

## High Speed Quad MOSFET Driver

### Features

- ▶ 6.0ns rise and fall time
- ▶ 2.0A peak output source/sink current
- ▶ 1.8 to 5.0V input CMOS compatible
- ▶ 5.0 to 12V total supply voltage
- ▶ Smart logic threshold
- ▶ Low jitter design
- ▶ Quad matched channels
- ▶ Drives two P- and two N-channel MOSFETs
- ▶ Outputs can swing below ground
- ▶ Low inductance quad flat no-lead package
- ▶ High-performance thermally-enhanced QFN

### Applications

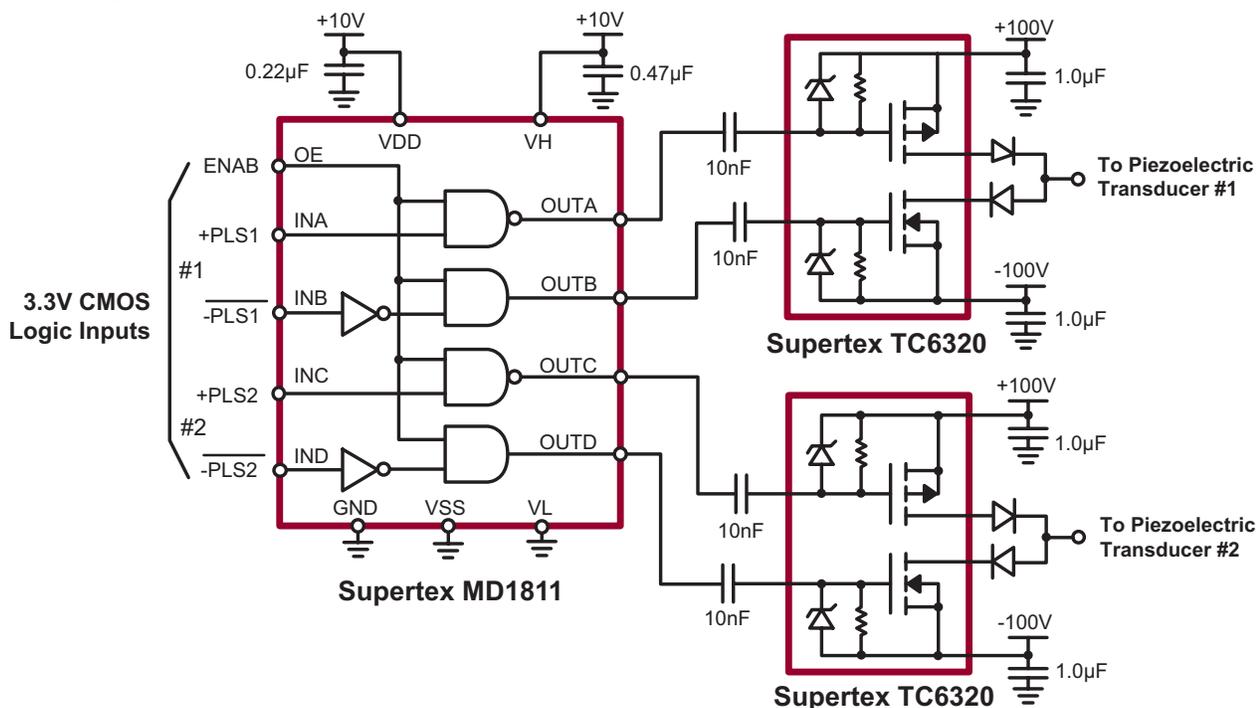
- ▶ Medical ultrasound imaging
- ▶ Piezoelectric transducer drivers
- ▶ Non-Destructive Testing (NDT)
- ▶ PIN diode driver
- ▶ Clock driver/buffer
- ▶ High speed level translator

### General Description

The Supertex MD1811 is a high speed, quad MOSFET driver designed to drive high voltage P and N-channel MOSFETs for medical ultrasound applications and other applications requiring a high output current for a capacitive load. The high-speed input stage of the MD1811 can operate from a 1.8 to 5.0V logic interface with an optimum operating input signal range of 1.8 to 3.3V. An adaptive threshold circuit is used to set the level translator switch threshold to the average of the input logic 0 and logic 1 levels. The input logic levels may be ground referenced, even though the driver is putting out bipolar signals. The level translator uses a proprietary circuit, which provides DC coupling together with high-speed operation.

