	3000 / Tape & Reel
NCP431BCSNT1G	VRJ	0.5%	0°C to 70°C	SOT-23-3 (Pb-Free)	3000 / Tape & Reel
NCP432BCSNT1G	VRM	0.5%		SOT-23-3 (Pb-Free)	3000 / Tape & Reel
NCP431ACLPRAG	ACLP	1%	-	TO-92 (TO-226) (Pb-Free)	2000 / Tape & Reel
NCP431AIDR2G	AI	1%		SOIC-8 (Pb-Free)	2500 / Tape & Reel
NCP431AISNT1G	VRG	1%		SOT-23-3 (Pb-Free)	3000 / Tape & Reel
NCP431BISNT1G	VRK	0.5%	–40°C to 85°C	SOT-23-3 (Pb-Free)	3000 / Tape & Reel
NCP432BISNT1G	VRN	0.5%		SOT-23-3 (Pb-Free)	3000 / Tape & Reel
NCP431AILPRAG	AILP	1%		TO–92 (TO–226) (Pb–Free)	2000 / Tape & Reel
NCP431AVDR2G	AV	1%		SOIC-8 (Pb-Free)	2500 / Tape & Reel
NCP431AVSNT1G / SC431AVSNT1G*	VRH	1%		SOT-23-3 (Pb-Free)	3000 / Tape & Reel
NCP431AVLPRAG	AVLP	1%		TO–92 (TO–226) (Pb–Free)	2000 / Tape & Reel
NCP431AVLPG	AVLP	1%	–40°C to 125°C -	TO-92 (TO-226) (Pb-Free)	2000 Units / Bag
NCP431BVSNT1G / SC431BVSNT1G*	VRL	0.5%		SOT-23-3 (Pb-Free)	3000 / Tape & Reel
NCP432BVSNT1G / SC432BVSNT1G*	VRP	0.5%		SOT-23-3 (Pb-Free)	3000 / Tape & Reel

+For information on tape and reel specifications, including part orientation and tape sizes, please refer to our Tape and Reel Packaging Specifications Brochure, BRD8011/D. \*SC Prefix for Automotive and Other Applications Requiring Unique Site and Control Change Requirements; AEC–Q100 Qualified and PPAP

Capable.

## PACKAGE DIMENSIONS

TO-92 (TO-226) CASE 29-11 **ISSUE AN** 





**BENT LEAD** 

D

SECTION X-X

J

STRAIGHT LEAD

- NOTES: 1. DIMENSIONING AND TOLERANCING PER ANSI Y14.5M, 1982. 2. CONTROLLING DIMENSION: INCH. 3. CONTOUR OF PACKAGE BEYOND DIMENSION R IS UNCONTROLLED. 4. LEAD DIMENSION IS UNCONTROLLED IN P AND BEYOND DIMENSION K MINIMUM.

	INC	HES	MILLIMETERS		
DIM	MIN	MAX	MIN	MAX	
Α	0.175	0.205	4.45	5.20	
В	0.170	0.210	4.32	5.33	
С	0.125	0.165	3.18	4.19	
D	0.016	0.021	0.407	0.533	
G	0.045	0.055	1.15	1.39	
Н	0.095	0.105	2.42	2.66	
J	0.015	0.020	0.39	0.50	
K	0.500		12.70		
L	0.250		6.35		
Ν	0.080	0.105	2.04	2.66	
Ρ		0.100		2.54	
R	0.115		2.93		
٧	0.135		3.43		

- NOTES: 1. DIMENSIONING AND TOLERANCING PER ASME Y14.5M, 1994. 2. CONTROLLING DIMENSION: MILLIMETERS. 3. CONTROLLING DIMENSION IS UNCONTROLLED. 4. LEAD DIMENSION IS UNCONTROLLED IN P AND BEYOND DIMENSION K MINIMUM.

	MILLIMETERS					
DIM	MIN	MAX				
Α	4.45	5.20				
В	4.32	5.33				
C	3.18	4.19				
D	0.40	0.54				
G	2.40	2.80				
J	0.39	0.50				
K	12.70					
Ν	2.04	2.66				
Ρ	1.50	4.00				
R	2.93					
V	3.43					





#### PACKAGE DIMENSIONS



\*For additional information on our Pb-Free strategy and soldering details, please download the ON Semiconductor Soldering and Mounting Techniques Reference Manual, SOLDERRM/D.

NOTES:

- NOTES:
  DIMENSIONING AND TOLERANCING PER ANSI Y14.5M, 1982.
  CONTROLLING DIMENSION: MILLIMETER.
  DIMENSION A AND B DO NOT INCLUDE MOLD PROTRUSION.
  MAXIMUM MOLD PROTRUSION 0.15 (0.006)
- MAXIMUM MOLD PROTRUSION 0.15 (0.006) PER SIDE.
  DIMENSION D DOES NOT INCLUDE DAMBAR PROTRUSION. ALLOWABLE DAMBAR PROTRUSION SHALL BE 0.127 (0.005) TOTAL IN EXCESS OF THE D DIMENSION AT MAXIMUM MATERIAL CONDITION.
  751–01 THRU 751–06 ARE OBSOLETE. NEW STANDARD IS 751–07
- STANDARD IS 751-07.

	MILLIN	IETERS	INC	HES				
DIM	MIN	MIN MAX		MAX				
Α	4.80	5.00	0.189	0.197				
в	3.80	4.00	0.150	0.157				
c	1.35	1.75	0.053	0.069				
D	0.33	0.51	0.013	0.020				
G	1.27	7 BSC	0.050 BSC					
н	0.10	0.25	0.004	0.010				
J	0.19	0.25	0.007	0.010				
ĸ	0.40	1.27	0.016	0.050				
м	0 °	8 °	0 °	8 °				
Z	0.25	0.50	0.010	0.020				
S	5 80	6.20	0.228	0 244				

#### PACKAGE DIMENSIONS

SOT-23 (TO-236) CASE 318-08 **ISSUE AR** 







NOTES:

- DIMENSIONING AND TOLERANCING PER ASME Y14.5M, 1994. CONTROLLING DIMENSION: MILLIMETERS. MAXIMUM LEAD THICKNESS INCLUDES LEAD FINISH. MINIMUM LEAD THICKNESS IS THE MINIMUM THICKNESS OF 3.
- THE BASE MATERIAL. DIMENSIONS D AND E DO NOT INCLUDE MOLD FLASH, 4. PROTRUSIONS, OR GATE BURRS.

	MILLIMETERS			INCHES		
DIM	MIN	NOM	MAX	MIN	NOM	MAX
Α	0.89	1.00	1.11	0.035	0.039	0.044
A1	0.01	0.06	0.10	0.000	0.002	0.004
b	0.37	0.44	0.50	0.015	0.017	0.020
С	0.08	0.14	0.20	0.003	0.006	0.008
D	2.80	2.90	3.04	0.110	0.114	0.120
E	1.20	1.30	1.40	0.047	0.051	0.055
е	1.78	1.90	2.04	0.070	0.075	0.080
L	0.30	0.43	0.55	0.012	0.017	0.022
L1	0.35	0.54	0.69	0.014	0.021	0.027
HE	2.10	2.40	2.64	0.083	0.094	0.104
Т	0°		10°	0°		10°



SEE VIEW C **END VIEW** 

#### RECOMMENDED SOLDERING FOOTPRINT\*



\*For additional information on our Pb-Free strategy and soldering details, please download the ON Semiconductor Soldering and Mounting Techniques Reference Manual, SOLDERRM/D.

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