The output stage of the MD1811 has separate power connections enabling the output signal L and H levels to be chosen independently from the supply voltages used for the majority of the circuit. As an example, the input logic levels may be 0 and 1.8V, the control logic may be powered by +5.0 and -5.0V, and the output L and H levels may be varied anywhere over the range of -5.0 to +5.0V. The output stage is capable of peak currents of up to  $\pm 2.0A$ , depending on the supply voltages used and load capacitance present. The OE pin serves a dual purpose. First, its logic H level is used to compute the threshold voltage level for the channel input level translators. Secondly, when OE is low, the outputs are disabled, with the A & C output high and the B & D output low. This assists in properly pre-charging the AC coupling capacitors that may be used in series in the gate drive circuit of an external PMOS and NMOS transistor pair.

### Typical Application Circuit



## Ordering Information

Device	16-Lead QFN 4.00x4.00mm body 1.00mm height (max) 0.65mm pitch
MD1811	MD1811K6-G

-G indicates package is RoHS compliant ("Green")



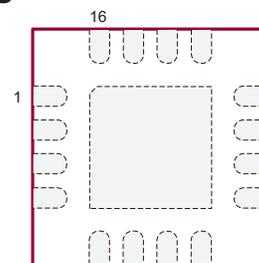
## Absolute Maximum Ratings

Parameter	Value
$V_{DD} - V_{SS}$ , Logic supply voltage	-0.5V to +13.5V
$V_H$ , Output high supply voltage	$V_L - 0.5V$ to $V_{DD} + 0.5V$
$V_L$ , Output low supply voltage	$V_{SS} - 0.5V$ to $V_H + 0.5V$
$V_{SS}$ , Low side supply voltage	-7.0V to +0.5V
Logic input levels	$V_{SS} - 0.5V$ to GND +7.0V
Maximum junction temperature	+125°C
Storage temperature	-65°C to 150°C
Operating temperature	-20°C to +85°C
Package power dissipation	2.2W
Thermal resistance ( $\theta_{JA}$ )*	45°C/W

Absolute Maximum Ratings are those values beyond which damage to the device may occur. Functional operation under these conditions is not implied. Continuous operation of the device at the absolute rating level may affect device reliability. All voltages are referenced to device ground.

\* 1.0oz 4-layer 3x4" PCB

## Pin Configuration



16-Lead QFN (K6)  
(top view)

## Product Marking



Y = Last Digit of Year Sealed  
W = Code for Week Sealed  
L = Lot Number  
— = "Green" Packaging

Package may or may not include the following marks: Si or

16-Lead QFN (K6)

## DC Electrical Characteristics ( $V_H = V_{DD} = 12V$ , $V_L = V_{SS} = GND = 0V$ , $V_{OE} = 3.3V$ , $T_A = 25^\circ C$ )

Sym	Parameter	Min	Typ	Max	Units	Conditions
$V_{DD} - V_{SS}$	Logic supply voltage	4.5	-	13	V	$2.5V \leq V_{DD} \leq 13V$
$V_{SS}$	Low side supply voltage	-5.5	-	0	V	--
$V_H$	Output high supply voltage	$V_{SS} + 2.0$	-	$V_{DD}$	V	--
$V_L$	Output low supply voltage	$V_{SS}$	-	$V_{DD} - 2.0$	V	--
$I_{DDQ}$	$V_{DD}$ quiescent current	-	0.8	-	mA	No input transitions, OE = 1
$I_{HQ}$	$V_H$ quiescent current	-	-	10	$\mu A$	
$I_{DD}$	$V_{DD}$ average current	-	8.0	-	mA	One channel on at 5.0Mhz, No load
$I_H$	$V_H$ average current	-	26	-	mA	
$V_{IH}$	Input logic voltage high	$V_{OE} - 0.3$	-	5.0	V	For logic inputs INA, INB, INC, and IND
$V_{IL}$	Input logic voltage low	0	-	0.3	V	
$I_{IH}$	Input logic current high	-	-	1.0	$\mu A$	
$I_{IL}$	Input logic current low	-	-	1.0	$\mu A$	
$V_{IH}$	OE Input logic voltage high	1.7	-	5.0	V	For logic input OE
$V_{IL}$	OE Input logic voltage low	0	-	0.3	V	
$R_{IN}$	Input logic impedance to GND	10	20	30	K $\Omega$	
$C_{IN}$	Logic input capacitance	-	5.0	10	pF	---

**Outputs** ( $V_H = V_{DD} = 12V$ ,  $V_L = V_{SS} = GND = 0V$ ,  $V_{OE} = 3.3V$ ,  $T_A = 25^\circ C$ )

Sym	Parameter	Min	Typ	Max	Units	Conditions
$R_{SINK}$	Output sink resistance	-	-	12.5	$\Omega$	$I_{SINK} = 50mA$
$R_{SOURCE}$	Output source resistance	-	-	12.5	$\Omega$	$I_{SOURCE} = 50mA$
$I_{SINK}$	Peak output sink current	-	2.0	-	A	--
$I_{SOURCE}$	Peak output source current	-	2.0	-	A	--

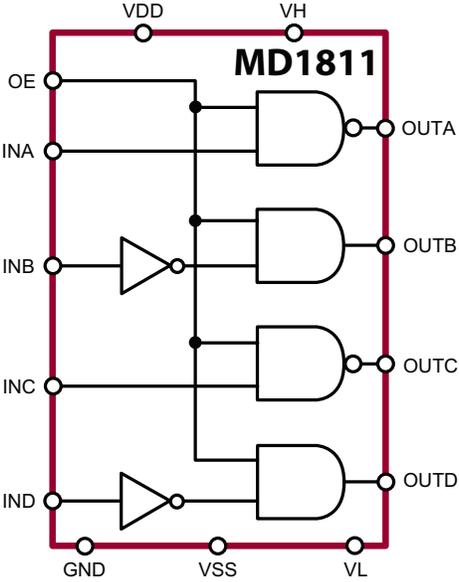
**AC Electrical Characteristics** ( $V_H = V_{DD} = 12V$ ,  $V_L = V_{SS} = GND = 0V$ ,  $V_{OE} = 3.3V$ ,  $T_A = 25^\circ C$ )

Sym	Parameter	Min	Typ	Max	Units	Conditions
$t_{ir}$	Input or OE rise & fall time	-	-	10	ns	Logic input edge speed requirement
$t_{PLH}$	Propagation delay when output is from low to high	-	7.0	-	ns	$C_{LOAD} = 1000pF$ , see timing diagram
$t_{PHL}$	Propagation delay when output is from high to low	-	7.0	-	ns	
$t_r$	Output rise time	-	6.0	-	ns	Input signal rise/fall time 2ns
$t_f$	Output fall time	-	6.0	-	ns	
$ t_r - t_f $	Rise and fall time matching	-	1.0	-	ns	For each channel
$ t_{PLH} - t_{PHL} $	Propagation low to high and high to low matching	-	1.0	-	ns	
$\Delta t_{dm}$	Propagation delay matching	-	$\pm 2.0$	-	ns	Device to device delay match
$t_{OE}$	Output enable time	-	9.0	-	ns	---

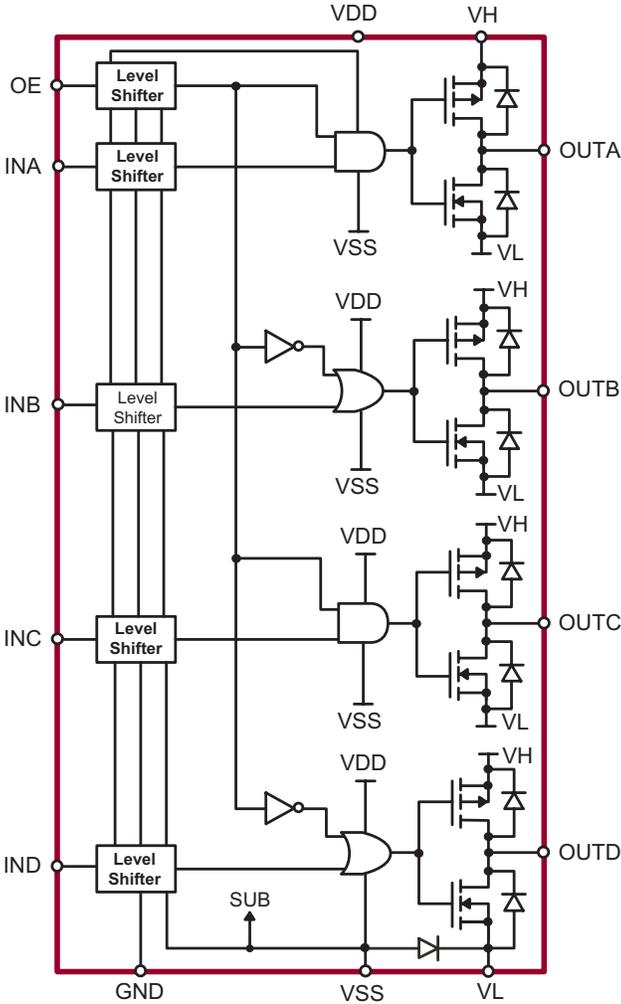
**Logic Truth Table**

Logic Inputs			Output	
OE	INA	INB	OUTA	OUTB
H	L	L	$V_H$	$V_H$
H	L	H	$V_H$	$V_L$
H	H	L	$V_L$	$V_H$
H	H	H	$V_L$	$V_L$
L	X	X	$V_H$	$V_L$
OE	INC	IND	OUTC	OUTD
H	L	L	$V_H$	$V_H$
H	L	H	$V_H$	$V_L$
H	H	L	$V_L$	$V_H$
H	H	H	$V_L$	$V_L$
L	X	X	$V_H$	$V_L$

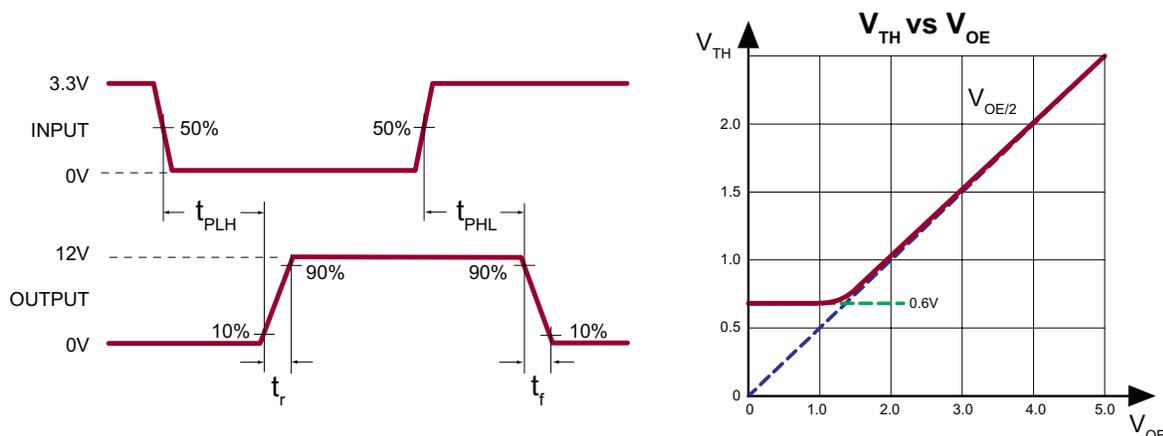
Simplified Block Diagram



Detailed Block Diagram



## Timing Diagram and $V_{TH} / V_{OE}$ Curve



## Application Information

For proper operation of the MD1811, low inductance bypass capacitors should be used on the various supply pins. The GND pin should be connected to the logic ground. The INA, INB, INC, IND, and OE pins should be connected to a logic source with a swing of GND to OE, where OE is 1.8 to 5.0V. Good trace practices should be followed corresponding to the desired operating speed. The internal circuitry of the MD1811 is capable of operating up to 100MHz, with the primary speed limitation being the loading effects of the load capacitance. Because of this speed and the high transient currents that result with capacitive loads, the bypass capacitors should be as close to the chip pins as possible. Unless the load specifically requires bipolar drive, the VSS and VL pins should have low inductance feed-through connections directly to a ground plane. If these voltages are not zero, then they need bypass capacitors in a manner similar to the positive power supplies. The power connection VDD should have a ceramic bypass capacitor to the ground plane with short leads and decoupling components to prevent resonance in the power leads.

The voltages of VH and VL decide the output signal levels. These two pins can draw fast transient currents of up to 2.0A, so they should be provided with an appropriate bypass capacitor located next to the chip pins. A ceramic capacitor

of up to 1.0 $\mu$ F may be appropriate, with a series ferrite bead to prevent resonance in the power supply lead coming to the capacitor. Pay particular attention to minimizing trace lengths, current loop area and using sufficient trace width to reduce inductance. Surface mount components are highly recommended. Since the output impedance of this driver is very low, in some cases it may be desirable to add a small series resistance in series with the output signal to obtain better waveform transitions at the load terminals. This will of course reduce the output voltage slew rate at the terminals of a capacitive load.

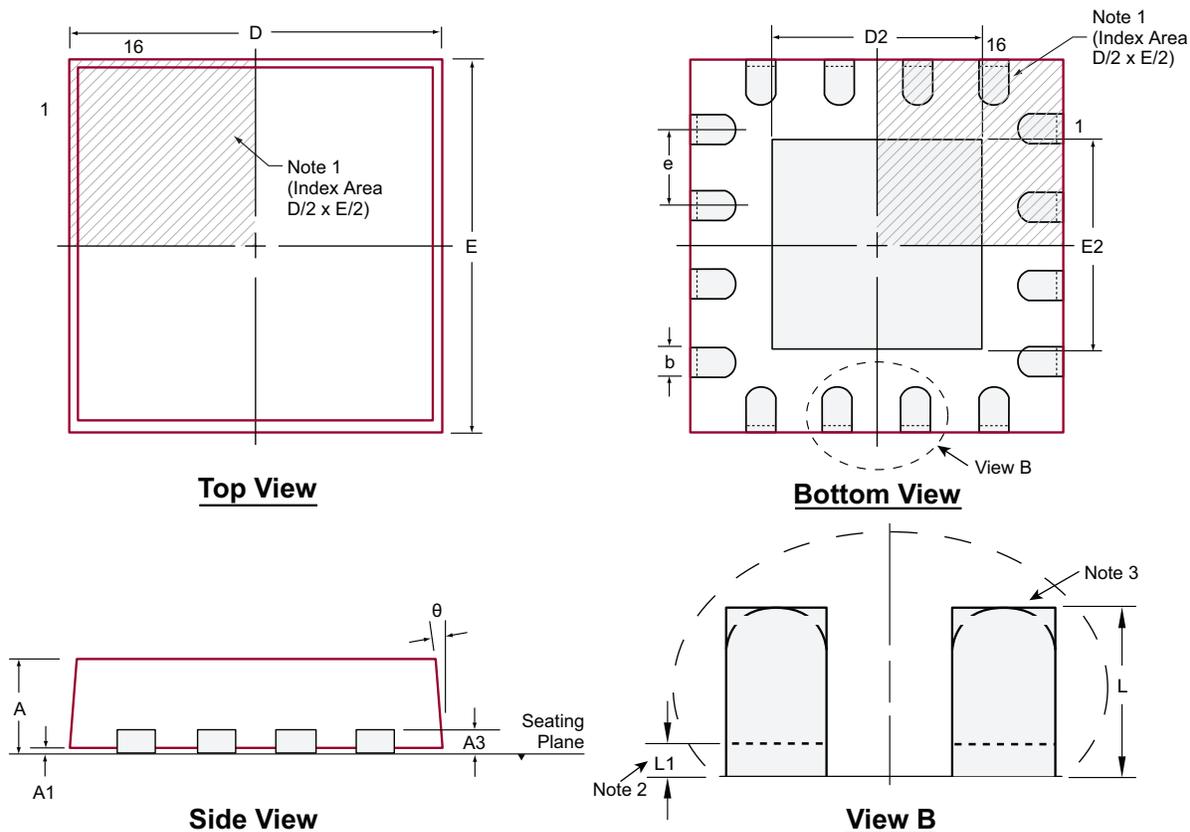
Pay particular attention that parasitic couplings are minimized from the output to the input signal terminals. The parasitic feedback may cause oscillations or spurious waveform shapes on the edges of signal transitions. Since the input operates with signals down to 1.8V even small coupled voltages may cause problems. Use of a solid ground plane and good power and signal layout practices will prevent this problem. Be careful that a circulating ground return current from a capacitive load cannot react with common inductance to cause noise voltages in the input logic circuitry.

## Pin Description

Pin #	Function	Description
1	INB	Logic input. Input logic low will cause the output to swing to VH. Input logic high will cause the output to swing to VL. Keep all logic inputs low until IC powered up.
2	VL	Supply voltage for N-channel output stage.
3	GND	Logic input ground reference.
4	VL	Supply voltage for N-channel output stage.
5	INC	Logic input. Input logic low will cause the output to swing to VH. Input logic high will cause the output to swing to VL. Keep all logic inputs low until IC powered up.
6	IND	
7	VSS	Low side supply voltage. VSS is also connected to the IC substrate. It is required to connect to the most negative potential of voltage supplies and powered-up first.
8	OUTD	Output driver. Swings from VH to VL. Intended to drive the gate of an external N-channel MOSFET via a series capacitor. When OE is low, the output is disabled. OUTD will swing to VL turning off the external N-channel MOSFET.
9	OUTC	Output driver. Swings from VH to VL. Intended to drive the gate of an external P-channel MOSFET via a series capacitor. When OE is low, the output is disabled. OUTC will swing to VH turning off the external P-channel MOSFET.
10, 11	VH	Supply voltage for P-channel output stage.
12	OUTB	Output driver. Swings from VH to VL. Intended to drive the gate of an external N-channel MOSFET via a series capacitor. When OE is low, the output is disabled. OUTB will swing to VL turning off the external N-channel MOSFET.
13	OUTA	Output driver. Swings from VH to VL. Intended to drive the gate of an external P-channel MOSFET via a series capacitor. When OE is low, the output is disabled. OUTA will swing to VH turning off the external P-channel MOSFET.
14	VDD	High side supply voltage.
15	INA	Logic input. Input logic low will cause the output to swing to VH. Input logic high will cause the output to swing to VL. Keep all logic inputs low until IC powered up.
16	OE	Output-Enable logic input. When OE is high, $(V_{OE} + V_{GND})/2$ sets the logic threshold level for inputs, When OE is low, OUTA and OUTC are at VH, OUTB and OUTD are at VL, regardless of the inputs INA, INB, INC or IND. Keep OE low until IC powered up.
Substrate		The IC substrate is internally connected to the thermal pad. Thermal Pad and VSS must be connected externally.

# 16-Lead QFN Package Outline (K6)

4.00x4.00mm body, 1.00mm height (max), 0.65mm pitch



**Notes:**

1. A Pin 1 identifier must be located in the index area indicated. The Pin 1 identifier can be: a molded mark/identifier; an embedded metal marker; or a printed indicator.
2. Depending on the method of manufacturing, a maximum of 0.15mm pullback (L1) may be present.
3. The inner tip of the lead may be either rounded or square.

Symbol		A	A1	A3	b	D	D2	E	E2	e	L	L1	$\theta$
Dimension (mm)	MIN	0.80	0.00	0.20 REF	0.25	3.85*	2.50	3.85*	2.50	0.65 BSC	0.30†	0.00	0°
	NOM	0.90	0.02		0.30	4.00	2.65	4.00	2.65		0.40†	-	-
	MAX	1.00	0.05		0.35	4.15*	2.80	4.15*	2.80		0.50†	0.15	14°

JEDEC Registration MO-220, Variation VGGC-2, Issue K, June 2006.

\* This dimension is not specified in the JEDEC drawing.

† This dimension differs from the JEDEC drawing.

**Drawings not to scale.**

**Supertex Doc.#:** DSPD-16QFNK64X4P065, Version C041009.

(The package drawing(s) in this data sheet may not reflect the most current specifications. For the latest package outline information go to <http://www.supertex.com/packaging.html>.)

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